INTEGRATED CIRCUITS

DEVICE SPECIFICATION

TDA935X/6X/8X PS/N2 series TV signal processor-Teletext decoder with embedded μ -Controller

Tentative Device Specification

2001 Jan 18

Version: 2.8 Previous date: 2000 Nov 29





TDA935X/6X/8X PS/N2 series

GENERAL DESCRIPTION

The various versions of the TDA935X/6X/8X PS/N2 series combine the functions of a TV signal processor together with a $\mu\text{-Controller}$ and US Closed Caption decoder. Most versions have a Teletext decoder on board. The Teletext decoder has an internal RAM memory for 1or 10 page text. The ICs are intended to be used in economy television receivers with 90° and 110° picture tubes.

The ICs have supply voltages of 8 V and 3.3 V and they are mounted in S-DIP envelope with 64 pins.

The features are given in the following feature list. The differences between the various ICs are given in the table on page 4.



FEATURES

TV-signal processor

- Multi-standard vision IF circuit with alignment-free PLL demodulator
- Internal (switchable) time-constant for the IF-AGC circuit
- A choice can be made between versions with mono intercarrier sound FM demodulator and versions with QSS IF amplifier.
- The mono intercarrier sound versions have a selective FM-PLL demodulator which can be switched to the different FM sound frequencies (4.5/5.5/6.0/6.5 MHz).
 The quality of this system is such that the external band-pass filters can be omitted.
- Source selection between 'internal' CVBS and external CVBS or Y/C signals
- Integrated chrominance trap circuit
- Integrated luminance delay line with adjustable delay time
- Picture improvement features with peaking (with variable centre frequency and positive/negative overshoot ratio) and black stretching
- Integrated chroma band-pass filter with switchable centre frequency
- Only one reference (12 MHz) crystal required for the μ-Controller, Teletext- and the colour decoder
- PAL/NTSC or multi-standard colour decoder with automatic search system
- · Internal base-band delay line

- RGB control circuit with 'Continuous Cathode Calibration', white point and black level offset adjustment so that the colour temperature of the dark and the light parts of the screen can be chosen independently.
- Linear RGB or YUV input with fast blanking for external RGB/YUV sources. The Text/OSD signals are internally supplied from the μ-Controller/Teletext decoder
- Contrast reduction possibility during mixed-mode of OSD and Text signals
- Horizontal synchronization with two control loops and alignment-free horizontal oscillator
- Vertical count-down circuit
- Vertical driver optimized for DC-coupled vertical output stages
- Horizontal and vertical geometry processing
- Horizontal and vertical zoom function for 16:9 applications
- Horizontal parallelogram and bow correction for large screen picture tubes
- · Low-power start-up of the horizontal drive circuit

TDA935X/6X/8X PS/N2 series

μ-Controller

- 80C51 μ-controller core standard instruction set and timing
- 1 µs machine cycle
- 32 128Kx8-bit late programmed ROM
- 3 12Kx8-bit Auxiliary RAM (shared with Display and Acquisition)
- Interrupt controller for individual enable/disable with two level priority
- Two 16-bit Timer/Counter registers
- One 16 bit Timer with 8-bit Pre-scaler
- · WatchDog timer
- Auxiliary RAM page pointer
- 16-bit Data pointer
- · Stand-by, Idle and Power Down (PD) mode
- 14 bits PWM for Voltage Synthesis Tuning
- 8-bit A/D converter
- 4 pins which can be programmed as general I/O pin, ADC input or PWM (6-bit) output

Data Capture

- Text memory for 0, 1 or 10 pages
- In the 10 page versions inventory of transmitted Teletext pages stored in the Transmitted Page Table (TPT) and Subtitle Page Table (SPT)
- · Data Capture for US Closed Caption
- Data Capture for 525/625 line WST, VPS (PDC system A) and Wide Screen Signalling (WSS) bit decoding
- Automatic selection between 525 WST/625 WST
- Automatic selection between 625 WST/VPS on line 16 of VRI
- Real-time capture and decoding for WST Teletext in Hardware, to enable optimized μ-processor throughput
- · Automatic detection of FASTEXT transmission
- Real-time packet 26 engine in Hardware for processing accented, G2 and G3 characters
- Signal quality detector for video and WST/VPS data types
- Comprehensive teletext language coverage
- Full Field and Vertical Blanking Interval (VBI) data capture of WST data

Display

- · Teletext and Enhanced OSD modes
- Features of level 1.5 WST and US Close Caption
- · Serial and Parallel Display Attributes
- Single/Double/Quadruple Width and Height for characters
- Scrolling of display region
- · Variable flash rate controlled by software
- Enhanced display features including overlining, underlining and italics
- Soft colours using CLUT with 4096 colour palette
- Globally selectable scan lines per row (9/10/13/16) and character matrix [12x10, 12x13, 12x16 (VxH)]
- Fringing (Shadow) selectable from N-S-E-W direction
- · Fringe colour selectable
- · Meshing of defined area
- · Contrast reduction of defined area
- Cursor
- Special Graphics Characters with two planes, allowing four colours per character
- 32 software redefinable On-Screen display characters
- 4 WST Character sets (G0/G2) in single device (e.g. Latin, Cyrillic, Greek, Arabic)
- G1 Mosaic graphics, Limited G3 Line drawing characters
- WST Character sets and Closed Caption Character set in single device

IC VERSION (TDA)	9350	9351	9352	9353	9360	9361	9362	9363	9364	9365	9366	9367	9380	9381	9382	9383	9384	9385	9386	9387	9388	938
TV range	90°	90°	90°	110°	90°	90°	110°	110°	110°	110°	90°	90°	90°	90°	90°	110°	110°	110°	110°	90°	110°	110°
Mono intercarrier multi-standard sound demodulator (4.5 - 6.5 MHz) with switchable centre frequency	1	V		1	1	√	√	V					√	√		√	√			√	1	
Audio switch	√	√		√	√	V	V	√					V	V		√	V			√	√	
Automatic Volume Levelling	V	√	V		√	√					√	√	V	√	√					√		
Automatic Volume Levelling or subcarrier output (for comb filter applications)				√			√	V	√	√						√	V	√	√		√	1
QSS sound IF amplifier with separate input and AGC circuit			√						√	√	V	V			V			√	V			V
AM sound demodulator without extra reference circuit										√									V			
PAL decoder	√		√	√	√	√	$\sqrt{}$		√	√	√	√	$\sqrt{}$	√	√	√	√	√	√			
SECAM decoder		√	V	V		V		√		V		V		√	√		√		√			
NTSC decoder	V	√	V	√	√	√	V	√	V	√	√	V	V	√	√	√	√	V	√	√	√	√
Horizontal geometry (E-W)				√			$\sqrt{}$		√	√						√		√	√		√	
Horizontal and Vertical Zoom				√			$\sqrt{}$		√	√						√		√	√		√	
ROM size	32- 64 k	32- 64 k	32- 64 k	32- 64 k	64- 128k	16- 64 k																
User RAM size	1 k	1 k	1 k	1 k	2 k	2 k	2 k	2 k	2 k	2 k	2 k	2 k	1 k	1 k	1 k	1 k	1 k	1 k	1 k	1 k	1 k	1 k
Teletext	1 page	1 page	1 page	1 page	10 page																	
Closed captioning	√	√	√	√	√	√	$\sqrt{}$	√	√	√	√	√	√	√	√	√	√	V	√	√	√	

embedded µ-Controller TV signal processor-Teletext decoder with

TDA935X/6X/8X PS/N2 series

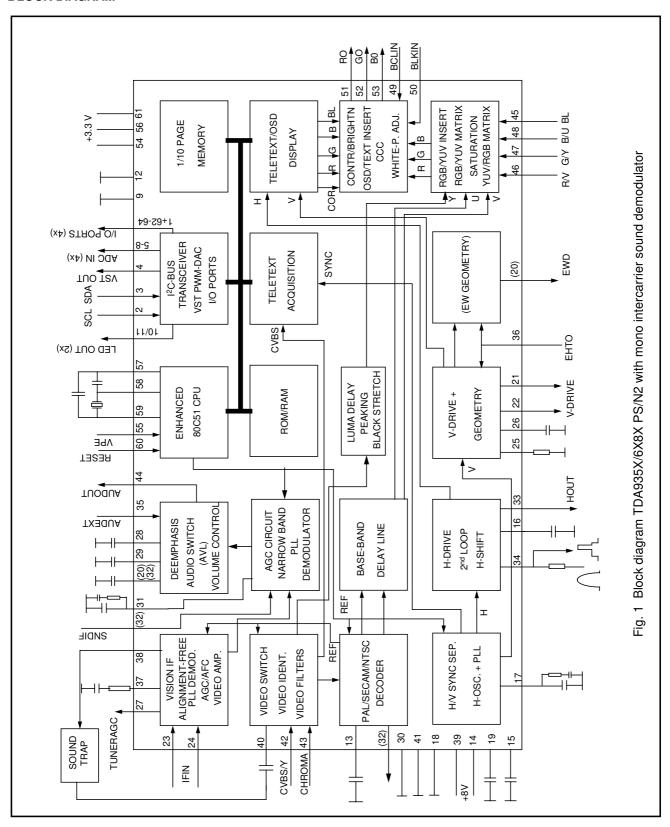
TDA935X/6X/8X PS/N2 series

QUICK REFERENCE DATA

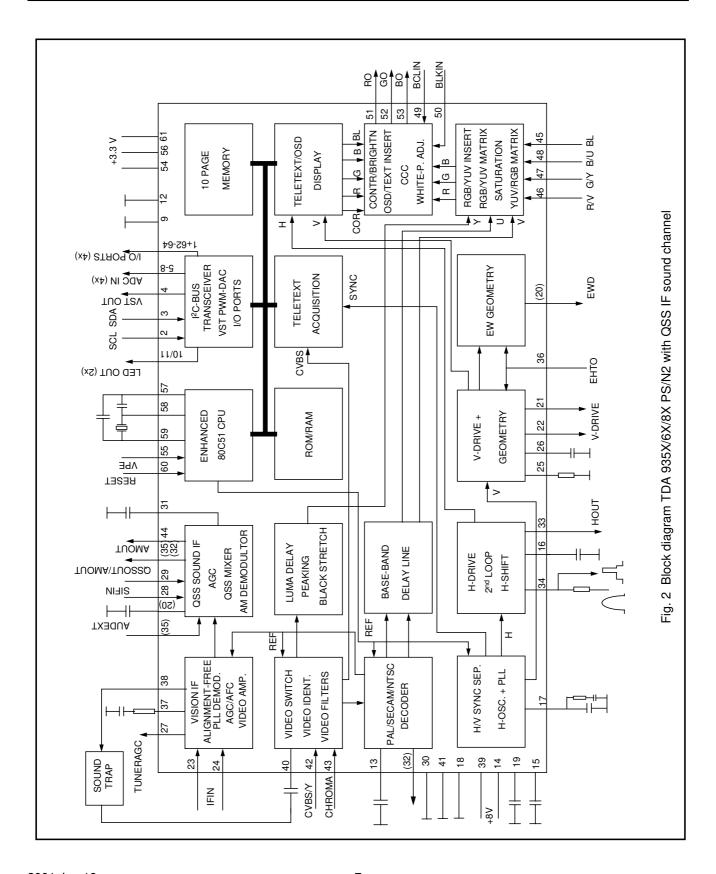
SYMBOL	PARAMETER	MIN.	TYP.	MAX.	UNIT
Supply		!	-!	!	!
V _P	supply voltages	_	8.0/3.3	_	٧
I _P	supply current	_	tbf	_	mA
Input voltage	s				
V _{iVIFrms)}	video IF amplifier sensitivity (RMS value)	_	75	_	μV
V _{iSIF(rms)}	QSS sound IF amplifier sensitivity (RMS value)	_	60	_	μV
V _{iAUDIO(rms)}	external audio input (RMS value)	_	500	_	mV
V _{iCVBS(p-p)}	external CVBS/Y input (peak-to-peak value)	_	1.0	_	V
$V_{\text{iCHROMA}(p-p)}$	external chroma input voltage (burst amplitude) (peak-to-peak value)	_	0.3	_	V
V _{iRGB(p-p)}	RGB inputs (peak-to-peak value)	_	0.7	_	V
V _{iYIN(p-p)}	luminance input signal (peak-to-peak value)	_	1.4	_	V
$V_{iUVIN(p-p)}$	U/V input signal (peak-to-peak value)	_	1.33/1.05	_	V
Output signal	s				
$V_{o(IFVO)(p-p)}$	demodulated CVBS output (peak-to-peak value)	_	2.5	_	V
V _{o(QSSO)(rms)}	sound IF intercarrier output in QSS versions (RMS value)	_	100	_	mV
$V_{o(AMOUT)(rms)}$	demodulated AM sound output in QSS versions (RMS value)	_	500	_	mV
I _{o(AGCOUT)}	tuner AGC output current range	0	_	5	mA
V _{oRGB(p-p)}	RGB output signal amplitudes (peak-to-peak value)	_	2.0	_	V
I _{oHOUT}	horizontal output current	10	_	_	mA
I _{oVERT}	vertical output current (peak-to-peak value)	1	_	_	mA
I _{oEWD}	EW drive output current	1.2	_	_	mA

TDA935X/6X/8X PS/N2 series

BLOCK DIAGRAM



TDA935X/6X/8X PS/N2 series



TDA935X/6X/8X PS/N2 series

PINNING

SYMBOL	PIN	DESCRIPTION
P1.3/T1	1	port 1.3 or Counter/Timer 1 input
P1.6/SCL	2	port 1.6 or I ² C-bus clock line
P1.7/SDA	3	port 1.7 or I ² C-bus data line
P2.0/TPWM	4	port 2.0 or Tuning PWM output
P3.0/ADC0	5	port 3.0 or ADC0 input
P3.1/ADC1	6	port 3.1 or ADC1 input
P3.2/ADC2	7	port 3.2 or ADC2 input
P3.3/ADC3	8	port 3.3 or ADC3 input
VSSC/P	9	digital ground for μ-Controller core and periphery
P0.5	10	port 0.5 (8 mA current sinking capability for direct drive of LEDs)
P0.6	11	port 0.6 (8 mA current sinking capability for direct drive of LEDs)
VSSA	12	analog ground of Teletext decoder and digital ground of TV-processor
SECPLL	13	SECAM PLL decoupling
VP2	14	2 nd supply voltage TV-processor (+8V)
DECDIG	15	decoupling digital supply of TV-processor
PH2LF	16	phase-2 filter
PH1LF	17	phase-1 filter
GND3	18	ground 3 for TV-processor
DECBG	19	bandgap decoupling
AVL/EWD (1)	20	Automatic Volume Levelling /East-West drive output
VDRB	21	vertical drive B output
VDRA	22	vertical drive A output
IFIN1	23	IF input 1
IFIN2	24	IF input 2
IREF	25	reference current input
VSC	26	vertical sawtooth capacitor
TUNERAGC	27	tuner AGC output
AUDEEM/SIFIN1(1)	28	audio deemphasis or SIF input 1
DECSDEM/SIFIN2(1)	29	decoupling sound demodulator or SIF input 2
GND2	30	ground 2 for TV processor
SNDPLL/SIFAGC ⁽¹⁾	31	narrow band PLL filter /AGC sound IF
AVL/SNDIF/REF0/ AMOUT ⁽¹⁾	32	Automatic Volume Levelling / sound IF input / subcarrier reference output /AM output (non controlled)
HOUT	33	horizontal output
FBISO	34	flyback input/sandcastle output
AUDEXT/ QSSO/AMOUT ⁽¹⁾	35	external audio input /QSS intercarrier out /AM audio output (non controlled)
EHTO	36	EHT/overvoltage protection input
PLLIF	37	IF-PLL loop filter
IFVO/SVO	38	IF video output / selected CVBS output
VP1	39	main supply voltage TV-processor (+8 V)
CVBSINT	40	internal CVBS input
GND1	41	ground 1 for TV-processor
CVBS/Y	42	external CVBS/Y input
CHROMA	43	chrominance input (SVHS)
AUDOUT /AMOUT ⁽¹⁾	44	audio output /AM audio output (volume controlled)

TDA935X/6X/8X PS/N2 series

SYMBOL	PIN	DESCRIPTION
INSSW2	45	2 nd RGB / YUV insertion input
R2/VIN	46	2 nd R input / V (R-Y) input
G2/YIN	47	2 nd G input / Y input
B2/UIN	48	2 nd B input / U (B-Y) input
BCLIN	49	beam current limiter input / (V-guard input, note 2)
BLKIN	50	black current input / (V-guard input, note 2)
RO	51	Red output
GO	52	Green output
ВО	53	Blue output
VDDA	54	analog supply of Teletext decoder and digital supply of TV-processor (3.3 V)
VPE	55	OTP Programming Voltage
VDDC	56	digital supply to core (3.3 V)
OSCGND	57	oscillator ground supply
XTALIN	58	crystal oscillator input
XTALOUT	59	crystal oscillator output
RESET	60	reset
VDDP	61	digital supply to periphery (+3.3 V)
P1.0/INT1	62	port 1.0 or external interrupt 1 input
P1.1/T0	63	port 1.1 or Counter/Timer 0 input
P1.2/INT0	64	port 1.2 or external interrupt 0 input

Note

- 1. The function of pin 20, 28, 29, 31, 32, 35 and 44 is dependent on the IC version (mono intercarrier FM demodulator / QSS IF amplifier and East-West output or not) and on some software control bits. The valid combinations are given in table 1.
- 2. The vertical guard function can be controlled via pin 49 or pin 50. The selction is made by means of the IVG bit in subaddress 2BH.

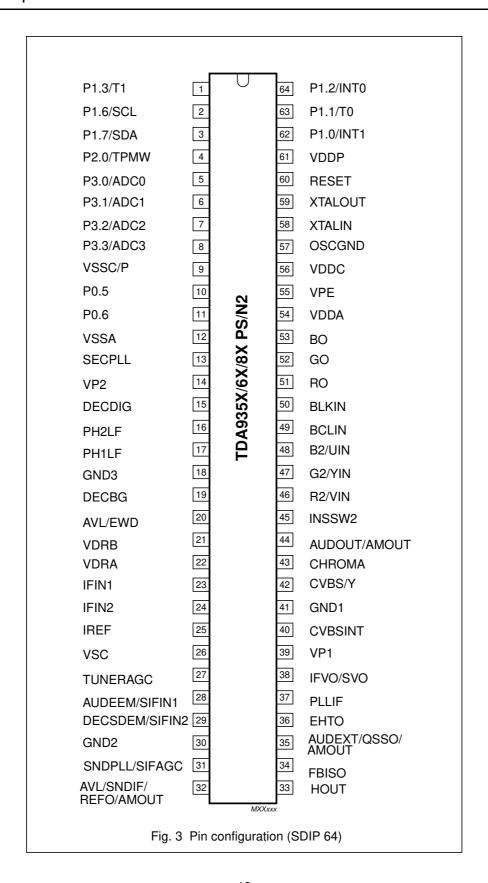
Table 1 Pin functions for various versions

IC version		FM-P	LL version		QSS version							
East-West Y/N		N	Y			N			Υ			
CMB1/CMB0 bits	00	01/10/11	00	01/10/11	00	01/	01/10/11		01/	10/11		
AM bit	_	_	_	_	_ 0 1 _ 0					1		
Pin 20	A	VL	EWD)	AVL EWD							
Pin 28		Al	JDEEM		SIFIN1							
Pin 29		DE	CSDEM		SIFIN2							
Pin 31		S	NDPLL				SIF	AGC				
Pin 32	SNDIF ⁽¹⁾	REFO ⁽²⁾	AVL/SNDIF(1)	REFO ⁽²⁾	AMOUT	REFO ⁽²⁾		AMOUT	RE	FO ⁽²⁾		
Pin 35		Α	UDEXT		AUDEXT QSSO AMOUT AUDEXT QSSO AMO					AMOUT		
Pin 44		Al	UDOUT		controlled AM or audio out							

Note

- 1. When additional (external) selectivity is required for FM-PLL system pin 32 can be used as sound IF input. This function is selected by means of SIF bit in subaddress 28H.
- 2. The reference output signal is only available for the CMB1/CMB0 setting of 0/1. For the other settings this pin is a switch output (see also table 67).

TDA935X/6X/8X PS/N2 series



TDA935X/6X/8X PS/N2 series

FUNCTIONAL DESCRIPTION OF THE 80C51

The functionality of the micro-controller used on this device is described here with reference to the industry standard 80C51 micro-controller. A full description of its functionality can be found in the 80C51 based 8-bit micro-controllers - Philips Semiconductors (ref. IC20).

Features of the 80c51

- 80C51 micro-controller core standard instruction set and timing.
- 1µs machine cycle.
- Maximum 128K x 8-bit Program ROM.
- Maximum of 12K x 8-bit Auxiliary RAM.
 - 2K (OSD only version) Auxiliary RAM, maximum of 1.25K required for Display
 - 3K (1 page teletext version) Auxiliary RAM, maximum of 2K required for Display
 - 12K (10 page teletext version) Auxiliary RAM, maximum of 10K required for Display
- 8-Level Interrupt Controller for individual enable/disable with two level priority.
- Two 16-bit Timer/Counters.
- Additional 16-bit Timer with 8-bit Pre-scaler.
- · WatchDog Timer.
- Auxiliary RAM Page Pointer.
- 16-bit Data pointer
- Idle, Stand-by and Power-Down modes.
- 13 General I/O.
- Four 6-bit Pulse Width Modulator (PWM) outputs for control of TV analogue signals.
- One 14-bit PWM for Voltage Synthesis tuner control.
- 8-bit ADC with 4 multiplexed inputs.
- 2 high current outputs for directly driving LED's etc.
- I²C Byte Level bus interface.

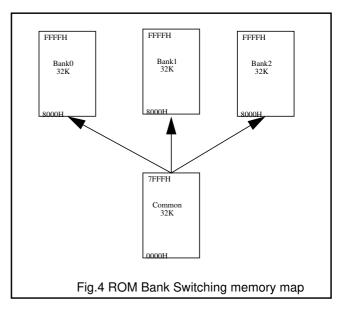
Memory Organisation

The device has the capability of a maximum of 128K Bytes of PROGRAM ROM and 12K Bytes of DATA RAM. The OSD (& Closed Caption) only version has a 2K RAM and a maximum of 64K ROM, the 1 page teletext version has a 3K RAM and also a maximum of 64K ROM whilst the 10 page teletext version has a 12K RAM and a maximum of 128K ROM.

ROM Organisation

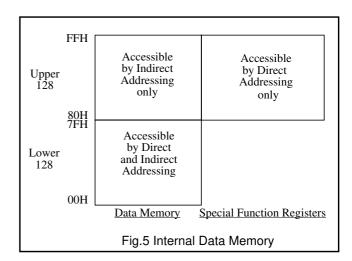
The 64K device has a continuous address space from 0 to 64K. The 128K is arranged in four banks of 32K. One of

the 32K banks is common and is always addressable. The other three banks (Bank0, Bank1, Bank2) can be accessed by selecting the right bank via the SFR ROMBK bits 1/0.



RAM Organisation

The Internal Data RAM is organised into two areas, Data Memory and Special Function Registers (SFRs) as shown in Fig.5.



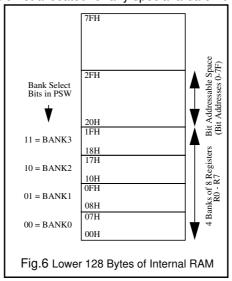
DATA MEMORY

The Data memory is 256 x 8-bits and occupies the address range 00 to FF Hex when using Indirect addressing and 00 to 7F Hex when using direct addressing. The SFRs occupy the address range 80 Hex to FF Hex and are accessible using Direct addressing only. The lower 128 Bytes of Data memory are mapped as shown in Fig.6. The lowest 32

2001 Jan 18

TDA935X/6X/8X PS/N2 series

bytes are grouped into 4 banks of 8 registers, the next 16 bytes above the register banks form a block of bit addressable memory space. The upper 128 bytes are not allocated for any special area or functions.



SFR MEMORY

The Special Function Register (SFR) space is used for port latches, counters/timers, peripheral control, data capture and display. These registers can only be accessed by direct addressing. Sixteen of the addresses in the SFR space are both bit and byte addressable. The bit addressable SFRs are those whose address ends in 0H or 8H. A summary of the SFR map in address order is shown in Table 2.

ADD	R/W	Names	BIT7	BIT6	BIT5	BIT4	BIT3	BIT2	BIT1	BIT0
80H	R/W	P0	Reserved	P0<6>	P0<5>	Reserved	Reserved	Reserved	Reserved	Reserved
81H	R/W	SP	SP<7>	SP<6>	SP<5>	SP<4>	SP<3>	SP<2>	SP<1>	SP<0>
82H	R/W	DPL	DPL<7>	DPL<6>	DPL<5>	DPL<4>	DPL<3>	DPL<2>	DPL<1>	DPL<0>
83H	R/W	DPH	DPH<7>	DPH<6>	DPH<5>	DPH<4>	DPH<3>	DPH<2>	DPH<1>	DPH<0>
84H	R/W	IEN1	-	-	-	-	-	-	-	ET2
85H	R/W	IP1	-	-	-	-	-	-	-	PT2
87H	R/W	PCON	0	ARD	RFI	WLE	GF1	GF0	PD	IDL
88H	R/W	TCON	TF1	TR1	TF0	TR0	IE1	IT1	IE0	IT0
89H	R/W	TMOD	GATE	C/T	M1	M0	GATE	C/T	M1	M0
8AH	R/W	TL0	TL0<7>	TL0<6>	TL0<5>	TL0<4>	TL0<3>	TL0<2>	TL0<1>	TL0<0>
8BH	R/W	TL1	TL1<7>	TL1<6>	TL1<5>	TL1<4>	TL1<3>	TL1<2>	TL1<1>	TL1<0>
8CH	R/W	тно	TH0<7>	TH0<6>	TH0<5>	TH0<4>	TH0<3>	TH0<2>	TH0<1>	TH0<0>
8DH	R/W	TH1	TH1<7>	TH1<6>	TH1<5>	TH1<4>	TH1<3>	TH1<2>	TH1<1>	TH1<0>
90H	R/W	P1	P1<7>	P1<6>	Reserved	Reserved	P1<3>	P1<2>	P1<1>	P1<0>
91H	R/W	TP2L	TP2L<7>	TP2L<6>	TP2L<5>	TP2L<4>	TP2L<3>	TP2L<2>	TP2L<1>	TP2L<0>

Table 2 SFR Map

TDA935X/6X/8X PS/N2 series

ADD	R/W	Names	BIT7	BIT6	BIT5	BIT4	BIT3	BIT2	BIT1	BIT0
92H	R/W	ТР2Н	TP2H<15>	TP2H<14>	TP2H<13>	TP2H<12>	TP2H<11>	TP2H<10>	TP2H<9>	TP2H<8>
93H	R/W	TP2PR	TP2PR<7>	TP2PR<6>	TP2PR<5>	TP2PR<4>	TP2PR<3>	TP2PR<2>	TP2PR<1>	TP2PR<0>
94H	R/W	TP2CRL	-	-	-	-	-	-	TP2CRL<1>	TP2CRL<0>
96H	R/W	P0CFGA	Reserved	P0CFGA<6>	P0CFGA<5>	Reserved	Reserved	Reserved	Reserved	Reserved
97H	R/W	P0CFGB	Reserved	P0CFGB<6>	P0CFGB<5>	Reserved	Reserved	Reserved	Reserved	Reserved
98H	R/W	SADB	-	-	-	DC_COMP	SAD<3>	SAD<2>	SAD<1>	SAD<0>
9СН	R	TP2CL	TP2CL<7>	TP2CL<6>	TP2CL<5>	TP2CL<4>	TP2CL<3>	TP2CL<2>	TP2CL<1>	TP2CL<0>
9DH	R	ТР2СН	TP2CH<7>	TP2CH<6>	TP2CH<5>	TP2CH<4>	TP2CH<3>	TP2CH<2>	TP2CH<1>	TP2CH<0>
9EH	R/W	P1CFGA	P1CFGA<7>	P1CFGA<6>	Reserved	Reserved	P1CFGA<3>	P1CFGA<2>	P1CFGA<1>	P1CFGA<0>
9FH	R/W	P1CFGB	P1CFGB<7>	P1CFGB<6>	Reserved	Reserved	P1CFGB<3>	P1CFGB<2>	P1CFGB<1>	P1CFGB<0>
АОН	R/W	P2	Reserved	-	-	-	-	-	-	P2<0>
А6Н	R/W	P2CFGA	Reserved	P2CFGA<6>	P2CFGA<5>	P2CFGA<4>	P2CFGA<3>	P2CFGA<2>	P2CFGA<1>	P2CFGA<0>
А7Н	R/W	P2CFGB	Reserved	P2CFGB<6>	P2CFGB<5>	P2CFGB<4>	P2CFGB<3>	P2CFGB<2>	P2CFGB<1>	P2CFGB<0>
A8H	R/W	IE	EA	EBUSY	ES2	ECC	ET1	EX1	ET0	EX0
ВОН	R/W	Р3	Reserved	Reserved	Reserved	Reserved	P3<3>	P3<2>	P3<1>	P3<0>
В2Н	R/W	TXT18	NOT<3>	NOT<2>	NOT<1>	NOT<0>	0	0	BS<1>	BS<0>
взн	R/W	TXT19	TEN	TC<2>	TC<1>	TC<0>	0	0	TS<1>	TS<0>
В4Н	R/W	TXT20	DRCS ENABLE	OSD PLANES	0	0	OSD LANG ENABLE	OSD LAN<2>	OSD LAN<1>	OSD LAN<0>
В5Н	R/W	TXT21	DISP LINE<1>	DISP LINES<0>	CHAR SIZE<1>	CHAR SIZE<0>	Reserved	CC ON	I2C PORT0	CC/TXT
В6Н	R	TXT22	GPF1<7>	GPF1<6>	GPF1<5>	GPF1<4>	GPF1<3>	GPF1<2>	GPF1<1>	GPF1<0>
В7Н	R/W	CCLIN	0	0	0	CS<4>	CS<3>	CS<2>	CS<1>	CS<0>
В8Н	R/W	IP	0	PBUSY	PES2	PCC	PT1	PX1	PT0	PX0
В9Н	R/W	TXT17	0	FORCE ACQ<1>	FORCE ACQ<0>	FORCE DISP<1>	FORCE DISP<0>	SCREEN COL<2>	SCREEN COL<1>	SCREEN COL<0>
ВАН	R	WSS1	0	0	0	WSS<3:0> ERROR	WSS<3>	WSS<2>	WSS<1>	WSS<0>
ввн	R	WSS2	0	0	0	WSS<7:4> ERROR	WSS<7>	WSS<6>	WSS<5>	WSS<4>
ВСН	R	WSS3	WSS<13:11> ERROR	WSS<13>	WSS<12>	WSS<11>	WSS<10:8> ERROR	WSS<10>	WSS<9>	WSS<8>
BEH	R/W	P3CFGA	Reserved	Reserved	Reserved	Reserved	P3CFGA<3>	P3CFGA<2>	P3CFGA<1>	P3CFGA<0>
BFH	R/W	P3CFGB	Reserved	Reserved	Reserved	Reserved	P3CFGB<3>	P3CFGB<2>	P3CFGB<1>	P3CFGB<0>
СОН	R/W	TXT0	X24 POSN	DISPLAY X24	AUTO FRAME	DISABLE HEADER ROLL	DISPLAY STATUS ROW ONLY	DISABLE FRAME	VPS ON	INV ON

Table 2 SFR Map

TDA935X/6X/8X PS/N2 series

ADD	R/W	Names	BIT7	BIT6	BIT5	BIT4	BIT3	BIT2	BIT1	BIT0
С1Н	R/W	TXT1	EXT PKT OFF	8 BIT	ACQ OFF	X26 OFF	FULL FIELD	FIELD POLARITY	H POLARITY	V POLARITY
С2Н	R/W	TXT2	ACQ BANK	REQ<3>	REQ<2>	REQ<1>	REQ<0>	SC<2>	SC<1>	SC<0>
СЗН	W	TXT3	-	-	-	PRD<4>	PRD<3>	PRD<2>	PRD<1>	PRD<0>
С4Н	R/W	TXT4	OSD BANK ENABLE	QUAD WIDTH ENABLE	EAST/WES T	DISABLE DOUBLE HEIGHT	B MESH ENABLE	C MESH ENABLE	TRANS ENABLE	SHADOW ENABLE
С5Н	R/W	TXT5	BKGND OUT	BKGND IN	CORB OUT	CORB IN	TEXT OUT	TEXT IN	PICTURE ON OUT	PICTURE ON IN
С6Н	R/W	TXT6	BKGND OUT	BKGND IN	CORB OUT	CORB IN	TEXT OUT	TEXT IN	PICTURE ON OUT	PICTURE ON IN
С7Н	R/W	TXT7	STATUS ROW TOP	CURSOR ON	REVEAL	BOTTOM/ TOP	DOUBLE HEIGHT	BOX ON 24	BOX ON 1-23	BOX ON 0
C8H	R/W	TXT8	(Reserved)	FLICKER STOP ON	HUNT	DISABLE SPANISH	PKT 26 RECEIVED	WSS RECEIVED	WSS ON	CVBS1/ CVBS0
С9Н	R/W	TXT9	CURSOR FREEZE	CLEAR MEMORY	A0	R<4>	R<3>	R<2>	R<1>	R<0>
САН	R/W	TXT10	0	0	C<5>	C<4>	C<3>	C<2>	C<1>	C<0>
СВН	R/W	TXT11	D<7>	D<6>	D<5>	D<4>	D<3>	D<2>	D<1>	D<0>
ССН	R	TXT12	525/ <u>625</u> SYNC	ROM VER<4>	ROM VER<3>	ROM VER<2>	ROM VER<1>	ROM VER<0>	1	VIDEO SIGNAL QUALITY
CDH	R/W	TXT14	0	0	0	DISPLAY BANK	PAGE<3>	PAGE<2>	PAGE<1>	PAGE<0>
СЕН	R/W	TXT15	0	0	0	MICRO BANK	BLOCK<3>	BLOCK<2>	BLOCK<1>	BLOCK<0>
D0H	R/W	PSW	С	AC	F0	RS1	RS0	ov	-	P
D2H	R/W	TDACL	TD<7>	TD<6>	TD<5>	TD<4>	TD<3>	TD<2>	TD<1>	TD<0>
D3H	R/W	TDACH	TPWE	1	TD<13>	TD<12>	TD<11>	TD<10>	TD<9>	TD<8>
D5H	R/W	PWM0	PW0E	1	PW0V<5>	PW0V<4>	PW0V<3>	PW0V<2>	PW0V<1>	PW0V<0>
D6H	R/W	PWM1	PW1E	1	PW1V<5>	PW1V<4>	PW1V<3>	PW1V<2>	PW1V<1>	PW1V<0>
D7H	R	CCDAT1	CCD1<7>	CCD1<6>	CCD1<5>	CCD1<4>	CCD1<3>	CCD1<2>	CCD1<1>	CCD1<0>
D8H	R/W	S1CON	CR<2>	ENSI	STA	STO	SI	AA	CR<1>	CR<0>
D9H	R	S1STA	STAT<4>	STAT<3>	STAT<2>	STAT<1>	STAT<0>	0	0	0
DAH	R/W	S1DAT	DAT<7>	DAT<6>	DAT<5>	DAT<4>	DAT<3>	DAT<2>	DAT<1>	DAT<0>
DBH	R/W	S1ADR	ADR<6>	ADR<5>	ADR<4>	ADR<3>	ADR<2>	ADR<1>	ADR<0>	GC
DCH	R/W	PWM3	PW3E	1	PW3V<5>	PW3V<4>	PW3V<3>	PW3V<2>	PW3V<1>	PW3V<0>
ЕОН	R/W	ACC	ACC<7>	ACC<6>	ACC<5>	ACC<4>	ACC<3>	ACC<2>	ACC<1>	ACC<0>
Е4Н	R/W	PWM2	PW2E	1	PW2V<5>	PW2V<4>	PW2V<3>	PW2V<2>	PW2V<1>	PW2V<0>

Table 2 SFR Map

TDA935X/6X/8X PS/N2 series

ADD	R/W	Names	BIT7	BIT6	BIT5	BIT4	BIT3	BIT2	BIT1	BIT0
Е7Н	R	CCDAT2	CCD2<7>	CCD2<6>	CCD2<5>	CCD2<4>	CCD2<3>	CCD2<2>	CCD2<1>	CCD2<0>
E8H	R/W	SAD	VHI	CH<1>	CH<0>	ST	SAD<7>	SAD<6>	SAD<5>	SAD<4>
F0H	R/W	В	B<7>	B<6>	B<5>	B<4>	B<3>	B<2>	B<1>	B<0>
F8H	R/W	TXT13	VPS RECEIVED	PAGE CLEARING	525 DISPLAY	525 TEXT	625 TEXT	PKT 8/30	FASTEXT	0
FAH	R/W	XRAMP	XRAMP<7>	XRAMP<6>	XRAMP<5>	XRAMP<4>	XRAMP<3>	XRAMP<2>	XRAMP<1>	XRAMP<0>
FBH	R/W	ROMBK	STANDBY	IIC_LUT<1>	IIC_LUT<0>	0	0	0	ROMBK<1>	ROMBK<0>
FDH	R	TEST	TEST<7>	TEST<6>	TEST<5>	TEST<4>	TEST<3>	TEST<2>	TEST<1>	TEST<0>
FEH	W	WDTKEY	WKEY<7>	WKEY<6>	WKEY<5>	WKEY<4>	WKEY<3>	WKEY<2>	WKEY<1>	WKEY<0>
FFH	R/W	WDT	WDV<7>	WDV<6>	WDV<5>	WDV<4>	WDV<3>	WDV<2>	WDV<1>	WDV<0>

Table 2 SFR Map

A description of each of the SFR bits is shown in Table 3, The SFRs are in alphabetical order.

Names	ADD	BIT7	BIT6	BIT5	BIT4	BIT3	BIT2	BIT1	BIT0	RESET					
ACC	ЕОН	ACC<7>	ACC<6>	ACC<5>	ACC<4>	ACC<3>	ACC<2>	ACC<1>	ACC<0>	00H					
ACC<7:0>	Accumu	ılator value.													
В	F0H	B<7>	B<6>	B<5>	B<4>	B<3>	B<2>	B<1>	B<0>	00H					
B<7:0>	B Regis	ter value.													
CCDAT1	D7H	CCD1<7>	CCD1<6>	CCD1<5>	CCD1<4>	CCD1<3>	CCD1<2>	CCD1<1>	CCD1<0>	00H					
CCD1<7:0>	Closed	Caption first data	byte.												
CCDAT2	Е7Н	CCD2<7>	CCD2<7> CCD2<6> CCD2<5> CCD2<4> CCD2<3> CCD2<2> CCD2<1> CCD2<0> 00H												
CCD2<7:0>	Closed	Caption second of	lata byte.												
CCLIN	В7Н	0 0 0 CS<4> CS<3> CS<2> CS<1> CS<0> 15H													
CS<4:0>	Closed	Caption Slice lin	e using 525 line	number.											
DPH	83H	DPH<7>	DPH<6>	DPH<5>	DPH<4>	DPH<3>	DPH<2>	DPH<1>	DPH<0>	00H					
DPH<7:0>	Data Po	inter High byte,	used with DPL to	o address display	and auxiliary m	emory.									
DPL	82H	DPL<7>	DPL<6>	DPL<5>	DPL<4>	DPL<3>	DPL<2>	DPL<1>	DPL<0>	00H					
DPL<7:0>	Data po	inter low byte, u	sed with DPH to	address display	and auxiliary me	mory.									
IE	A8H	EA	EBUSY	ES2	ECC	ET1	EX1	ET0	EX0	00H					
EA	Disable	able all interrupts (0), or use individual interrupt enable bits (1).													
EBUSY	Enable l	Enable BUSY Interrupt.													
ES2	Enable l	² C Interrupt.													
ECC	Enable (Closed Caption I	nterrupt.												

Table 3 SFR Bit description

TDA935X/6X/8X PS/N2 series

Names	ADD	BIT7	BIT6	BIT5	BIT4	ВІТ3	BIT2	BIT1	BIT0	RESET				
ET1	Enable '	Timer 1 Interrupt	t.											
EX1	Enable 1	External Interrup	ot 1.											
ET0	Enable '	Timer 0 Interrup	t.											
EX0	Enable 1	External Interrup	ot 0.											
IEN1	84H	-	-	-	-	-	-	-	ET2	00Н				
ET2	Enable '	Timer 2 Interrup	t.											
IP	В8Н	0	PBUSY	PES2	PCC	PT1	PX1	PT0	PX0	00Н				
PBUSY	Priority	EBUSY Interrup	ot.		•									
PES2	Priority	ES2 Interrupt.												
PCC	Priority	ECC Interrupt.												
PT1	Priority	Timer 1 Interrup	ot.											
PX1	Priority	External Interru	pt 1.											
PT0	Priority	Timer 0 Interrup	ot.											
PX0	Priority	External Interru	pt 0.											
IP1	85H	-	-	-	-	-	-	-	PT2	00Н				
PT2	Priority	Timer 2 Interrup	ot.		•				<u> </u>	•				
P0	80H	Reserved	P0<6>	P0<5>	Reserved	Reserved	Reserved	Reserved	Reserved	FFH				
P0<6:5>	Port 0 I/	O register conne	ected to external p	pins.	•			1	•					
P1	90H	P1<7>	P1<6>	Reserved	Reserved	P1<3>	P1<2>	P1<1>	P1<0>	FFH				
P1<7:6>	Port 1 I/	O register conne	ected to external p	pins.	•				•	'				
P1<3:0>	Port 1 I/	O register conne	ected to external 1	oins.										
P2	АОН	Reserved	P2<6>	P2<5>	P2<4>	P2<3>	P2<2>	P2<1>	P2<0>	FFH				
P2<6:0>	Port 2 I/	O register conne	ected to external p	pins.						•				
Р3	ВОН	Reserved	Reserved	Reserved	Reserved	P3<3>	P3<2>	P3<1>	P3<0>	FFH				
P3<3:0>	Port 3 I/	O register conne	ected to external p	oins.										
POCFGA	96H	Reserved	P0CFGA<6>	P0CFGA<5>	Reserved	Reserved	Reserved	Reserved	Reserved	FFH				
P0CFGB	97H	Reserved	P0CFGB<6>	P0CFGB<5>	Reserved	Reserved	Reserved	Reserved	Reserved	00H				
P0CFG	B <x>/P0C</x>	CFGA < x > = 00	MODE 0 Oper	n Drain.			-	•		•				
P0CFG	B <x>/P0C</x>	CFGA < x > = 01	A <x> = 01 MODE 1 Quasi Bi-Directional.</x>											
P0CFG	B <x>/P0C</x>	CFGA < x > = 10	MODE2 High Impedance.											
P0CFG	B <x>/P0C</x>	CFGA <x> = 11</x>	MODE3 Push	Pull.										
P1CFGA	9EH	P1CFGA<7>	P1CFGA<6>	Reserved	Reserved	P1CFGA<3>	P1CFGA<2>	P1CFGA<1>	P1CFGA<0>	FFH				
P1CFGB	9FH	P1CFGB<7>	P1CFGB<6>	Reserved	Reserved	P1CFGB<3>	P1CFGB<2>	P1CFGB<1>	P1CFGB<0>	00H				

Table 3 SFR Bit description

TDA935X/6X/8X PS/N2 series

Names	ADD	BIT7	BIT6	BIT5	BIT4	BIT3	BIT2	BIT1	BIT0	RESET			
P1CFG:	B <x>/P1C</x>	CFGA < x > = 00	MODE 0 Oper	n Drain.									
P1CFG:	B <x>/P1C</x>	CFGA <x> = 01</x>	MODE 1 Qua	si Bi-Directional									
P1CFG:	B <x>/P1C</x>	CFGA <x> = 10</x>	MODE2 High	Impedance.									
P1CFG:	B <x>/P1C</x>	FGA <x> = 11</x>	MODE3 Push	Pull.									
P2CFGA	А6Н	Reserved	P2CFGA<6>	P2CFGA<5>	P2CFGA<4>	P2CFGA<3>	P2CFGA<2>	P2CFGA<1>	P2CFGA<0>	FFH			
P2CFGB	А7Н	Reserved	P2CFGB<6>	P2CFGB<5>	P2CFGB<4>	P2CFGB<3>	P2CFGB<2>	P2CFGB<1>	P2CFGB<0>	00Н			
P2CFG	B <x>/P2C</x>	CFGA < x > = 00	MODE 0 Oper	n Drain.									
P2CFG	B <x>/P2C</x>	CFGA < x > = 01	MODE 1 Qua	si Bi-Directional	•								
P2CFG	B <x>/P2C</x>	CFGA < x > = 10	MODE2 High	Impedance.									
P2CFG	B <x>/P2C</x>	CFGA <x> = 11</x>	MODE3 Push	Pull.									
P3CFGA	BEH	Reserved	Reserved										
P3CFGB	BFH	Reserved	Reserved	Reserved	Reserved	P3CFGB<3>	P3CFGB<2>	P3CFGB<1>	P3CFGB<0>	00Н			
P3CFG	B <x>/P3C</x>	$\langle x \rangle / P3CFGA \langle x \rangle = 00$ MODE 0 Open Drain.											
P3CFG	B <x>/P3C</x>	x > P3CFGA < x > = 01 MODE 1 Quasi Bi-directional.											
P3CFG	B <x>/P3C</x>	CFGA < x > = 10	MODE2 High	Impedance.									
P3CFG	B <x>/P3C</x>	CFGA <x> = 11</x>	MODE3 Push	Pull.									
PCON	87H	SMOD	ARD	RFI	WLE	GF1	GF0	PD	IDL	00H			
SMOD	UART I	Baud Rate Doubl	e Control.							•			
ARD	Auxilia	ry RAM Disable	, All MOVX inst	ructions access the	he external data r	nemory.							
RFI	Disable	ALE during inte	rnal access to re	duce Radio Frequ	uency Interference	ee.							
WLE	Watch I	Oog Timer enable	2.										
GF1	General	purpose flag.											
GF0	General	purpose flag.											
PD	Power-d	lown activation b	oit.										
IDL	Idle mo	de activation bit.											
PSW	D0H	С	AC	F0	RS<1>	RS<0>	OV	-	P	00Н			
С	Carry B	it.								•			
AC	Auxilia	ry Carry bit.											
F0	Flag 0, 0	General purpose	flag.										
RS<1:0>	RS<1:0: RS<1:0: RS<1:0: RS<1:0:	Bank selector b > = 00, Bank0 (0 > = 01, Bank1 (0 > = 10, Bank2 (1 > = 11, Bank3 (1	00H - 07H). 08H - 0FH). 0H - 17H).										
OV Table 2 SEE	Overflo	w flag.											

Table 3 SFR Bit description

TDA935X/6X/8X PS/N2 series

Names	ADD	BIT7	BIT6	BIT5	BIT4	BIT3	BIT2	BIT1	BIT0	RESET	
P	Parity bi	t.									
PWM0	D5H	PW0E	1	PW0V<5>	PW0V<4>	PW0V<3>	PW0V<2>	PW0V<1>	PW0V<0>	40H	
PW0E		- Disable Pulse Width Modulator 0 Enable Pulse Width Modulator 0.									
PW0V<5:0>	Pulse W	idth Modulator	high time.								
PWM1	D6H	PW1E 1 PW1V<5> PW1V<4> PW1V<3> PW1V<2> PW1V<1> PW1V<0>								40H	
PW1E		ble Pulse Width ble Pulse Width I									
PW1V<5:0>	Pulse W	idth Modulator	high time.								
PWM2	E4H	PW2E	1	PW2V<5>	PW2V<4>	PW2V<3>	PW2V<2>	PW2V<1>	PW2V<0>	40H	
PW2E		ble Pulse Width ble Pulse Width I			•				•		
PW2V<5:0>	Pulse W	idth Modulator	high time.								
PWM3	DCH	PW3E	1	PW3V<5>	PW3V<4>	PW3V<3>	PW3V<2>	PW3V<1>	PW3V<0>	40H	
PW3E		ble Pulse Width ble Pulse Width I			•					•	
PW3V<5:0>	Pulse W	idth Modulator	high time.								
ROMBK	FBH	STANDBY	IIC_LUT<1>	IIC_LUT<0>	0	0	0	ROMBK<1>	ROMBK<0>	00Н	
STANDBY		ble Stand-by Mo								•	
IIC_LUT<1:0 >	IIC_LU IIC_LU IIC_LU	kup table selecti T<1:0>=00, 558 T<1:0>=01, 558 T<1:0>=10, 558 T<1:0>=11, Res	Normal Mode. Fast Mode. Slow Mode.								
ROMBK<1:0	ROMBI ROMBI ROMBI	ank selection K<1:0>=00, Ban K<1:0>=01, Ban K<1:0>=10, Ban K<1:0>=11, Rese	k1 k2								
S1ADR	DBH	ADR<6>	ADR<5>	ADR<4>	ADR<3>	ADR<2>	ADR<1>	ADR<0>	GC	00H	
ADR<6:0>	I2C Sla	ve Address.	1	ı	ı	ı		I	ı	1	
GC		ble I ² C general c									
S1CON	D8H	CR<2>	ENSI	STA	STO	SI	AA	CR<1>	CR<0>	00Н	
CR<2:0>	Clock ra	ate bits. s are selectable (three tables)	1	ı	1	ı	ı	1	ı	
ENSI		ble I ² C interface ble I ² C interface.									

Table 3 SFR Bit description

TDA935X/6X/8X PS/N2 series

Names	ADD	BIT7	BIT6	BIT5	BIT4	BIT3	BIT2	BIT1	BIT0	RESET
STA			bit is set in slav					condition if the	bus is free or at	ter the bus
STO	be set ir	slave mode in	set in a master me order to recover and switches to t	from an error cor	ndition. In this ca	ase no STOP con	dition is generat	ed to the I2C bus		
SI	-A STA -The ow -The ge -A data -A data A STOR	RT condition is in slave address neral call address byte has been re byte has been re P or START con	is flag is set and a generated in mas has been receive as has been receive exceived or transm dition is received SCL remains LC	ter mode. d during AA=1. ved while S1ADF itted in master m itted as selected as selected slave	R.GC and AA=1. node (even if arbitalized) slave. e receiver or trans	tration is lost).				
AA	-Own sl -Genera -A data -A data	ave address is real call address is byte is received byte is received	g. When this bit eccived. received(S1ADR, while the device, while the device acknowledge is	R.GC=1). The is programmed to is selected slave.	to be a master re	ceiver.			call address is re	ceived.
S1DAT	DAH	DAT<7>	DAT<6>	DAT<5>	DAT<4>	DAT<3>	DAT<2>	DAT<1>	DAT<0>	00H
DAT<7:0>	I ² C Data	a.								
S1STA	D9H	STAT<4>	STAT<3>	STAT<2>	STAT<1>	STAT<0>	0	0	0	F8H
STAT<4:0>	I ² C Inte	rface Status.	•	•				•	•	
SAD	E8H	VHI	CH<1>	CH<0>	ST	SAD<7>	SAD<6>	SAD<5>	SAD<4>	00Н
VHI			age less than or e age greater then I	-	tage.					
CH<1:0>	CH<1:0 CH<1:0 CH<1:0	put channel sele > = 00,ADC3. > = 01,ADC0. > = 10,ADC1. > = 11,ADC2.	ect.							
ST			son between AD nd reset by Hard	-	and SADB<3:0>	value.				
SAD<7:4>	Most Si	gnificant nibble	of DAC input wo	ord						
SADB	98H	0	0	0	DC_COMP	SAD<3>	SAD<2>	SAD<1>	SAD<0>	00H
DC_COMP		ble DC Compar ble DC Compara		1	1	1	1	1	1	
SAD<3:0>	4-bit SA	D value.								
SP	81H	SP<7>	SP<6>	SP<5>	SP<4>	SP<3>	SP<2>	SP<1>	SP<0>	07H
SP<7>	Stack Po	ointer.	1	1	1	1	1	1		1
TCON	88H	TF1	TR1	TF0	TR0	IE1	IT1	IE0	IT0	00Н

Table 3 SFR Bit description

TDA935X/6X/8X PS/N2 series

Names	ADD	BIT7	BIT6	BIT5	BIT4	BIT3	BIT2	BIT1	BIT0	RESET		
TR1	Timer 1	Run control bit.	Set/Cleared by	software to turn	Γimer/Counter or	n/off.						
TF0	Timer 0	overflow Flag. S	Set by hardware	on Timer/Counte	r overflow.Cleare	ed by hardware v	hen processor v	ectors to interrup	ot routine.			
TR0	Timer 0	Run control bit.	Set/Cleared by	software to turn	Γimer/Counter or	n/off.						
IE1	Interrup	t 1 Edge flag (bo	oth edges generat	te flag). Set by ha	ardware when ex	ternal interrupt e	dge detected.Cle	ared by hardware	e when interrupt	processed.		
IT1	Interrup	t 1 Type control	bit. Set/Cleared	by Software to sp	pecify edge/low	evel triggered ex	ternal interrupts	•				
IE0	Interrup	t 0 Edge 1 flag. S	Set by hardware v	when external int	errupt edge detec	cted.Cleared by h	ardware when in	nterrupt processe	d.			
IT0	Interrup	t 0 Type flag.Set	/Cleared by Soft	ware to specify f	alling edge/low l	evel triggered ex	ternal interrupts.					
TDACH	D3H	TPWE	WE 1 TD<13> TD<12> TD<11> TD<10> TD<9> TD<8> 40									
TPWE		_	Width Modulat Width Modulate							•		
TD<13:8>	Tuning	Pulse Width Mo	dulator High Byt	te.								
TDACL	D2H	TD<7>	TD<6>	TD<5>	TD<4>	TD<3>	TD<2>	TD<1>	TD<0>	00Н		
TD<7:0>	Tuning	Pulse Width Mo	dulator Low Byte	e.	'	•	•	1	•			
ТН0	8CH	TH0<7>	TH0<6>	TH0<5>	TH0<4>	TH0<3>	TH0<2>	TH0<1>	TH0<0>	00Н		
TH0<7:0>	Timer 0	high byte.	1	•	•	•	•	1	1	•		
TH1	8DH	TH1<7>	TH1<6>	TH1<5>	TH1<4>	TH1<3>	TH1<2>	TH1<1>	TH1<0>	00Н		
TH1<7:0>	Timer 1	high byte.	!	!	!	!	!	·	!			
TL0	8AH	TL0<7>	TL0<6>	TL0<5>	TL0<4>	TL0<3>	TL0<2>	TL0<1>	TL0<0>	00Н		
TL0<7:0>	Timer 0	low byte.	1	•	•	•	•	1	1			
TL1	8BH	TL1<7>	TL1<6>	TL1<5>	TL1<4>	TL1<3>	TL1<2>	TL1<1>	TL1<0>	00Н		
TL1<7:0>	Timer 1	low byte.		'	'			1	•	!		
TMOD	89H	GATE	C/T	M1	M0	GATE	C/T	M1	M0	00Н		
						,				,		
			Timer / G	Counter 1			Timer /	Counter 0				
GATE	Gating (ting Control Timer /Counter 1.										
C/T	Counter	/Timer 1 selecto	r.									
M1,M0	M1,M0 M1,M0 M1,M0	= 01, 16 bit time	or 8 bit counter interval or even		2 pre-scaler. omatic reload up	on overflow. Rel	oad value stored	in TH1.				
GATE	Gating of	control Timer/Co	ounter 0.									
C/T	Counter	/Timer 0 selecto	r.									

Table 3 SFR Bit description

TDA935X/6X/8X PS/N2 series

Names	ADD	BIT7	BIT6	BIT5	BIT4	BIT3	BIT2	BIT1	BIT0	RESET
M1,M0	M1,M0 M1,M0 M1,M0		or 8 bit counter interval or even interval or event	t counter. counter with aut	2 pre-scaler. omatic reload up one 8 bit time in		oad value stored	in TH0.		
TP2CL	9CH	TP2CL<7>	TP2CL<6>	TP2CL<5>	TP2CL<4>	TP2CL<3>	TP2CL<2>	TP2CL<1>	TP2CL<0>	00Н
TP2CL<7:0>	Indicate	the low byte of	the Time 2 curre	nt value.	.!	1			!	<u>.l</u>
TP2CH	9DH	TP2CH<7>	TP2CH<6>	TP2CH<5>	TP2CH<4>	TP2CH<3>	TP2CH<2>	TP2CH<1>	TP2CH<0>	00Н
TP2CH<7:0>	Indicate	the high byte of	the Time 2 curre	ent value.	•	ı			•	.1
ТР2Н	92H	TP2H<7>	TP2H<6>	TP2H<5>	TP2H<4>	TP2H<3>	TP2H<2>	TP2H<1>	TP2H<0>	00Н
TP2H<7:0>	Timer 2	high byte, never	change unless u	pdated by the so	ftware.			!	!	1
TP2L	91H	TP2L<7>	TP2L<6>	TP2L<5>	TP2L<4>	TP2L<3>	TP2L<2>	TP2L<1>	TP2L<0>	00Н
TP2L<7:0>	Timer 2	low byte, never	change unless up	odated by the sof	tware.					
TP2PR	93H	TP2PR<7>	TP2PR<6>	TP2PR<5>	TP2PR<4>	TP2PR<3>	TP2PR<2>	TP2PR<1>	TP2PR<0>	00Н
TP2H<7:0>	Timer 2	Pre-scaler, neve	r change unless t	pdated by the so	oftware.					1
TP2CRL	94H	-	-	-	-	-	-	TP2CRL<1>	TP2CRL<0>	00Н
TP2CRL<0>	0 - Time	Control. er 2 disabled. er 2 enabled.						l		
TP2CRL<1>	Timer 2 0 - No 0 1 - Over	Overflow.								
TEST	FDH	TEST<7>	TEST<6>	TEST<5>	TEST<4>	TEST<3>	TEST<2>	TEST<1>	TEST<0>	00Н
TEST<2:0>	011 - D 001 - A	n Type bit SEL isplay Dram test. cquisition1 test. cquisition2 test			,				,	
TEST<4:3>	Functio	nal test mode bit	s, set via mode s	elect logic.						
TEST<7:5>	001 - 2H 010 - 6H 011 - 7H 100 - 12 101 - 14 110 - 1H	5K x 16. K x 16. K x 16. K x 16. K x 16. K x 16.								
TXT0	СОН	X24 POSN	DISPLAY X24	AUTO FRAME	DISABLE HEADER ROLL	DISPLAY STATUS ROW ONLY	DISABLE FRAME	VPS ON	INV ON	00Н
X24 POSN		e X/24 in extensi e X/24 in basic p	-	h packets 0 to 23					•	

Table 3 SFR Bit description

TDA935X/6X/8X PS/N2 series

Names	ADD	BIT7	BIT6	BIT5	BIT4	BIT3	BIT2	BIT1	BIT0	RESET
DISLAY X24	-	=	basic page memo appropriate local	=	memory					
AUTO FRAME		nal Frame output e output is switc	t hed off automati	cally if any vide	o displayed					
DISABLE HEADER ROLL		_	and time to curr							
DISPLAY STATUS ROW ONLY	-	lay normal page ay only row 24	rows 0 to 24							
DISABLE FRAME		nal Frame output Frame output to								
VPS ON		acquisition off acquisition on								
INV ON		ntory page off ntory page on								
TXT1	С1Н	EXT PKT OFF	8 BIT	ACQ OFF	ACQ OFF	FULL FIELD	FIELD POLARITY	H POLARITY	V POLARITY	00Н
EXT PKT OFF	-	-	ckets X/24,X/27, f extension packe							
8 BIT			orrect packets 0 to 24 w		ory					
ACQ OFF		-	into display men ta into Display n	-						
X26 OFF		-	cessing of X/26							
FULL FIELD	-	-	on selected line							
FIELD POLARIY	-	-	alf of line during d half of line dur							
H POLARITY	-	_	e is positive going	=						
V POLARITY		U	e is positive going	_						
TXT2	С2Н	ACQ BANK	REQ<3>	REQ<2>	REQ<1>	REQ<0>	SC<2>	SC<1>	SC<0>	00Н
ACQ_BANK		et Acquisition ba								
REQ<3:0>	Page req	uest								
SC<2:0>	Start col	umn of page req	uest							
TXT3	СЗН				PRD<4>	PRD<3>	PRD<2>	PRD<1>	PRD<0>	00Н
PRD<4:0>	Page Re	quest data		•	•	•	•	•		

Table 3 SFR Bit description

TDA935X/6X/8X PS/N2 series

Names	ADD	BIT7	BIT6	BIT5	BIT4	BIT3	BIT2	BIT1	BIT0	RESET
TXT4	С4Н	OSD BANK ENABLE	QUAD WIDTH ENABLE	EAST/WEST	DISABLE DBL HEIGHT	B MESH ENABLE	C MESH ENABLE	TRANS ENABLE	SHADOW ENABLE	00Н
OSD BANK ENABLE		•	OSD characters a			cation				
QUAD WIDTH ENABLE			uadruple width c							
EAST/WEST			lection of charact							
DISABLE DOUBLE HEIGHT			ing of double hei	_						
B MESH ENABLE		mal display of blole meshing of b	ack background lack background							
C MESH ENABLE			loured backgrour oloured backgrou							
TRANS ENABLE	_	olay black backg olay black backg								
SHADOW ENABLE		able display of sh olay shadow/ frin	adow/fringing ge (default SE bl	ack)						
TXT5	С5Н	BKGND OUT	BKGND IN	COR OUT	COR IN	TEXT OUT	TEXT IN	PICTURE ON OUT	PICTURE ON IN	03Н
BKGND OUT		_	ot displayed outs lisplayed outside		es					
BKGND IN		_	ot displayed insi lisplayed inside t							
COR OUT			de teletext and C eletext and OSD							
COR IN			e teletext and OS letext and OSD b							
TEXT OUT			outside teletext boxe							
TEXT IN			inside teletext bo de teletext boxes							
PICTURE ON OUT			d outside teletext itside teletext box							
PICTURE ON IN			d inside teletext boxe							
TXT6	С6Н	BKGND OUT	BKGND IN	COR OUT	COR IN	TEXT OUT	TEXT IN	PICTURE ON OUT	PICTURE ON IN	03Н
BKGND OUT		_	ot displayed outs displayed outside		es					

Table 3 SFR Bit description

TDA935X/6X/8X PS/N2 series

Names	ADD	BIT7	BIT6	BIT5	BIT4	віт3	BIT2	BIT1	BIT0	RESET
BKGND IN		0 - Background colour not displayed inside teletext boxes 1 - Background colour displayed inside teletext boxes								
COR OUT			de teletext and O eletext and OSD							
COR IN			e teletext and OS etext and OSD b							
TEXT OUT			outside teletext b							
TEXT IN			inside teletext bo	oxes						
PICTURE ON OUT			l outside teletext tside teletext box							
PICTURE ON IN			l inside teletext b							
TXT7	С7Н	STATUS ROW TOP	CURSOR ON	REVEAL	BOTTOM/ TOP	DOUBLE HEIGHT	BOX ON 24	BOX ON 1-23	BOX ON 0	00Н
STATUS ROW TOP	•			•	ge (on display ro ge (on display ro	*				
CURSOR ON		ble display of cu lay cursor at pos	rsor ition given by TX	XT9 and TXT10						
REVEAL	-	-	area with concea		oute set					
BOTTOM/TO P	_		s 0 to 11 when do	-						
DOUBLE HEIGHT	-	=	ers with normal h	_						
BOX ON 24			etext boxes in me	-						
BOX ON 1-23			etext boxes in me	=						
BOX ON 0			etext boxes in me	•						
TXT8	С8Н	(Reserved)	FLICKER STOP ON	HUNT	DISABLE SPANISH	PKT 26 RECEIVED	WSS RECEIVED	WSS ON	(Reserved)	00Н
FLICKER STOP ON		ole 'Flicker Stopp ble 'Flicker Stop	-		ı					
HUNT			ting for amplitud		cquired					
DISABLE SPANISH	0 - Enable special treatment of Spanish packet 26 characters 1 - Disable special treatment of Spanish packet 26 characters									

Table 3 SFR Bit description

TDA935X/6X/8X PS/N2 series

Names	ADD	BIT7	BIT6	BIT5	BIT4	BIT3	BIT2	BIT1	BIT0	RESET
PKT 26 RECEIVED	1 - Pack	packet 26 data ha tet 26 data has be his flag is set by	en processed.		oftware					
WSS RECEIVED	1 - Wide	Wide Screen Sign e Screen signallin his flag is set by	ng data has been	processed	oftware.					
WSS ON		ble acquisition of								
TXT9	С9Н	CURSOR FREEZE	CLEAR MEMORY	A0	R<4>	R<3>	R<2>	R<1>	R<0>	00Н
CURSOR FREEZE		current TXT9 an		for cursor positi	on.					
CLEAR MEMORY		r memory block his flag is set by	-							
A0		ess memory block	-	XT15						
R<4:0>		memory ROW valid range TXT r		mode 0 to 15						
TXT10	CAH	0	0	C<5>	C<4>	C<3>	C<2>	C<1>	C<0>	00Н
C<5:0>		memory COLUI		2 mode 0 to 47						
TXT11	СВН	D<7>	D<6>	D<5>	D<4>	D<3>	D<2>	D<1>	D<0>	00Н
D<7:0>	Data va	lue written or rea	d from memory	location defined	by TXT9, TXT1	0 and TXT15				
TXT12	ССН	625/525 SYNC	ROM VER<4>	ROM VER<3>	ROM VER<2>	ROM VER<1>	ROM VER<0>	1	VIDEO SIGNAL QUALITY	xxxxx x1xB
625/525 SYNC		line CVBS signa	-							
ROM VER<4:0>	Rom Ve	rogrammable ide ersion <4> : nish Flicker Stopp nish Flicker Stopp	oer Disabled.		3 Bit-6).					
1	Reserve	d								
VIDEO SIGNAL QUALITY	-	uisition can not b	-	-						
TXT13	F8H	VPS RECEIVED	PAGE CLEARING	525 DISPLAY	525 TEXT	625 TEXT	PKT 8/30	FASTEXT	0	xxxxx xx0B
VPS RECEIVED	0 - 1 - VPS	data								
PAGE CLEARING	_	page clearing acti		rogress						

Table 3 SFR Bit description

TDA935X/6X/8X PS/N2 series

Names	ADD	BIT7	BIT6	BIT5	BIT4	BIT3	BIT2	BIT1	BIT0	RESET
525 DISPLAY		=	ation for Display ation for Display							
525 TEXT		Line WST not b	_							
625 TEXT		Line WST not b	_							
PKT 8/30		,	5) or Packet 4/30 r Packet 4/30/x(, ,						
FASTEXT		Packet x/27 data xet x/27 data dete								
0	Reserve	ed								
TXT14	CDH	0	0	0	DISPLAY BANK	PAGE<3>	PAGE<2>	PAGE<1>	PAGE<0>	00Н
DISPLAY BANK		ct lower bank for ct upper bank fo								
PAGE<3:0>	Current	Display page								
TXT15	СЕН	0	0	0	MICRO BANK	BLOCK<3	BLOCK<2	BLOCK<1	BLOCK<0	00Н
MICRO BANK		ct lower bank fo								
BLOCK<3:0>	Current	Micro block to	be accessed by T	XT9, TXT10 at	nd TXT11					
TXT17	В9Н	0	FORCE ACQ<1>	FORCE ACQ<0>	FORCE DISP<1>	FORCE DISP<0>	SCREEN COL2	SCREEN COL1	SCREEN COL0	00Н
FORCE ACQ<1:0>	01 - For 10 - For	rce 625 timing, F	n Force 525 Teletex Force 625 Teletex Force 525 Teletex	t Standard						
FORCE DISP<1:0>	01 - For 10 - For		25 mode (9 lines 25 mode (10 lines	-						
SCREEN COL<2:0>	Defines colour to be displayed instead of TV picture and black background. The bits <2:0> are equivalent to the RGB components 000 - Transparent 001 - CLUT entry 9 010 - CLUT entry 10 011 - CLUT entry 11 100 - CLUT entry 12 101 - CLUT entry 13 110- CLUT entry 14 111 - CLUT entry 15									
TXT18	В2Н	NOT<3>	NOT<2>	NOT<1>	NOT<0>	0	0	BS<1>	BS<0>	00H
NOT<3:0>	Nationa	l Option table se	election, maximu	m of 32 when u	sed with East/Wes	st bit				
BS<1:0>	Basic C	haracter set sele	ction							
	•									

 Table 3
 SFR Bit description

TDA935X/6X/8X PS/N2 series

Names	ADD	BIT7	BIT6	BIT5	BIT4	BIT3	BIT2	BIT1	BIT0	RESET
TXT19	ВЗН	TEN	TC<2>	TC<1>	TC<0>	0	0	TS<1>	TS<0>	00H
TEN		ble Twist function								
TC<2:0>	Langua	ge control bits (C	C12/C13/C14) tha	nt has Twisted ch	aracter set					
TS<1:0>	Twist C	haracter set selec	etion							
TXT20	В4Н	DRCS ENABLE	OSD PLANES	0	0	OSD LANG ENABLE	OSD LAN<2>	OSD LAN<1>	OSD LAN<0>	00Н
DRCS ENABLE		mal OSD charact nap column 9 to	ers used DRCS (TXT and	I CC modes),						
OSD PLANES			nns 8 and 9 defin ns 8 and 9 define							
OSD LANG ENABLE	Enable	use of OSD LAN	I<2:0> to define	language option	for display, inste	ad of C12/C13/C	C14			
OSD LAN<2:0>	Alternat	tive C12/C13/C1	4 bits for use wit	h OSD menus						_
TXT21	В5Н	DISP LINES<1>	DISP LINES<0>	CHAR SIZE<1>	CHAR SIZE<0>	Reserved	CC ON	I2C PORT0	CC/TXT	02Н
DISP LINES<1:0>	00 - 10 01 - 13	lines per characte lines per characte lines per characte			e)					
CHAR SIZE<1:0>	00 - 10 01 - 13	lines per characte	er (matrix 12x10 er (matrix 12x13 er (matrix 12x16))						
CCON		ed Caption acquied Caption acqui								
I2C PORT0		ble I2C PORT0 ble I2C PORT0 s	election (P1.7/SI	DA0, P1.6/SCL0)					
CC/TXT	0 - Display configured for TXT mode 1 - Display configured for CC mode									
TXT22	B6H GPF<7> GPF<6> GPF<5> GPF<4> GPF<3> GPF<2> GPF<1> GPF<0> X								ХХН	
GPF<7:6>	General purpose register, bits defined by mask programmable bits									
GPF<5>	0 - Standard Painter device 1 - Enhanced Painter device									
GPF<4> (Used for software only)		ose 6 page telete: ose 10 page telet								
GPF<3>	0 - PWM0, PWM1, PWM2 & PWM3 output on Port 3.0 to Port 3.3 respectively 1 - PWM0, PWM1, PWM2 & PWM3 output on Port 2.1 to Port 2.4 respectively									

Table 3 SFR Bit description

TDA935X/6X/8X PS/N2 series

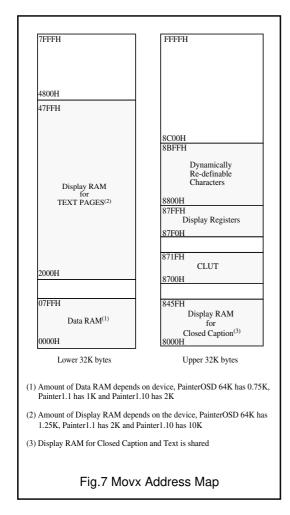
Names	ADD	BIT7	BIT6	BIT5	BIT4	BIT3	BIT2	BIT1	BIT0	RESET
GPF<2>		ble Closed Capti ble Closed Caption								
GPF<1>		ble Text acquisit								
GPF<0> (Polarity reversed in Painter1_Plus standalone)	0 - Stan 1 - UOC	dalone (Painter1	_Plus) mode							
WDT	FFH	WDV<7>	WDV<6>	WDV<5>	WDV<4>	WDV<3>	WDV<2>	WDV<1>	WDV<0>	00Н
WDv<7:0>	Watch I	Oog Timer period	i			•		•		1
WDTKEY	FEH	WKEY<7>	WKEY<6>	WKEY<5>	WKEY<4>	WKEY<3>	WKEY<2>	WKEY<1>	WKEY<0>	00Н
WKEY<7:0>		Dog Timer Key. Iust be set to 55H	H to disable Watc	h dog timer whe	en active.					
WSS1	ВАН	0	0	0	WSS<3:0> ERROR	WSS<3>	WSS<2>	WSS<1>	WSS<0>	00Н
WSS<3:0> ERROR		error in WSS<3:0 r in WSS<3:0>)>		1					
WSS<3:0>	Signalli	ng bits to define	aspect ratio (grou	up 1)						
WSS2	ВВН	0	0	0	WSS<7:4> ERROR	WSS<7>	WSS<6>	WSS<5>	WSS<4>	00Н
WSS<7:4> ERROR		errors in WSS<7: r in WSS<7:4>	4>							
WSS<7:4>	Signalli	ng bits to define	enhanced service	es (group 2)						
WSS3	ВСН	WSS<13:11 < ERROR	WSS<13>	WSS<12>	WSS<11>	WSS<10:8> ERROR	WSS<10>	WSS<9>	WSS<8>	00Н
WSS<13:11> ERROR		error in WSS<13: r in WSS<13:11								
WSS<13:11>	Signalli	ng bits to define	reserved element	is (group 4)						
WSS<10:8> ERROR		error in WSS<10: r in WS<10:8>	:8>							
WSS<10:8>	Signalli	ng bits to define	subtitles (group 3	3)						
XRAMP	FAH	XRAMP<7>	XRAMP<6>	XRAMP<5>	XRAMP<4>	XRAMP<3>	XRAMP<2>	XRAMP<1>	XRAMP<0>	00Н
					-	•	-	•		

Table 3 SFR Bit description

TDA935X/6X/8X PS/N2 series

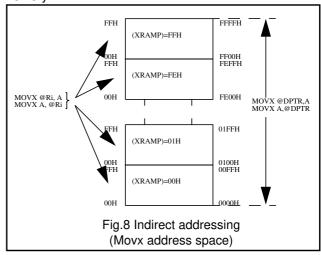
External (Auxiliary + Display) Memory

The normal 80C51 external memory area has been mapped internally to the device, this means that the MOVX instruction accesses data memory internal to the device. The movx memory map is shown in Fig.7.



Auxiliary RAM Page Selection

The Auxiliary RAM page pointer is used to select one of the 256 pages within the auxiliary RAM, not all pages are allocated, refer to Fig.8. A page consists of 256 consecutive bytes. XRAMP only works on internal MOVX memory.



Power-on Reset

Power on reset is generated internally to the TDA935X/6x/8x device, hence no external reset circuitry is required. The TV processor die shall generate the master reset in the system, which in turn will reset the microcontroller die

A external reset pin is still present and is logically ORed with the internal Power on reset. This pin will only be used for test modes and OTP/ISP programming. The active high reset pin incorporates an internal pull-down, thus it can be left unconnected in application.

Power Saving modes of Operation

There are three Power Saving modes, Idle, Stand-by and Power Down, incorporated into the Painter1_Plus die. When utilizing either mode, the 3.3v power to the device (Vddp, Vddc & Vdda) should be maintained, since Power Saving is achieved by clock gating on a section by section basis.

STAND-BY MODE

During Stand-by mode, the Acquisition and Display sections of the device are disabled. The following functions remain active:-

- 80c51 CPU Core
- · Memory Interface
- 12C
- Timer/Counters
- WatchDog Timer
- SAD and PWMs

TDA935X/6X/8X PS/N2 series

To enter Stand-by mode, the STAND-BY bit in the ROMBANK register must be set. Once in Stand-By, the XTAL oscillator continues to run, but the internal clock to Acquisition and Display are gated out. However, the clocks to the 80c51 CPU Core, Memory Interface, I2C, Timer/Counters, WatchDog Timer and Pulse Width Modulators are maintained. Since the output values on RGB and VDS are maintained the display output must be disabled before entering this mode.

This mode may be used in conjunction with both Idle and Power-Down modes. Hence, prior to entering either Idle or Power-Down, the STAND-BY bit may be set, thus allowing wake-up of the 80c51 CPU core without fully waking the entire device (This enables detection of a Remote Control source in a power saving mode).

IDLE MODE

During Idle mode, Acquisition, Display and the CPU sections of the device are disabled. The following functions remain active:-

- · Memory Interface
- I2C
- Timer/Counters
- WatchDog Timer
- SAD & PWMs

To enter Idle mode the IDL bit in the PCON register must be set. The WatchDog timer must be disabled prior to entering Idle to prevent the device being reset. Once in Idle mode, the XTAL oscillator continues to run, but the internal clock to the CPU, Acquisition and Display are gated out. However, the clocks to the Memory Interface, I2C, Timer/Counters, WatchDog Timer and Pulse Width Modulators are maintained. The CPU state is frozen along with the status of all SFRs, internal RAM contents are maintained, as are the device output pin values. Since the output values on RGB and VDS are maintained the Display output must be disabled before entering this mode.

There are three methods available to recover from Idle:-

- Assertion of an enabled interrupt will cause the IDL bit to be cleared by hardware, thus terminating Idle mode.
 The interrupt is serviced, and following the instruction RETI, the next instruction to be executed will be the one after the instruction that put the device into Idle mode.
- A second method of exiting Idle is via an Interrupt generated by the SAD DC Compare circuit. When Painter is configured in this mode, detection of an analogue threshold at the input to the SAD may be used to trigger wake-up of the device i.e. TV Front Panel Key-press. As above, the interrupt is serviced, and following the instruction RETI, the next instruction to be

- executed will be the one following the instruction that put the device into Idle.
- The third method of terminating Idle mode is with an external hardware reset. Since the oscillator is running, the hardware reset need only be active for two machine cycles (24 clocks at 12MHz) to complete the reset operation. Reset defines all SFRs and Display memory to a pre-defined state, but maintains all other RAM values. Code execution commences with the Program Counter set to '0000'.

Power Down Mode

In Power Down mode the XTAL oscillator still runs, and differential clock transmitter is active. The contents of all SFRs and Data memory are maintained, however, the contents of the Auxiliary/Display memory are lost. The port pins maintain the values defined by their associated SFRs. Since the output values on RGB and VDS are maintained the Display output must be made inactive before entering Power Down mode.

The power down mode is activated by setting the PD bit in the PCON register. It is advised to disable the WatchDog timer prior to entering Power down. Recovery from Power-Down takes several milli-seconds as the oscillator must be given time to stabilise.

There are three methods of exiting power down:-

- An External interrupt provides the first mechanism for waking from Power-Down. Since the clock is stopped, external interrupts needs to be set level sensitive prior to entering Power-Down. The interrupt is serviced, and following the instruction RETI, the next instruction to be executed will be the one after the instruction that put the device into Power-Down mode.
- A second method of exiting Power-Down is via an
 Interrupt generated by the SAD DC Compare circuit.
 When Painter is configured in this mode, detection of a
 certain analogue threshold at the input to the SAD may
 be used to trigger wake-up of the device i.e. TV Front
 Panel Key-press. As above, the interrupt is serviced,
 and following the instruction RETI, the next instruction to
 be executed will be the one following the instruction that
 put the device into Power-Down.
- The third method of terminating the Power-Down mode is with an external hardware reset. Reset defines all SFRs and Display memory, but maintains all other RAM values. Code execution commences with the Program Counter set to '0000'.

TDA935X/6X/8X PS/N2 series

I/O Facility

I/O PORTS

The IC has 13 I/O lines, each is individually addressable, or form part of 4 parallel addressable ports which are port0, port1, port2 and port3.

PORT TYPE

All individual ports can be programmed to function in one of four modes, the mode is defined by two Port Configuration SFRs. The modes available are Open Drain, Quasi-bidirectional, High Impedance and Push-Pull.

Open Drain

The Open drain mode can be used for bi-directional operation of a port. It requires an external pull-up resistor, the pull-up voltage has a maximum value of 5.5V, to allow connection of the device into a 5V environment.

Quasi bi-directional

The quasi-bidirectional mode is a combination of open drain and push pull. It requires an external pull-up resistor to VDDp (nominally 3.3V). When a signal transition from 0->1 is output from the device, the pad is put into push-pull mode for one clock cycle (166ns) after which the pad goes into open drain mode. This mode is used to speed up the edges of signal transitions. This is the default mode of operation of the pads after reset.

High Impedance

The high impedance mode can be used for Input only operation of the port. When using this configuration the two output transistors are turned off.

Push-Pull

The push pull mode can be used for output only. In this mode the signal is driven to either 0V or VDDp, which is nominally 3.3V.

Interrupt System

The device has 8 interrupt sources, each of which can be enabled or disabled. When enabled, each interrupt can be assigned one of two priority levels. There are four interrupts that are common to the 80C51, two of these are external interrupts (EX0 and EX1) and the other two are timer interrupts (ET0 and ET1). There is also one interrupt (ES2) connected to the 80c51 micro-controller IIC peripheral for Transmit and Receive operation.

The TDA935X/6x/8x family of devices have an additional 16-bit Timer (with 8-bit Pre-scaler). To accommodate this, another interrupt ET2PR has been added to indicate timer overflow.

In addition to the conventional 80c51, two application specific interrupts are incorporated internal to the device which have the following functionality:-

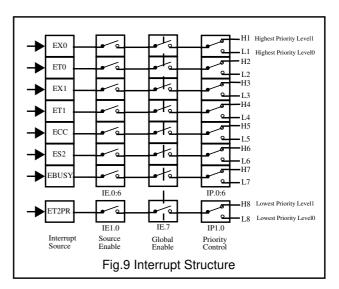
CC (Closed Caption Data Ready Interrupt) - This interrupt is generated when the device is configured for Closed Caption acquisition. The interrupt is activated at the end of the currently selected Slice Line as defined in the CCLIN SFR.

BUSY (Display Busy Interrupt) - An interrupt is generated when the Display enters either a Horizontal or Vertical Blanking Period. i.e. Indicates when the micro-controller can update the Display RAM without causing undesired effects on the screen. This interrupt can be configured in one of two modes using the MMR Configuration Register (Address 87FF, Bit-3 [TXT/V]):-

- TeXT Display Busy: An interrupt is generated on each active horizontal display line when the Horizontal Blanking Period is entered.
- Vertical Display Busy: An interrupt is generated on each vertical display field when the Vertical Blanking Period is entered.

INTERRUPT ENABLE STRUCTURE

Each of the individual interrupts can be enabled or disabled by setting or clearing the relevant bit in the interrupt enable SFRs (IE and IEN1). All interrupt sources can also be globally disabled by clearing the EA bit (IE.7).



INTERRUPT ENABLE PRIORITY

Each interrupt source can be assigned one of two priority levels. The interrupt priorities are defined by the interrupt priority SFRs (IP and IP1). A low priority interrupt can be interrupted by a high priority interrupt, but not by another low priority interrupt. A high priority interrupt can not be interrupted by any other interrupt source. If two requests of

TDA935X/6X/8X PS/N2 series

different priority level are received simultaneously, the request with the highest priority level is serviced. If requests of the same priority level are received simultaneously, an internal polling sequence determines which request is serviced. Thus, within each priority level there is a second priority structure determined by the polling sequence as defined in Table 4.

Source	Priority within level	Interrupt Vector
EX0	Highest	0003H
ET0		000BH
EX1		0013H
ET1		001BH
ECC		0023H
ES2		002BH
EBUSY		0033H
ET2PR	Lowest	003BH

Table 4 Interrupt Priority (within same level)

INTERRUPT VECTOR ADDRESS

The processor acknowledges an interrupt request by executing a hardware generated LCALL to the appropriate servicing routine. The interrupt vector addresses are shown in Table 4.

LEVEL/EDGE INTERRUPT

The external interrupt can be programmed to be either level-activated or transition activated by setting or clearing the IT0/1 bits in the Timer Control SFR(TCON).

ITx	Level	Edge	
0	Active low		
1		INT0 = Negative Edge INT1 = Positive and Negative Edge	

Table 5 External Interrupt Activation

The external interrupt INT1 differs from the standard 80C51 in that it is activated on both edges when in edge sensitive mode. This is to allow software pulse width measurement for handling remote control inputs.

Timer/Counter

Two 16 bit timers/counters are incorporated Timer0 and Timer1. Both can be configured to operate as either timers or event counters.

In Timer mode, the register is incremented on every machine cycle. It is therefore counting machine cycles. Since the machine cycle consists of 12 oscillator periods, the count rate is 1/12 Fosc = 1MHz.

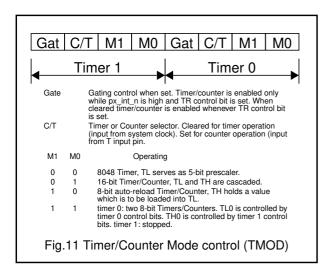
In Counter mode, the register is incremented in response to a negative transition at its corresponding external pin T0/1. Since the pins T0/1 are sampled once per machine cycle it takes two machine cycles to recognise a transition, this gives a maximum count rate of 1/24 Fosc = 0.5MHz. There are six special function registers used to control the timers/counters as defined in Table 6.

SFR	Address
TCON	88H
TMOD	89Н
TL0	8AH
TH0	8BH
TL1	8CH
TH1	8DH

Table 6 Timer/Counter Registers

TF1 TR	TF0	TR	IE1	IT1	IE0	IT0
Symbol TF1	Position TCON.7		Name and Significance Timer 1 overflow flag. Set by hard- ware on timer/counter overflow. Cleared by hardware when processor vectors to interrupt routine.			
TR1	TCON.6		Timer 1 F	Run contro	ol bit. Set/ timer.cou	
TF0	TCON.5		Timer 0 overflow flag. Set by hard- ware on timer/counter overflow. Cleared by hardware when processor vectors to interrupt routine.		ow.	
TR0	TCON.4		Timer 0 Run control bit. Set/cleared by software to turn timer.counter on/off.			
Symbol IE1	Position TCON.3		Name and Significance Interrupt 1 Edge flag. Set by hardware when external interrupt edge detected. Cleared when interrupt processed.		9	
IT1	TCON.2		Interrupt 1 Type control bit. Set/cleared by software to specify falling edge/low level triggered external interrupts.			
IE0	TCON.1		Interrupt 0 Edge flag. Set by hardware when external interrupt edge detected. Cleared when interrupt processed.		9	
IT0	TCON.0		Interrupt 0 Type control bit. Set/cleared by software to specify falling edge/low level triggered external interrupts.			
Fig.10 Timer/Counter Control (TCON) register						

TDA935X/6X/8X PS/N2 series



The Timer/Counter function is selected by control bits C/T in the Timer Mode SFR (TMOD). These two Timer/Counter have four operating modes, which are selected by bit-pairs (M1.M0) in the TMOD. Refer to the 80C51 based 8-bit micro-controllers - Philips Semiconductors (ref. IC20) for detail of the modes and operation.

TL0/TL1 and TH0/TH1 are the actual timer/counter registers for timer0 / timer1. TL0/TL1 is the low byte and TH0/TH1 is the high byte.

TIMER WITH PRE-SCALER

An additional 16-bit timer with 8-bit pre-scaler is provided to allow timer periods up to 16.777 seconds. This timer remains active during IDLE mode.

TP2L sets the lower value of the period for timer 2 and TP2H is the upper timer value. TP2PR provides an 8-bit pre-scaler for timer 2. The value on TP2PR, TP2H and TP2L shall never change unless updated by the software. If the micro reads TP2R, TP2H orTP2L at any stage, this should return the value written and not the current timer 2 value. The timer 2 should continue after overflow by re-loading the timer with the values of SFRs TP2PR, TP2H and TP2L.

TP2CL and TP2CH indicate the current timer 2 value. These should be readable both when the timer 2 is active and inactive. Once the timer 2 is disable, the timer 2 value at the time of disabling should be maintained on the SFRs TP2CL and TP2CH. At a count of zero (on TP2CL and TP2CH), the overflow flag should be set:- TP2CRL<1>- '0' = no timer 2 overflow, '1'= timer 2 overflow.

TP2CRL is the control and status for timer 2. TP2CRL.0 is the timer enable and TP2CRL.1 is the timer overflow status. The overflow flag will need to be reset by software. Hence, if required, software may poll flag rather than use interrupt. Upon overflow an interrupt should also be generated.

Reset values of all registers should be 00 hex. In Timer mode, Timer 2 should count down from the value set on SFRs TP2PR, TP2H and TP2L. It is therefore counting machine cycles. Since the machine cycle consists of 12 oscillator periods, the count rate is 1/12 fosc (1MHz).

Timer2 interval = (TP2H * 256 + TP2L) * (TP2PR + 1) * 1 us

WatchDog Timer

The WatchDog timer is a counter that once in an overflow state forces the micro-controller in to a reset condition. The purpose of the WatchDog timer is to reset the micro-controller if it enters an erroneous processor state (possibly caused by electrical noise or RFI) within a reasonable period of time. When enabled, the WatchDog circuitry will generate a system reset if the user program fails to reload the WatchDog timer within a specified length of time known as the WatchDog interval.

The WatchDog timer consists of an 8-bit counter with an 16-bit pre-scaler. The pre-scaler is fed with a signal whose frequency is 1/12 fosc (1MHz).

The 8 bit timer is incremented every 't' seconds where:

 $t=12x65536x1/fosc=12x65536x1/12x10^6=65.536ms$

WATCHDOG TIMER OPERATION

The WatchDog operation is activated when the WLE bit in the Power Control SFR (PCON) is set. The WatchDog can be disabled by Software by loading the value 55H into the WatchDog Key SFR (WDTKEY). This must be performed before entering Idle/Power Down mode to prevent exiting the mode prematurely.

Once activated the WatchDog timer SFR (WDT) must be reloaded before the timer overflows. The WLE bit must be set to enable loading of the WDT SFR, once loaded the WLE bit is reset by hardware, this is to prevent erroneous Software from loading the WDT SFR.

The value loaded into the WDT defines the WatchDog interval.

WatchDog interval = (256 - WDT) * t = (256 - WDT) * 65.536ms.

The range of intervals is from WDT=00H which gives 16.777s to WDT=FFH which gives 65.536ms.

PORT Alternate Functions

The Ports 1,2 and 3 are shared with alternate functions to enable control of external devices and circuitry. The alternate functions are enabled by setting the appropriate

TDA935X/6X/8X PS/N2 series

SFR and also writing a '1' to the Port bit that the function occupies.

PWM Pulse Width Modulators

The device has four 6-bit Pulse Width Modulated (PWM) outputs for analogue control of e.g. volume, balance, bass and treble. The PWM outputs generate pulse patterns with a repetition rate of 21.33us, with the high time equal to the PWM SFR value multiplied by 0.33us. The analogue value is determined by the ratio of the high time to the repetition time, a D.C. voltage proportional to the PWM setting is obtained by means of an external integration network (low pass filter).

PWM Control

The relevant PWM is enabled by setting the PWM enable bit PWxE in the PWMx Control register. The high time is defined by the value PWxV<5:0>

TPWM TUNING PULSE WIDTH MODULATOR

The device has a single 14-bit PWM that can be used for Voltage Synthesis Tuning. The method of operation is similar to the normal PWM except the repetition period is 42.66us.

TPWM Control

Two SFRs are used to control the TPWM, they are TDACL and TDACH. The TPWM is enabled by setting the TPWE bit in the TDACH SFR. The most significant bits TD<13:7> alter the high period between 0 and 42.33us. The 7 least significant bits TD<6:0> extend certain pulses by a further 0.33us. e.g. if TD<6:0> = 01H then 1 in 128 periods will be extended by 0.33us, if TD<6:0>=02H then 2 in 128 periods will be extended.

The TPWM will not start to output a new value until TDACH has been written to. Therefore, if the value is to be changed, TACL should be written before TDACH.

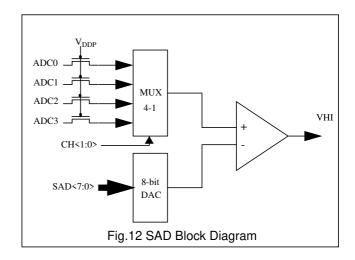
SAD SOFTWARE A/D

Four successive approximation Analogue to Digital Converters can be implemented in software by making use of the on board 8-bit Digital to Analogue Converter and Analogue Comparator.

SAD Control

The control of the required analogue input is done using the channel select bits CH<1:0> in the SAD SFR, this selects the required analogue input to be passed to one of the inputs of the comparator. The second comparator input is generated by the DAC whose value is set by the bits SAD<7:0> in the SAD and SADB SFRs. A comparison between the two inputs is made when the start compare bit ST in the SAD SFR is set, this must be at least one

instruction cycle after the SAD<7:0> value has been set. The result of the comparison is given on VHI one instruction cycle after the setting of ST.



SAD Input Voltage

The external analogue voltage that is used for comparison with the internally generated DAC voltage, does not have the same voltage range due to the 5 V tolerance of the pin. It is limited to V_{DDP} - V_{tn} where V_{tn} is a maximum of 0.75 V. For further details refer to the SAA55XX and SAA56XX Software Analogue to Digital Converter Application Note: SPG/AN99022.

SAD DC Comparator Mode

The SAD module incorporates a DC Comparator mode which is selected using the 'DC_COMP' control bit in the SADB SFR. This mode enables the micro-controller to detect a threshold crossing at the input to the selected analogue input pin (P3.0, P3.1, P3.2 or P3.3) of the Software A/D Converter. A level sensitive interrupt is generated when the analogue input voltage level at the pin falls below the analogue output level of the SAD D/A converter.

This mode is intended to provide the device with a wake-up mechanism from Power-Down or Idle when a key-press on the front panel of the TV is detected. The following software sequence should be used when utilizing this mode for Power-Down or Idle:-

- 1. Disable INT1 using the IE SFR.
- 2. Set INT1 to level sensitive using the TCON SFR.
- 3. Set the D/A Converter digital input level to the desired threshold level using the SAD/SADB SFRs and select the required input pin (P3.0, P3.1, P3.2 or P3,3) using CH1, CH0 in the SAD SFR.
- 4. Enter DC Compare mode by setting the 'DC_COMP' enable bit in the SADB SFR.

TDA935X/6X/8X PS/N2 series

- 5. Enable INT1 using the IE SFR.
- Enter Power-Down/Idle. Upon wake-up the SAD should be restored to its conventional operating mode by disabling the 'DC_COMP' control bit.

I2C Serial I/O Bus

The I²C bus consists of a serial data line (SDA) and a serial clock line (SCL). The definition of the I²C protocol can be found in the 80C51 based 8-bit micro-controllers - Philips Semiconductors (ref. IC20).

The device operates in four modes: -

- Master Transmitter
- Master Receiver
- Slave Transmitter
- Slave Receiver

The micro-controller peripheral is controlled by the Serial Control SFR (S1CON) and its Status is indicated by the status SFR (S1STA). Information is transmitted/received to/from the I²C bus using the Data SFR (S1DAT) and the Slave Address SFR (S1ADR) is used to configure the slave address of the peripheral.

The byte level I²C serial port is identical to the I²C serial port on the 8xC558, except for the clock rate selection bits CR<2:0>. The operation of the subsystem is described in detail in the 8xC558 data sheet and can be found in the 80C51 based 8-bit micro-controllers - Philips Semiconductors (ref. IC20).

Three different IIC selection tables for CR<2:0> can be configured using the ROMBANK SFR (IIC_LUT<1:0>) as follows: -

'558 nominal mode' (iic_lut="00")

This option accommodates the 558 I2C. The various serial rates are shown below: -

CR2	CR1	CR0	f _{clk} (6MHz) divided by	I2C Bit Frequency (KHz) at f _{clk}
0	0	0	60	100
0	0	1	1600	3.75
0	1	0	40	150
0	1	1	30	200
1	0	0	240	25
1	0	1	3200	1.875
1	1	0	160	37.5
1	1	1	120	50

Table 7 IIC Serial Rates '558 nominal mode'

'558 fast mode' (iic_lut="01")

This option accommodates the 558 I²C doubled rates as shown below: -

CR2	CR1	CR0	f _{clk} (6MHz) divided by	I2C Bit Frequency (KHz) at f _{clk}
0	0	0	30	200
0	0	1	800	7.5
0	1	0	20	300
0	1	1	15	400
1	0	0	120	50
1	0	1	1600	3.75
1	1	0	80	75
1	1	1	60	100

Table 8 IIC Serial Rates '558 fast mode'

'558 slow mode' (iic_lut="10")

This option accommodates the 558 I²C rates divided by 2 as shown below: -

CR2	CR1	CR0	f _{clk} (6MHz) divided by	I2C Bit Frequency (KHz) at f _{clk}
0	0	0	120	50
0	0	1	3200	1.875
0	1	0	80	75
0	1	1	60	100
1	0	0	480	12.5
1	0	1	6400	0.9375
1	1	0	320	18.75
1	1	1	240	25

Table 9 IIC Serial Rates '558 slow mode'

Note: In the above tables the f_{clk} relates to the clock rate of the 80c51 IIC module (6MHz).

12C Port Enable

One external I²C port is available. This port is enabled using TXT21.I2C PORT0. Any information transmitted to the device can only be acted upon if the port is enabled. Internal communication between the 80c51 micro-controller and the TV Signal Processor will continue regardless of the value written to TXT21.I2C PORT0.

LED Support

Port pins P0.5 and P0.6 have a 8mA current sinking capability to enable LEDs in series with current limiting resistors to be driven directly, without the need for additional buffering circuitry.

TDA935X/6X/8X PS/N2 series

MEMORY INTERFACE

The memory interface controls the access to the embedded DRAM, refreshing of the DRAM and page clearing. The DRAM is shared between Data Capture, Display and Microcontroller sections. The Data Capture section uses the DRAM to store acquired information that has been requested. The Display reads the DRAM information and converts it to RGB output values. The Microcontroller uses the DRAM as embedded auxiliary RAM.

DATA CAPTURE

The Data Capture section takes in the analogue Composite Video and Blanking Signal (CVBS) from One Chip, and from this extracts the required data, which is then decoded and stored in SFR memory.

The extraction of the data is performed in the digital domain. The first stage is to convert the analogue CVBS signal into a digital form. This is done using an ADC sampling at 12MHz. The data and clock recovery is then performed by a Multi-Rate Video Input Processor (MulVIP). From the recovered data and clock the following data types are extracted WST Teletext (625/525),Closed Caption, VPS, WSS. The extracted data is stored in either memory (DRAM) via the Memory Interface or in SFR locations.

Data Capture Features

- Video Signal Quality detector.
- Data Capture for 625 line WST
- Data Capture for 525 line WST
- Data Capture for US Closed Caption
- Data Capture for VPS data (PDC system A)
- Data Capture for Wide Screen Signalling (WSS) bit decoding
- Automatic selection between 525 WST/625WST
- Automatic selection between 625WST/VPS on line 16 of VBI
- Real-time capture and decoding for WST Teletext in Hardware, to enable optimised microprocessor throughput
- Up to 10 pages stored On-Chip
- Inventory of transmitted Teletext pages stored in the Transmitted Page Table (TPT) and Subtitle Page Table (SPT)
- Automatic detection of FASTEXT transmission
- Real-time packet 26 engine in Hardware for processing accented, G2 and G3 characters
- Signal quality detector for WST/VPS data types
- Comprehensive Teletext language coverage
- Full Field and Vertical Blanking Interval (VBI) data capture of WST data

Analogue to Digital Converter

The CVBS input is passed through a differential to single ended converter (DIVIS), although in this device it is used in single ended configuration with a reference. The analogue output of DIVIS is converted into a digital representation by a full flash ADC with a sampling rate of 12MHz.

Multi Rate Video Input Processor

The multi rate video input processor is a Digital Signal Processor designed to extract the data and recover the clock from the digital CVBS signal.

TDA935X/6X/8X PS/N2 series

Data Standards

The data and clock standards that can be recovered are shown in Table 10 below:-

Data Standard	Clock Rate
625WST	6.9375 MHz
525WST	5.7272 MHz
VPS	5.0 MHz
WSS	5.0 MHz
Closed Caption	500 KHz

Table 10 Data Slicing Standards

Data Capture Timing

The Data Capture timing section uses the Synchronisation information extracted from the CVBS signal to generate the required Horizontal and Vertical reference timings. The timing section automatically recognises and selects the appropriate timings for either 625 (50Hz) synchronisation or 525 (60Hz) synchronisation. A flag TXT12.Video Signal Quality is set when the timing section is locked correctly to the incoming CVBS signal. When TXT12.Video Signal Quality is set another flag TXT12.625/525 SYNC can be used to identify the standard.

Acquisition

The acquisition sections extracts the relevant information from the serial stream of data from the MulVIP and stores it in memory.

625 WST ACQUISITION

The family is capable of acquiring 625-line and 525-line World System Teletext. Teletext pages are identified by seven numbers: magazine (page hundreds), page tens, page units, hours tens, hours units, minutes tens and minutes units. The last four digits, hours and minutes, are known as the subcode, and were originally intended to be time related, hence their names.

Making a page request

A page is requested by writing a series of bytes into the TXT3.PRD<4:0> SFR which correspond to the number of the page required. The bytes written into TXT3 are stored in a RAM with an auto-incrementing address. The start address for the RAM is set using the TXT2.SC<2:0> to define which part of the page request is being written, and TXT2.REQ<3:0> is used to define which of the 10 page requests is being modified. If TXT2.REQ<3:0> is greater

than 09h, then data being written to TXT3 is ignored. Table 11 shows the contents of the page request RAM. Up to 10 pages of teletext can be acquired on the 10 page device, when TXT1.EXT PKT OFF is set to logic 1, and up to 9 pages can be acquired when this bit is set to logic 0. If the 'Do Care' bit for part of the page number is set to 0 then that part of the page number is ignored when the teletext decoder is deciding whether a page being received off air should be stored or not. For example, if the Do Care bits for the 4 subcode digits are all set to 0 then every subcode version of the page will be captured.

Start Column	Byte Identification	PRD<4>	PRD<3>	PRD<2>	PRD<1>	PRD<0>
0	Magazine	DO CARE	HOLD	MAG2	MAG1	MAG0
1	Page Tens	DO CARE	PT3	PT2	PT1	PT0
2	Page Units	DO CARE	PU3	PU2	PU1	PU0
3	Hours Tens	DO CARE	х	х	HT1	HT0
4	Hours Units	DO CARE	HU3	HU2	HU1	HU0
5	Minutes Tens	DO CARE	х	MT2	MT1	MT0
6	Minutes Units	DO CARE	MU3	MU2	MU1	MU0
7	Error Mode	х	х	х	E1	E0

Table 11 The contents of the Page request RAM

Note: MAG = Magazine PT = Page Tens PU = Page Units HT = Hours Tens HU = Hours Units

 $\mathsf{MT} = \mathsf{Minutes} \; \mathsf{Tens} \; \mathsf{MU} = \mathsf{Minutes} \; \mathsf{Units} \; \mathsf{E} = \mathsf{Error} \; \mathsf{check} \; \mathsf{mode}$

When the Hold bit is set to 0 the teletext decoder will not recognise any page as having the correct page number and no pages will be captured. In addition to providing the user requested hold function this bit should be used to prevent the inadvertent capture of an unwanted page when a new page request is being made. For example, if the previous page request was for page 100 and this was being changed to page 234, it would be possible to capture page 200 if this arrived after only the requested magazine number had been changed.

The E1 and E0 bits control the error checking which should be carried out on packets 1 to 23 when the page being requested is captured. This is described in more detail in a later section ('Error Checking').

For a multi page device, each packet can only be written into one place in the teletext RAM so if a page matches more than one of the page requests the data is written into the area of memory corresponding to the lowest numbered matching page request.

At power-up each page request defaults to any page, hold on and error check mode 0.

Rolling Headers and Time

When a new page has been requested it is conventional for the decoder to turn the header row of the display green

TDA935X/6X/8X PS/N2 series

and to display each page header as it arrives until the correct page has been found.

When a page request is changed (i.e.: when the TXT3 SFR is written to) a flag (PBLF) is written into bit 5, column 9, row 25 of the corresponding block of the page memory. The state of the flag for each block is updated every TV line, if it is set for the current display block, the acquisition section writes all valid page headers which arrive into the display block and automatically writes an alpha-numerics green character into column 7 of row 0 of the display block every TV line.

When a requested page header is acquired for the first time, rows 1 to 23 of the relevant memory block are cleared to space, i.e.: have 20h written into every column, before the rest of the page arrives. Row 24 is also cleared if the TXT0.X24 POSN bit is set. If the TXT1.EXT PKT OFF bit is set the extension packets corresponding to the page are also cleared.

The last 8 characters of the page header are used to provide a time display and are always extracted from every valid page header as it arrives and written into the display block

The TXT0. DISABLE HEADER ROLL bit prevents any data being written into row 0 of the page memory except when a page is acquired off air i.e.: rolling headers and time are not written into the memory. The TXT1.ACQ OFF bit prevents any data being written into the memory by the teletext acquisition section.

When a parallel magazine mode transmission is being received only headers in the magazine of the page requested are considered valid for the purposes of rolling headers and time. Only one magazine is used even if don't care magazine is requested. When a serial magazine mode transmission is being received all page headers are considered to be valid.

Error Checking

Before teletext packets are written into the page memory they are error checked. The error checking carried out depends on the packet number, the byte number, the error check mode bits in the page request data and the TXT1.8 BIT bit.

If an incorrectable error occurs in one of the Hamming checked addressing and control bytes in the page header or in the Hamming checked bytes in packet 8/30, bit 4 of the byte written into the memory is set, to act as an error flag to the software. If incorrectable errors are detected in any other Hamming checked data the byte is not written into the memory.

Teletext Memory Organisation

The teletext memory is divided into 2 banks of 10 blocks. Normally, when the TXT1.EXT PKT OFF bit is logic 0,

each of blocks 0 to 8 contains a teletext page arranged in the same way as the basic page memory of the page device and block 9 contains extension packets. When the TXT1.EXT PKT OFF bit is logic 1, no extension packets are captured and block 9 of the memory is used to store another page. The number of the memory block into which a page is written corresponds to the page request number which resulted in the capture of the page.

Packet 0, the page header, is split into 2 parts when it is written into the text memory. The first 8 bytes of the header contain control and addressing information. They are Hamming decoded and written into columns 0 to 7 of row 25. Row 25 also contains the magazine number of the acquired page and the PBLF flag but the last 14 bytes are unused and may be used by the software, if necessary.

Row 25 Data Contents

The Hamming error flags are set if the on-board 8/4 Hamming checker detects that there has been an incorrectable (2 bit) error in the associated byte. It is possible for the page to still be acquired if some of the page address information contains incorrectable errors if that part of the page request was a 'don't care'. There is no error flag for the magazine number as an incorrectable error in this information prevents the page being acquired. The interrupted sequence (C9) bit is automatically dealt with by the acquisition section so that rolling headers do not contain a discontinuity in the page number sequence. The magazine serial (C11) bit indicates whether the transmission is a serial or a parallel magazine transmission. This affects the way the acquisition section operates and is dealt with automatically.

The newsflash (C5), subtitle (C6), suppress header (C7), inhibit display (C10) and language control (C12 to 14) bits are dealt with automatically by the display section, described below.

The update (C8) bit has no effect on the hardware. The remaining 32 bytes of the page header are parity checked and written into columns 8 to 39 of row 0. Bytes which pass the parity check have the MSB set to 0 and are written into the page memory. Bytes with parity errors are not written into the memory.

Inventory Page

If the TXT0.INV on bit is 1, memory block 8 is used as an inventory page. The inventory page consists of two tables, - the Transmitted Page Table (TPT) and the subtitle page table (SPT).

In each table, every possible combination of the page tens and units digit, 00 to FFh, is represented by a byte. Each bit of these bytes corresponds to a magazine number so each page number, from 100 to 8FF, is represented by a bit in the table. The bit for a particular page in the TPT is set

TDA935X/6X/8X PS/N2 series

when a page header is received for that page. The bit in the SPT is set when a page header for the page is received which has the 'subtitle' page header control bit (C6) set. The bit for a particular page in the TPT is set when a page header is received for that page. The bit in the SPT is set when a page header for the page is received which has the 'subtitle' page header control bit (C6) set.

Packet 26 Processing

One of the uses of packet 26 is to transmit characters which are not in the basic teletext character set. The family automatically decodes packet 26 data and, if a character corresponding to that being transmitted is available in the character set, automatically writes the appropriate character code into the correct location in the teletext memory. This is not a full implementation of the packet 26 specification allowed for in level 2 teletext, and so is often referred to as level 1.5.

By convention, the packets 26 for a page are transmitted before the normal packets. To prevent the default character data over writing the packet 26 data the device incorporates a mechanism which prevents packet 26 data from being overwritten. This mechanism is disabled when the Spanish national option is detected as the Spanish transmission system sends even parity (i.e. incorrect) characters in the basic page locations corresponding to the characters sent via packet 26 and these will not over write the packet 26 characters anyway. The special treatment of Spanish national option is prevented if TXT12. ROM VER R4 is logic 0 or if the TXT8.DISABLE SPANISH is set.

Packet 26 data is processed regardless of the TXT1. EXT PKT OFF bit, but setting the TXT1.X26 OFF disables packet 26 processing.

The TXT8. Packet 26 received bit is set by the hardware whenever a character is written into the page memory by the packet 26 decoding hardware. The flag can be reset by writing a 0 into the SFR bit.

525 WST

The 525 line format is similar to the 625 line format but the data rate is lower and there are less data bytes per packet (32 rather than 40). There are still 40 characters per display row so extra packets are sent each of which contains the last 8 characters for four rows. These packets can be identified by looking at the 'tabulation bit' (T), which replaces one of the magazine bits in 525 line teletext. When an ordinary packet with T=1 is received, the decoder puts the data into the four rows starting with that corresponding to the packet number, but with the 2 LSBs set to 0. For example, a packet 9 with T=1 (packet X/1/9) contains data for rows 8, 9, 10 and 11. The error checking carried out on data from packets with T=1 depends on the

setting of the TXT1. 8 BIT bit and the error checking control bits in the page request data and is the same as that applied to the data written into the same memory location in the 625 line format.

The rolling time display (the last 8 characters in row 0) is taken from any packets X/1/1, 2 or 3 received. In parallel magazine mode only packets in the correct magazine are used for rolling time. Packet number X/1/0 is ignored. The tabulation bit is also used with extension packets. The first 8 data bytes of packet X/1/24 are used to extend the Fastext prompt row to 40 characters. These characters are written into whichever part of the memory the packet 24 is being written into (determined by the 'X24 Posn' bit). Packets X/0/27/0 contain 5 Fastext page links and the link control byte and are captured, Hamming checked and stored by in the same way as are packets X/27/0 in 625 line text. Packets X/1/27/0 are not captured. Because there are only 2 magazine bits in 525 line text, packets with the magazine bits all set to 0 are referred to as being in magazine 4. Therefore, the broadcast service data packet is packet 4/30, rather than packet 8/30. As in 625 line text, the first 20 bytes of packet 4/30 contain encoded data which is decoded in the same way as that in packet 8/30. The last 12 bytes of the packet contains half of the parity encoded status message. Packet 4/0/30 contains the first half of the message and packet 4/1/30 contains the second half. The last 4 bytes of the message are not written into memory. The first 20 bytes of the each version of the packet are the same so they are stored whenever either version of the packet is acquired. In 525 line text each packet 26 only contains ten 24/18 Hamming encoded data triplets, rather than the 13 found in 625 line text. The tabulation bit is used as an extra bit (the MSB) of the designation code, allowing 32 packet 26s to be transmitted for each page. The last byte of each packet 26 is ignored.

FASTEXT DETECTION

When a packet 27, designation code 0 is detected, whether or not it is acquired, the TXT13. FASTEXT bit is set. If the device is receiving 525 line teletext, a packet X/0/27/0 is required to set the flag. The flag can be reset by writing a 0 into the SFR bit.

BROADCAST SERVICE DATA DETECTION

When a packet 8/30 is detected, or a packet 4/30 when the device is receiving a 525 line transmission, the TXT13. Packet 8/30. The flag can be reset by writing a 0 into the SFR bit.

VPS ACQUISITION

When the TXT0. VPS ON bit is set, any VPS data present on line 16, field 0 of the CVBS signal at the input of the

TDA935X/6X/8X PS/N2 series

teletext decoder is error checked and stored in row 25, block 9 of the basic page memory. The device automatically detects whether teletext or VPS is being transmitted on this line and decodes the data appropriately.

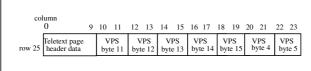


Fig.13 VPS Data Storage

Each VPS byte in the memory consists of 4 bi-phase decoded data bits (bits 0-3), a bi-phase error flag (bit 4) and three 0s (bits5-7). The TXT13. VPS Received bit is set by the hardware whenever VPS data is acquired. The flag can be reset by writing a 0 into the SFR bit.

WSS ACQUISITION

The Wide Screen Signalling data transmitted on line 23 gives information on the aspect ratio and display position of the transmitted picture, the position of subtitles and on the camera/film mode. Some additional bits are reserved for future use. A total of 14 data bits are transmitted. All of the available data bits transmitted by the Wide Screen Signalling signal are captured and stored in SFRs WSS1, WSS2 and WSS3. The bits are stored as groups of related bits and an error flag is provided for each group to indicate when a transmission error has been detected in one or more of the bits in the group. Wide screen signalling data is only acquired when the TXT8.WSS ON bit is set. The TXT8.WSS RECEIVED bit is set by the hardware whenever wide screen signalling data is acquired. The flag can be reset by writing a 0 into the SFR bit.

CLOSED CAPTION ACQUISITION

The US Closed Caption data is transmitted on line 21 (525 line timings) and is used for Captioning information, Text information and Extended Data Services. Closed Caption data is only acquired when TXT21.CC ON bit is set. Two bytes of data are stored per field in SFRs, the first bye is stored in CCDAT1 and the second byte is stored in CCDAT2. The value in the CCDAT registers are reset to 00h at the start of the Closed Caption line defined by CCLIN.CS<4:0>. At the end of the Closed Caption line an interrupt is generated if IE.ECC is active.

The processing of the Closed Caption data to convert into a displayable format is performed by Software.

DISPLAY

The display section is based on the requirements for a Level 1.5 WST Teletext and US Closed Caption. There are some enhancements for use with locally generated On-Screen Displays.

The display section reads the contents of the Display memory and interprets the control/character codes. Using this information and other global settings, the display produces the required RGB signals and Video/Data (Fast Blanking) signal for the TV signal processing. The display is synchronised to the TV signal processing by

way of Horizontal and Vertical sync signals generated within TDA935X/6x/8x. From these signals all display timings are derived.

Display Features

- Teletext and Enhanced OSD modes
- Level 1.5 WST features
- US Closed Caption Features
- · Serial and Parallel Display Attributes
- Single/Double/Quadruple Width and Height for characters
- · Scrolling of display region.
- Variable flash rate controlled by software.
- Globally selectable scan lines per row 9/10/13/16.
- Globally selectable character matrix (HxV) 12x9, 12x10, 12x13, 12x16.
- Italics, Underline and Overline.
- Soft Colours using CLUT with 4096 colour palette.
- Fringing (Shadow) selectable from N-S-E-W direction.
- Fringe colour selectable.
- · Meshing of defined area.
- Contrast reduction of defined area.
- · Cursor.
- Special Graphics characters with two planes, allowing four colours per character.
- 32 Software re-definable On-Screen Display characters.
- 4 WST Character sets(G0/G2) in single device (e.g. Latin,Cyrillic,Greek,Arabic).
- G1 Mosaic graphics, Limited G3 Line drawing characters.
- WST Character sets and Closed Caption Character set in single device.

Display Modes

The display section has two distinct modes with different features available in each. The two modes are:

TDA935X/6X/8X PS/N2 series

- TXT:This is the display configured as the WST mode with additional serial and global attributes to enable the same functionality as the SAA5497 (ETT) device.The display is configured as a fixed 25 rows with 40 characters per row.
- CC:This is the display configured as the US Closed Caption mode with the same functionality as the PC83C771 device. The display is configured as a maximum of 16 rows with a maximum of 48 characters per row.

In both of the above modes the Character matrix, and TV lines per row can be defined. There is an option of 9, 10, 13 & 16 TV lines per display row, and a Character matrix (HxV) of 12x9, 12x10, 12x13, or 12x16. Not all combinations of TV lines per row and maximum display rows give a sensible OSD display, since there is limited number of TV scan lines available. Special Function Register, TXT21 is used to control the

character matrix and lines per row.

Display Features available in each mode

The following is a list of features available in each mode. Each setting can either be a serial or parallel attribute, and some have a global effect on the display.

Feature	TXT	CC
Flash	serial	serial
Boxes	Txt/OSD (Serial)	serial
Horizontal Size	x1/x2/x4 (serial)	x1/x2 (serial)
Vertical Size	x1/x2 (serial) x4 (global)	x1/x2 (serial)
Italic	N/A	serial
Foreground colours	8 (serial)	8+8 (parallel)
Background colours	8 (serial)	16 (serial)
Soft Colours (CLUT)	16 from 4096	16 from 4096
Underline	N/A	serial
Overline	N/A	serial
Fringe	N+S+E+W	N+S+E+W
Fringe Colour	16 (Global)	16 (Serial)
Meshing of Background	Black or Colour (Global)	All (Global)
Fast Blanking Polarity	YES	YES
Screen Colour	16 (Global)	16 (Global)
DRCS	32 (Global)	32 (Global)
Character Matrix (HxV)	12x9/10/13/16	12x9/10/13/16
No. of Rows	25	16
No. of Columns	40	48
No of Characters displayable	1000	768
Cursor	YES	YES
Special Graphics (2 planes per character)	16	16
Scroll	NO	YES

Table 12 Display Features

TDA935X/6X/8X PS/N2 series

Display Feature Descriptions

FLASH

Flashing causes the foreground colour pixel to be displayed as the background pixels. The flash frequency is controlled by software setting and resetting display register REG0: Status at the appropriate interval. CC: This attribute is valid from the time set (see Table 18) until the end of the row or until otherwise modified. TXT: This attribute is set by the control character 'flash' (08h) and remains valid until the end of the row or until reset by the control character 'steady' (09h).

BOXES

CC: This attribute is valid from the time set until end of row or otherwise modified if set with Serial Mode 0. If set with Serial Mode 1, then it is set from the next character onwards.

In CC text mode the background colour is displayed regardless of the setting of the box attribute bit. Boxes take affect only during mixed mode, where boxes are set in this mode the background colour is displayed. Character locations where boxes are not set show video/screen colour (depending on the setting in the display control register. REG0: Display Control) in stead of the background colour.

TXT: Two types of boxes exist the Teletext box and the OSD box. The Teletext box is activated by the 'start box' control character (0Bh), Two start box characters are required begin a Teletext box, with box starting between the 2 characters. The box ends at the end of the line or after a 'end box' control character.

TXT mode can also use OSD boxes, they are started using size implying OSD control chracters(BCh/BDh/BEh/BFh). The box starts after the control character ('set after') and ends either at the end of the row or at the next size implying OSD character ('set at'). The attributes flash, teletext box, conceal, separate graphics, twist and hold graphics are all reset at the start of an OSD box, as they are at the start of the row. OSD Boxes are only valid in TV mode which is defined by TXT5=03h and TXT6=03h.

SIZE

The size of the characters can be modified in both the horizontal and vertical directions.

CC: Two sizes are available in both the horizontal and vertical directions. The sizes available are normal (x1), double (x2) height/width and any combination of these. The attribute setting is always valid for the whole row. Mixing of sizes within a row is not possible.

TXT: Three horizontal sizes are available normal(x1),double(x2),quadruple(x4). The control characters 'normal size' (0Ch/BCh) enables normal size,

the 'double width' or double size (0Eh/BEh/0Fh/BFh) enables double width characters. Any two consecutive combination of 'double width' or 'double size' (0Eh/BEh/0Fh/Bfh) activates quadruple width characters, provided quadruple width characters are enabled by TXT4.Quad Width Enable. Three vertical sizes are available normal(x1),double(x2),quadruple(x4). The control characters 'normal size' (0Ch/BCh) enable normal size, the 'double height' or 'double size' (0Dh/BDh/0Fh/BFh) enable double height characters. Quadruple height character are achieved by using double height characters and setting the global attributes TXT7.Double Height (expand) and TXT7.Bottom/Top. If double height characters are used in teletext mode, single height characters in the lower row of the double height character are automatically disabled.

ITALIC

CC: This attribute is valid from the time set until the end of the row or otherwise modified. The attribute causes the character foreground pixels to be offset horizontally by 1 pixel per 4 scan lines (interlaced mode). The base is the bottom left character matrix pixel. The pattern of the character is indented as shown in Fig.14.

TDA935X/6X/8X PS/N2 series

TXT: The Italic attribute is not available.

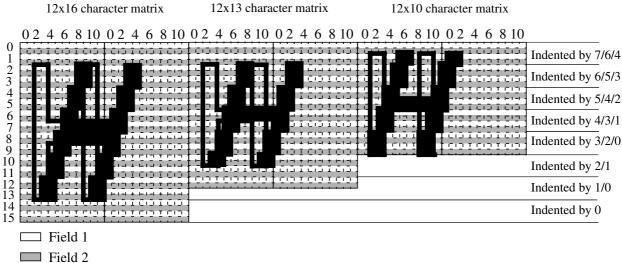


Fig.14 Italic Characters (12x10, 12x13 & 12x16).

COLOURS

CLUT (Colour Look Up Table)

A CLUT (Colour Look Up Table) with 16 colour entries is provided. The colours are programmable out of a palette of 4096(4 bits per R, G and B). The CLUT is defined by writing data to a RAM that resides in the MOVX address space of the 80C51.

RED3-0 b11b4	GRN3-0 b7b4	BLU3-0 b3b0	Colour entry
0000	0000	0000	0
0000	0000	1111	1
1111	1111	0000	14
1111	1111	1111	15

Table 13 CLUT Colour values

Foreground Colour

CC: The foreground colour can be chosen from 8 colours on a character by character basis. Two sets of 8 colours are provided. A serial attribute switches between the banks (see Table 18 Serial Mode 1, bit 7). The colours are the CLUT entries 0 to 7 or 8 to 15.

TXT: The foreground colour is selected via a control character. The colour control characters takes effect at the start of the next character ("Set-After") and remain valid until the end of the row, or until modified by a control character. Only 8 foreground colours are available. The TEXT foreground control characters map to the CLUT entries as shown below:

Control Code	Defined Colour	CLUT Entry
00h	Black	0
01h	Red	1
02h	Green	2
03h	Yellow	3
04h	Blue	4
05h	Magenta	5
06h	Cyan	6
07h	White	7

Table 14 Foreground CLUT mapping

Background Colour

CC: This attribute is valid from the time set until end of row or otherwise modified if set with Serial Mode 0. If set with Serial Mode 1, then the colour is set from the next character onwards.

The background colour can be chosen from all 16 CLUT entries.

TDA935X/6X/8X PS/N2 series

TXT: The control character "New background" ("1Dh") is used to change the background colour to the current foreground colour. The selection is immediate ("Set at") and remains valid until the end of the row or until otherwise modified.

The TEXT background control characters map to the CLUT entries as shown below:

Control Code	Defined Colour	CLUT Entry
00h+1Dh	Black	8
01h+1Dh	Red	9
02h+1Dh	Green	10
03h+1Dh	Yellow	11
04h+1Dh	Blue	12
05h+1Dh	Magenta	13
06h+1Dh	Cyan	14
07h+1Dh	White	15

Table 15 Background CLUT mapping

BACKGROUND DURATION

The attribute when set takes effect from the current position until to the end of the text display defined in REG4:Text Area End.

CC: The background duration attribute (see Table 18, Serial Mode 1, bit 8) in combination with the End Of Row attribute (see Table 18, Serial Mode 1, bit 9) forces the background colour to be display on the row until the end of the text area is reached.

TXT: This attribute is not available.

UNDERLINE

The underline attribute causes the characters to have the bottom scan line of the character cell forced to foreground colour, including spaces. If background duration is set, then underline is set until the end of the text area.

CC: The underline attribute (see Table 18, Serial Mode 0/1, bit 4) is valid from the time set until end of row or otherwise modified.

TXT: This attribute is not available.

OVERLINE

The overline attribute causes the characters to have the top scan line of the character cell forced to foreground colour, including spaces. If background duration is set, then overline is set until the end of the text area.

CC: The overline attribute (see Table 18, Serial Mode 0/1, bit 5) is valid from the time set until end of row or otherwise modified. Overlining of Italic characters is not possible. TXT: This attribute is not available.

END OF ROW

CC: The number of characters in a row is flexible and can determined by the end of row attribute (see Table 18, Serial Mode 1, bit 9). However the maximum number of character positions displayed is determined by the setting of the REG2:Text Position Horizontal and REG4:Text Area

NOTE: When using the end of row attribute the next character location after the attribute should always be occupied by a 'space'.

TXT: This attribute is not available, Row length is fixed at 40 characters.

FRINGING

A fringe (shadow) can be defined around characters. The fringe direction is individually selectable in any of the North, South, East and West direction using REG3:Fringing Control. The colour of the fringe can also be defined as one of the entries in the CLUT, again using REG3:Fringing Control.

CC: The fringe attribute (see Table 18, Serial Mode 0, bit 9) is valid from the time set until the end of the row or otherwise modified.

TXT: The display of fringing in TXT mode is controlled by the TXT4.SHADOW bit. When set all the alphanumeric characters being displayed are shadowed, graphics characters are not shadowed.

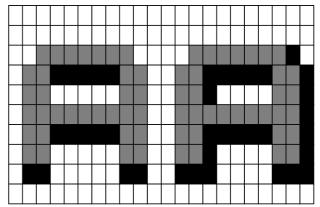


Fig.15 South and Southwest Fringing

MESHING

The attribute effects the background colour being displayed. Alternate pixels are displayed as the background colour or video. The structure is offset by 1

TDA935X/6X/8X PS/N2 series

pixel from scan line to scan line, thus achieving a checker board display of the background colour and video. TXT: There are two meshing attributes one that only affects black background colours TXT4.BMESH and a second that only affects backgrounds other than black TXT4.CMESH. A black background is defined as CLUT entry 8, a none black background is defined as CLUT entry 9-15.

CC: The setting of the Mesh bit in REG0:Display Control has the effect of meshing any background colour.

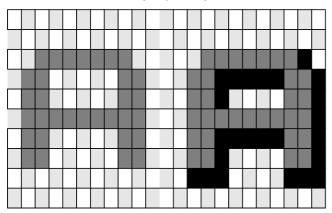


Fig.16 Meshing and Meshing / Fringing (South+West)

Cursor

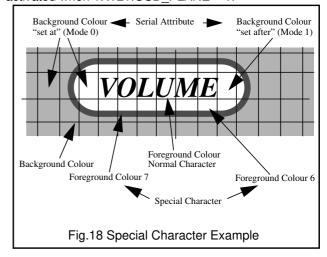
The cursor operates by reversing the background and foreground colours in the character position pointed to by the active cursor position. The cursor is enabled using TXT7.CURSOR ON. When active, the row the cursor appears on is defined by TXT9.R<4:0> and the column is defined by TXT10.C<5:0>. The position of the cursor can be fixed using TXT9.CURSOR FREEZE.

CC: The valid range for row is 0 to 15. The valid range for column is 0 to 47. The cursor remains rectangular at all times, it's shape is not affected by italic attribute, therefore it is not advised to use the cursor with italic characters. TXT: The valid range for row positioning is 0 to 24. The valid range for column is 0 to 39.



SPECIAL GRAPHICS CHARACTERS

CC/TXT: Several special characters are provided for improved OSD effects. These characters provide a choice of 4 colours within a character cell. The total number of special graphics characters is limited to 16. They are stored in the character codes 8Xh and 9Xh of the character table (32 ROM characters), or in the DRCs which overlay character codes 8Xh and 9Xh. Each special graphics character uses two consecutive normal characters. Fringing, underline and overline is not possible for special graphics characters. Special graphics characters are activated when TXT21.OSD PLANE = 1.



The example in Fig.18 can be done with 8 special graphics characters.

If the screen colour is transparent (implicit in mixed mode) and inside the object the box attribute is set, then the object is surrounded by video. If the box attribute is not set the background colour inside the object will also be displayed as transparent.

Plane 1 0	Colour Allocation	
0 0	Background Colour	
0 1	Foreground Colour	
1 0	CLUT entry 6	
1 1	CLUT entry 7	

Table 16 Special Character Colour allocation

TDA935X/6X/8X PS/N2 series

Character and Attribute Coding

CC MODE

Character coding is split into character oriented attributes (parallel) and character group coding (serial). The serial attributes take effect either at the position of the attribute (Set At), or at the following location (Set After) and remain effective until either modified by a new serial attribute or until the end of the row. A serial attribute is represented as a space (the space character itself however is not used for this purpose), the attributes that are still active, e.g. overline and underline will be visible during the display of the space. The default setting at the start of a row is:

- 1x size, flash and italics OFF
- · overline and underline OFF
- Display mode = superimpose
- fringing OFF
- background colour duration = 0
- end of row = 0

The coding is done in 12 bit words. The codes are stored sequentially in the display memory. A maximum of 768 character positions can be defined for a single display.

PARALLEL CHARACTER CODING

Bits	Description	
0-7	8 bit character code	
8-10	3 bits for 8 foreground colours	
11	Mode bit: 0 = Parallel code	

Table 17 Parallel Character Coding

TDA935X/6X/8X PS/N2 series

SERIAL CHARACTER CODING

Bits		Description			
	Serial Mode 0	Mode 1			
	("set at")	Char.Pos. 1 ("set at")	Char.Pos. >1 ("set after")		
0-3	4 bits for 16 Background colours	4 bits for 16 Background colours	4 bits for 16 Background colours		
4	0 = Underline OFF 1 = Underline ON	Horizontal Size: $0 = \text{Underline OFF}$ 0 = normal $1 = X2$			
5	0 = Overline OFF 1 = Overline ON	Vertical Size: 0 = normal 1 = x2	0 = Overline OFF 1 = Overline ON		
6	Display mode: 0 = Superimpose 1 = Boxing	Display mode: 0 = Superimpose 1 = Boxing	Display mode: 0 = Superimpose 1 = Boxing		
7	0 = Flash OFF 1 = Flash ON	Foreground colour switch 0 = Bank 0 (colours 0-7) 1 = Bank 1 (colours 8-15)	Foreground colour switch 0 = Bank 0 (colours 0-7) 1 = Bank 1 (colours 8-15)		
8	0 = Italics OFF 1 = Italics ON	Background colour duration: 0 = stop BGC 1 = set BGC to end of row Background colour duration: (set at): 0 = stop BGC 1 = set BGC to end of row			
9	0 = Fringing OFF 1 = Fringing ON	End of Row $0 = \text{Continue Row}$ $1 = \text{End Row}$ End of Row (set at): $0 = \text{Continue Row}$ $1 = \text{End Row}$			
10	Switch for Serial coding mode 0 and 1:	Switch for Serial coding mode 0 and 1: Switch for Serial coding mode 1:			
	0 = mode 0	1 = mode 1	1 = mode 1		
11	Mode bit:	Mode bit:	Mode bit:		
	1 = Serial code	1 = Serial code	1 = Serial code		

Table 18 Serial Character Coding

TXT MODE

Character coding is in a serial format, with only one attributes being changed at any single location. The serial attributes take effect either at the position of the attribute

(Set At), or at the following location (Set After). The attribute remainseffective until either modified by new serial attributes or until the end of the row. The default settings at the start of a row is:

- foreground colour white (CLUT Address 7)
- background colour black (CLUT Address 8)

TDA935X/6X/8X PS/N2 series

- Horizontal size x1, Vertical size x1 (normal size)
- Alphanumeric ON
- · Contiguous Mosaic Graphics
- · Release Mosaics
- Flash, Box, Conceal and Twist OFF

The attributes have individual codes which are defined in the basic character table below:

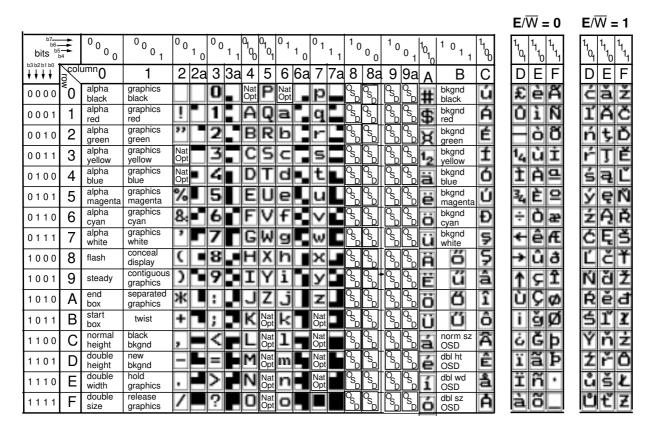


Fig.19 TXT Basic Character Set (Pan-European)

TDA935X/6X/8X PS/N2 series

Screen and Global Controls

A number of attributes are available that affect the whole display region, and cannot be applied selectively to regions of the display.

TV SCAN LINES PER ROW

The number of TV scan lines per field used for each display row can be defined, the value is independent of the character size being used. The number of lines can be either 10/13/16 per display row. The number of TV scan lines per row is defined TXT21.DISP_LINES<1:0>. A value of 9 lines per row can be achieved if the display is forced into 525 line display mode by

TXT17.DISP_FORCE<1:0>, or if the device is in 10 line mode and the automatic detection circuitry within display finds 525 line display syncs.

CHARACTER MATRIX (HXV)

There are four different character matrices available, these are 12x10, 12x13, and 12x16. The selection is made using TXT21.CHAR_SIZE<1:0> and is independent of the number of display lines per row.

If the character matrix is less than the number of TV scan lines per row then the matrix is padded with blank lines. If the character matrix is greater than the number of TV scan lines then the character is truncated.

DISPLAY MODES

CC: When attributes superimpose or when boxing (see Table 18, Serial Mode 0/1, bit 6) is set, the resulting display depends on the setting of the following screen control mode bits in REG0:Display Control.

Display Mode	MOD 1 0	Description
Video	0 0	Video mode disables all display activities and sets the RGB to true black and VDS to video.
Full Text	0 1	Full Text mode displays screen colour at all locations not covered by character foreground or background colour. The box attribute has no effect.
Mixed Screen Colour	1 0	Mixed Screen mode displays screen colour at all locations not covered by character foreground, within boxed areas or, background colour.
Mixed Video	1 1	Mixed Video mode displays video at all locations not covered by character foreground, within boxed areas or, background colour.

Table 19 Display Modes

TXT: The display mode is controlled by the bits in the TXT5 and TXT6. There are 3 control functions - Text on, Background on and Picture on. Separate sets of bits are used inside and outside Teletext boxes so that different display modes can be invoked. TXT6 is used if the newsflash (C5) or subtitle (C6) bits in row 25 of the basic page memory are set otherwise TXT5 is used. This allows the software to set up the type of display required on newsflash and subtitle pages (e.g. text inside boxes, TV picture outside) this will be invoked without any further software intervention when such a page is acquired.

Picture On	Text On	Background On	Effect
0	0	x	Text mode, black screen
0	1	0	Text mode, background always black
0	1	1	Text mode
1	0	x	Video mode
1	1	0	Mixed text and TV mode
1	1	1	Text mode, TV picture outside text area

Table 20 TXT Display Control Bits

TDA935X/6X/8X PS/N2 series

Screen Colour

Screen colour is displayed from 10.5 ms to 62.5 ms after the active edge of the HSync input and on TV lines 23 to 310 inclusive, for a 625 line display, and lines 17 to 260 inclusive for a 525 line display.

The screen colour is defined by REG0:Display Control and points to a location in the CLUT table. The screen colour covers the full video width. It is visible when the Full Text or Mixed Screen Colour mode is set and no foreground or background pixels are being displayed.

Text Display Controls

TEXT DISPLAY CONFIGURATION

Two types of area are possible. The one area is static and the other is dynamic. The dynamic area allows scrolling of a region to take place. The areas cannot cross each other. Only one scroll region is possible.

Display Map

The display map allows a flexible allocation of data in the memory to individual rows.

Sixteen words are provided in the display memory for this purpose. The lower 10 bits address the first word in the memory where the row data starts. This value is an offset in terms of 16-bit words from the start of Display Memory (8000 Hex). The most significant bit enables the display when not within the scroll (dynamic) area.

The display map memory is fixed at the first 16 words in the closed caption display memory.

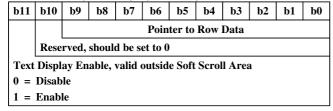
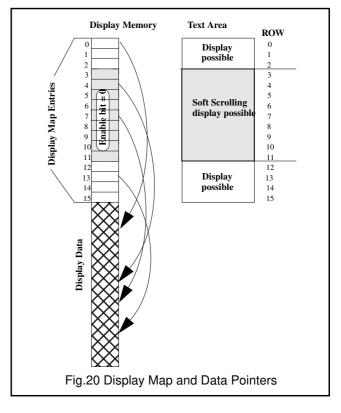


Table 21 Display map Bit Allocation



SOFT SCROLL ACTION

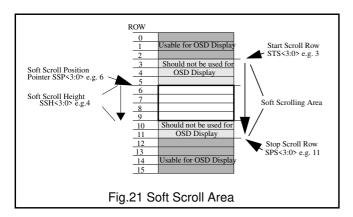
The dynamic scroll region is defined by the REG5:Scroll Area, REG6:Scroll Range, REG14:Top Scroll line and the REG8:Status Register. The scroll area is enabled when the SCON bit is set in REG8: Status.

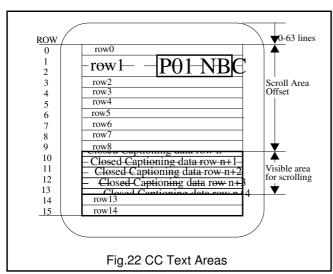
The position of the soft scroll area window is defined using the Soft Scroll Position (SSP<3:0), and the height of the window is defined using the Soft Scroll Height (SSH<3:0>) both are in REG6:Scroll Range. The rows that are scrolled through the window are defined using the Start Scroll Row (STS<3:0>) and the Stop Scroll Row (SPS<3:0>) both are in REG5:Scroll Area.

The soft scrolling function is done by modifying the Scroll Line (SCL<3:0>) in REG14: Top Scroll Line. and the first scroll row value SCR<3:0> in REG8:Status. If the number of rows allocated to the scroll counter is larger than the defined visible scroll area, this allows parts of rows at the top and bottom to be displayed during the scroll function. The registers can be written throughout the field and the values are updated for display with the next field sync. Care should be taken that the register pairs are written to by the software in the same field.

Only a region that contains only single height rows or only double height rows can be scrolled.

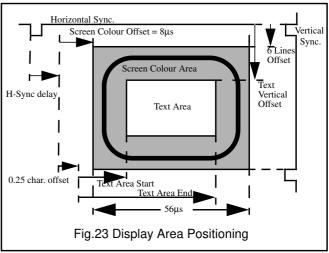
TDA935X/6X/8X PS/N2 series





Display Positioning

The display consists of the **Screen Colour** covering the whole screen and the **Text Area** that is placed within the visible screen area. The screen colour extends over a large vertical and horizontal range so that no offset is needed. The text area is offset in both directions relative to the vertical and horizontal sync pulses.



SCREEN COLOUR DISPLAY AREA

This area is covered by the screen colour. The screen colour display area starts with a fixed offset of 8 us from the leading edge of the horizontal sync pulse in the horizontal direction. A vertical offset is not necessary.

Horizontal	starts 8 us after the leading edge of H-Sync for 56 us.
Vertical	line 9, field 1 (321, field 2) with respect to leading edge of vertical sync (line numbering using 625 Standard).

Table 22 Screen Colour Display Area

TEXT DISPLAY AREA

The text area can be defined to start with an offset in both the horizontal and vertical direction.

Horizontal	Up to 48 full sized characters per row. Start position setting from 3 to 64 characters from the leading edge of H-Sync. Fine adjustment in quarter characters.
Vertical	256 lines (nominal 41- 297). Start position setting from leading edge of vertical sync legal values are 4 to 64 lines. (line numbering using 625 Standard)

Table 23 Text Display Area

The horizontal offset is set in REG2: Text Area Start. The offset is done in full width characters using TAS<5:0> and quarter characters using HOP<1:0> for fine setting. The

TDA935X/6X/8X PS/N2 series

values 00h to 03h for TAS<5:0> will result in a corrupted display.

The width of the text area is defined in REG4:Text Area End by setting the end character value TAE<5:0>. This number determines where the background colour of the Text Area will end if set to extend to the end of the row. It will also terminate the character fetch process thus eliminating the necessity of a row end attribute. This entails however writing to all positions.

The vertical offset is set in REG1:Text Position Vertical Register. The offset value VOL<5:0> is done in number of TV scan lines.

NOTE: REG1:Text Position Vertical Register should not be set to 00 Hex as the Display Busy interrupt is not generated in these circumstances.

Character Set

To facilitate the global nature of the device the character set has the ability to accommodate a large number of characters, which can be stored in different matrices.

CHARACTER MATRICES

The character matrices that can be accommodated are: - (HxVxPlanes) 12x9x1, 12x10x1, 12x13x1, 12x16x1. These modes allow two colours per character position. In CC mode two additional character matrices are available to allow four colours per character: - (HxVxPlanes) 12x13x2, 12x16x2.

The characters are stored physically in ROM in a matrix of size either 12x10 or 12x16.

CHARACTER SET SELECTION

Four character sets are available in the device. A set can consist of alphanumeric characters as required by the WST Teletext or FCC Closed Captioning, Customer definable On-Screen Display characters, and Special Graphic characters.

CC:- Only a single character set can be used for display and this is selected using the Basic Set selection TXT18.BS<1:0>. When selecting a character set in CC mode the Twist Set selection TXT18.TS<1:0> should be set to the same value as TXT18.BS<1:0> for correct operation.

TXT:- Two character sets can be displayed at once. These are the basic G0 set or the alternative G0 set (Twist Set). The basic set is selected using TXT18.BS<1:0>, The alternative/twist character set is defined by TXT19.TS<1:0>. Since the alternative character set is an option it can be enabled or disabled using TXT19.TEN, and the language code that is defined for the alternative set is defined by TXT19.TC<2:0>.

ROM ADDRESSING

Three ROM's are used to generate the correct pixel information. The first contains the National Option look-up table, the second contains the Basic Character look-up table and the third contains the Character Pixel information. Although these are individual ROM, since they do not need to be accessed simultaneously they are all combined into a single ROM unit.

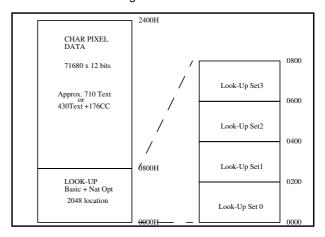


Fig.24 ROM Organisation

TDA935X/6X/8X PS/N2 series

CHARACTER TABLE

The character table is shown in Table 24:-

							Cha	aracter	code c	olumn	s (Bits	4-7)					
		0	1	2	3	4	5	6	7	8	9	A	В	C	D	Е	F
	0		®	SP	0	@	P	ú	р								
	1		0	••	1	A	Q	a	q								
	2		1/2	:	2	В	R	b	r								
	3		ડ	#	3	C	S	c	s								
0-3	4		TM	\$	4	D	T	d	t								
Character code rows (Bits 0-3)	5		¢	%	5	E	U	e	u								
Ns (I	6		£	&	6	F	V	f	v								
[] L	7		-	,	7	G	W	g	w								
oge	8		à	(8	Н	X	h	X								
er c	9		1)	9	I	Y	i	y								
-ract	A		è	á	:	J	Z	j	z								
Cha	В		â	+	;	K	[k	ç								
	C		ê	,	<	L	é	l									
	D		î	•	II	M]	m	Ñ								
	E		ô	•	>	N	í	n	ñ								
	F		û	/	?	0	ó	0	n								

Table 24 Closed Caption Character Table

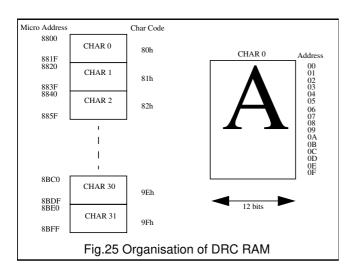
Special Characters are in column 8.
Additional table locations for normal characters
Table locations for normal characters

Re-definable Characters

A number of Dynamically Re-definable Characters (DRC) are available. These are mapped onto the normal character codes, and replace the pre-defined OTP character Rom value.

There are 32 DRCs which occupy character codes 80H to 9FH. Alternatively, These locations can be utilized as 16 special graphics characters. The remapping of the standard OSD to the DRCs is activated when the TXT21.DRCS ENABLE bit is set. The selection of Normal or Special OSD symbols is defined by the TXT21.OSD PLANES. Each character is stored in a matrix of 16x16x1 (V x H x planes), this allows for all possible character matrices to be defined within a single location.

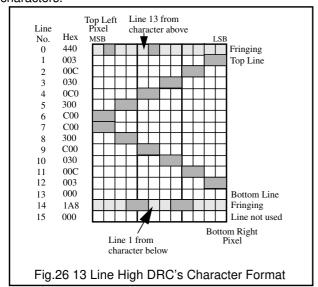
TDA935X/6X/8X PS/N2 series



DEFINING CHARACTERS

The DRC RAM is mapped on to the 80C51 RAM address space and starts at location 8800H. The character matrix is 12 bits wide and therefore requires two bytes to be written for each word, the first byte (even addresses), addresses the lower 8 bits and the second byte (odd addresses) addresses the upper 4 bits.

For characters of 9, 10 or 16 lines high the pixel information starts in the first address and continues sequentially for the required number of addresses. Characters of 13 lines high are defined with an initial offset of 1 address, this is to allow for correct generation of fringing across boundaries of clustered characters (see Fig.26). The characters continue sequentially for 13 lines after which a further line can again be used for generation of correct fringing across boundaries of clustered characters.



DRCs are defined by writing data to the DRC RAM using the 80C51 MOVX command. Setting bits 3 to 9 of the first line of a 12 wide by 16 line character would require setting the high byte of the 80C51 data pointer to 88H, the low byte of the 80C51 data pointer to 00H, using the MOVX command to load address 8800H with data F8H, incrementing the data pointer, and finally using the MOVX command to load address 8801H with data 03H.

Display Synchronization

The horizontal and vertical synchronizing signals from the TV deflection are used as inputs. Both signals can be inverted before being delivered to the Phase Selector section.

CC: The polarity is controlled using either VPOL or HPOL in REG2:Text position Vertical.

TXT: SFRs bits TXT1.HPOL & TXT1.VPOL control the polarity.

A line locked 12 MHz clock is derived from the 12MHz free running oscillator by the Phase Selector. This line locked clock is used to clock the whole of the Display block. The H & V Sync signals are synchronized with the 12 MHz clock before being used in the display section.

Video/Data Switch (Fast Blanking) Polarity

The polarity of the Video/Data (Fast Blanking) signal can be inverted. The polarity is set with the VDSPOL in REG7: RGB Brightness register.

VDSP OL	VDS	Condition
0	1	RGB display
0	0	Video Display
1	0	RGB display
1	1	Video Display

Table 25 Fast Blanking Signal Polarity

Video/Data Switch Adjustment

To take into account the delay between the RGB values and the VDS signal due to external buffering, the VDS signal can be moved in relation to the RGB signals. The VDS signal can be set to be either a clock cycle before or after the RGB signal, or coincident with the RGB signal. This is done using VDEL<2:0> in REG15:Configuration.

TDA935X/6X/8X PS/N2 series

RGB Brightness Control

A brightness control is provided to allow the RGB upper output voltage level to be modified. The RGB amplitude may be varied between 60% and 100%.

The brightness is set in the RGB Brightness register as follows: -

BRI3-0	RGB Brightness
0 0 0 0	Lowest value
1 1 1 1	Highest value

Table 26 RGB Brightness

Contrast Reduction

TXT: The COR bits in SFRs TXT5 & TXT6 control when the COR output of the device is activated (i.e. Pulled-low).

This output is intended to act on the TV's display circuits to reduce contrast of the video when it is active. The result of contrast reduction is to improve the readability of the text in a mixed teletext and video display.

The bits in the TXT5 & TXT6 SFRs allow the display to be set up so that, for example, the areas inside teletext boxes will be contrast reduced when a subtitle is being displayed but that the rest of the screen will be displayed as normal video.

CC: This feature is not available in CC mode.

Memory Mapped Registers

The memory mapped registers are used to control the display. The registers are mapped into the Microcontroller MOVX address space, starting at address 87F0h and extending to 87FF.

MMR MAP

ADD	R/W	Names	BIT7	BIT6	BIT5	BIT4	BIT3	BIT2	BIT1	BIT0
87F0	R/W	Display Control	SRC<3>	SRC<2>	SRC<1>	SRC<0>	-	MSH	MOD<1>	MOD<0>
87F1	R/W	Text Position Vertical	VPOL	HPOL	VOL<5>	VOL<4>	VOL<3>	VOL<2>	VOL<1>	VOL<0>
87F2	R/W	Text Area Start	HOP<1>	HOP<0>	TAS<5>	TAS<4>	TAS<3>	TAS<2>	TAS<1>	TAS<0>
87F3	R/W	Fringing Control	FRC<3>	FRC<2>	FRC<1>	FRC<0>	FRDN	FRDE	FRDS	FRDW
87F4	R/W	Text Area End	-	-	TAE<5>	TAE<4>	TAE<3>	TAE<2>	TAE<1>	TAE<0>
87F5	R/W	Scroll Area	SSH<3>	SSH<2>	SSH<1>	SSH<0>	SSP<3>	SSP<2>	SSP<1>	SSP<0>
87F6	R/W	Scroll Range	SPS<3>	SPS<2>	SPS<1>	SPS<0>	STS<3>	STS<2>	STS<1>	STS<0>
87F7	R/W	RGB Bright.ness	VDSPOL	-	-	-	BRI<3>	BRI<2>	BRI<1>	BRI<0>
87F8	R	Status read	BUSY	FIELD	SCON	FLR	SCR<3>	SCR<2>	SCR<1>	SCR<0>
87F8	W	Status write	-	-	SCON	FLR	SCR<3>	SCR<2>	SCR<1>	SCR<0>
87FC	R/W	H-Sync. Delay	-	HSD<6>	HSD<5>	HSD<4>	HSD<3>	HSD<3>	HSD<1>	HSD<0>
87FD	R/W	V-Sync. Delay	-	VSD<6>	VSD<5>	VSD<4>	VSD<3>	VSD<2>	VSD<1>	VSD<0>
87FE	R/W	Top Scroll Line	-	-	-	-	SCL<3>	SCL<2>	SCL<1>	SCL<0>
87FF	R/W	Configuration	CC	VDEL<2>	VDEL<1>	VDEL<0>	TXT/V	-	-	-

Table 27 MMR Memory Map

TDA935X/6X/8X PS/N2 series

MMR BIT DEFINITION

Names	ADD	BIT7	BIT6	BIT5	BIT4	BIT3	BIT2	BIT1	BIT0	RESET			
Display Control.	87F0	SRC<3>	SRC<2>	SRC<1>	SRC<0>	-	MSH	MOD<1>	MOD<0>	00H			
SRC<3:0>	Screen C	olour definition		'					•	!			
MSH		eshing of backg ing all backgrou											
MOD<1:0>	00 - Vide 01 - Full 10 - Mixe 11 - Mixe	Text ed Screen Colou	r										
Text Position Vertical	87F1	VPOL	HPOL	VOL<5>	VOL<4>	VOL<3>	VOL<2>	VOL<1>	VOL<0>	00H			
VPOL	0 - Input 1 - Invert	polarity ted input polarity	<i>y</i>										
HPOL	0 - Input 1 - Invert	Polarity ted input polarity	V										
VOL<5:0>	Display s	play start Vertical Offset from V-Sync. (lines)											
Text Area Start	87F2	HOP<1>	HOP<0>	TAS<5>	TAS<4>	TAS<3>	TAS<2>	TAS<1>	TAS<0>	00H			
HOP<1:0>	Fine Hor	Fine Horizontal Offset in quarter of characters											
TAS<5:0>	Text area	start											
Fringing Control.	87F3	FRC<3>	FRC<2>	FRC<1>	FRC<0>	FRDN	FRDE	FRDS	FRDW	00H			
FRC<3:0>	Fringing	colour, value ad	dress of CLUT							•			
FRDN		inge in North direct											
FRDE	l	inge in East dire e in East direction											
FRDS		inge in South direct											
FRDW	l	inge in West directi											
Text Area End	87F4	-	-	TAE<5>	TAE<4>	TAE<3>	TAE<2>	TAE<1>	TAE<0>	00H			
TAE<5:0>	Text Are	a End, in full ch	aracters							•			
Scroll Area	87F5	SSH<3>	SSH<2>	SSH<1>	SSH<0>	SSP<3>	SSP<2>	SSP<1>	SSP<0>	00H			
SSH<3:0>	Soft Scro	oll Height								•			
SSP<3:0>	Soft Scro	oll Position											
Scroll Range	87F6	SPS<3>	SPS<2>	SPS<1>	SPS<0>	STS<3>	STS<2>	STS<1>	STS<0>	00H			
SPS<3:0>	Stop Scro	oll row		•	•			•	•	•			
STS<3:0>	Start Scr	oll row											
RGB Brightness	87F7	VDSPOL	-	-	-	BRI<3>	BRI<2>	BRI<1>	BRI<0>	00H			

TDA935X/6X/8X PS/N2 series

VDSPOL		arity (1), Video (0) (0), Video (1)										
BRI<3:0>	RGB Bri	ghtness control										
Status read	87F8	BUSY	FIELD	SCON	FLR	SCR<3>	SCR<2>	SCR<1>	SCR<0>	00H		
BUSY	1		-	use display prob se display proble			1	1	1			
FIELD	0 - Odd I 1 - Even											
FLR	1	0 - Active flash region foreground and background displayed 1 - Active flash region background only displayed										
SCR<3:0>	First sere	First scroll row										
Status write	87F8	-	-	SCON	FLR	SCR<3>	SCR<2>	SCR<1>	SCR<0>	00H		
SCON		0 - Scroll area disabled 1 - Scroll area enabled										
FLR	0 - Active flash region foreground and background colour displayed 1 - Active flash region background colour only displayed											
SCR<3:0>	First Scroll Row											
H-Sync. delay	87FC	-	HSD<6>	HSD<5>	HSD<4>	HSD<3>	HSD<3>	HSD<1>	HSD<0>	00H		
HSD<6:0>	H-Sync o	lelay, in full size	characters	!	!				-!			
V-Sync Delay	87FD	-	VSD<6>	VSD<5>	VSD<4>	VSD<3>	VSD<2>	VSD<1>	VSD<0>	00H		
VSD<6:0>	V-Sync d	lelay in number	of TV lines									
Top Scroll Line	87FE	-	-	-	-	SCL<3>	SCL<2>	SCL<1>	SCL<0>	00H		
SCL<3:0>	Top line	for scroll			I.	ļ	1	1				
Configuration	87FF	CC	VDEL<2>	VDEL<1>	VDEL<0>	TXT/V	-	-	-	00H		
CC	0 - OSD 1 - Close	mode d Caption mode	;	l		1	1	1	1			
VDEL<2:0>	000 - VE 001 - VE 010 - VE	Pixel delay between VDS and RGB output 000 - VDS switched to video, not active 001 - VDS active one pixel earlier then RGB 010 - VDS synchronous to RGB 100 - VDS active one pixel after RGB										
TXT/V	BUSY Si 1 - Horiz 0 - Vertic											

Table 28 MMR Descriptions

TDA935X/6X/8X PS/N2 series

OTP MEMORY

These may be programmed either using the Parallel Programming Interface or via the ISP Programming Interface.

Parallel Programming

The following pins form the parallel programming interface:-

Pin	Name	Function
P0.5	IO(0)	Bit 0:- Address/Data/Mode
P0.6	IO(1)	Bit 1:- Address/Data/Mode
P1.0	IO(2)	Bit 2:- Address/Data/Mode
P1.1	IO(3)	Bit 3:- Address/Data/Mode
P1.2	IO(4)	Bit 4:- Address/Data/Mode
P1.3	IO(5)	Bit 5:- Address/Data/Mode
P3.1	IO(6)	Bit 6:- Address/Data/Mode
P3.2	IO(7)	Bit 7:- Address/Data/Mode
P2.0	OEB	Output Enable 0 = IO is output 1 = IO is input
P3.0	WEB	Write Enable, programming pulse >100us 0 = Program
P1.6	MODE	0 = IO(7:0) defined by A/DB 1 = IO(7:0) contains mode information
P1.7	A/DB	0 = IO(7:0) contains Data 1 = IO(7:0) contains Address Information
P3.3		Unused
VPE	VPE	9V Programming Voltage
RESET	RESET	Device reset/ mode selection
XTALIN	CLK	Clock 4 MHz

Table 29 Parallel Programming Interface

ISecurity Bits

The family of devices have a set of security bits for the combined OTP Program ROM, Character ROM and Packet 26 ROM. The security bits are used to prevent the ROM from being overwritten once programmed, and also the contents being verified once programmed. The

security bits are one-time programmable and CANNOT be erased.

TDA935X/6X/8X PS/N2 series

The memory and security bits are structured as follows:-

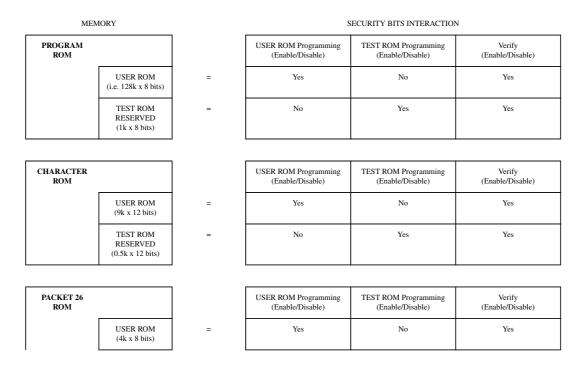


Table 30 Security bit structure

The security bits are set as follows for production programmed devices (i.e. programmed by Philips):-

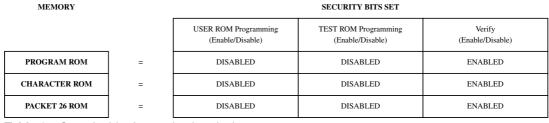


Table 31 Security bits for production devices

The security bits are set as follows for production un-programmed (blank) devices:-

MEMORY		SECURITY BITS SET							
		USER ROM Programming (Enable/Disable)	TEST ROM Programming (Enable/Disable)	Verify (Enable/Disable)					
PROGRAM ROM	=	ENABLED	DISABLED	ENABLED					
CHARACTER ROM	=	ENABLED	DISABLED	ENABLED					
PACKET 26 ROM	=	ENABLED	DISABLED	ENABLED					

Table 32 Security bits for Blank devices

TDA935X/6X/8X PS/N2 series

TDA935X/6X/8X PS/N2 series

FUNCTIONAL DESCRIPTION OF VIDEO PROCESSOR

Vision IF amplifier

The vision IF amplifier can demodulate signals with positive and negative modulation. The PLL demodulator is completely alignment-free.

The VCO of the PLL circuit is internal and the frequency is fixed to the required value by using the clock frequency of the $\mu\text{-Controller/Teletext}$ decoder as a reference. The setting of the various frequencies (38, 38.9, 45.75 and 58.75 MHz) can be made via the control bits IFA-IFC in subaddress 27H. Because of the internal VCO the IF circuit has a high immunity to EMC interferences.

QSS Sound circuit

The sound IF amplifier is similar to the vision IF amplifier and has an external AGC decoupling capacitor.

The single reference QSS mixer is realised by a multiplier. In this multiplier the SIF signal is converted to the intercarrier frequency by mixing it with the regenerated picture carrier from the VCO. The mixer output signal is supplied to the output via a high-pass filter for attenuation of the residual video signals. With this system a high performance hi-fi stereo sound processing can be achieved.

The AM sound demodulator is realised by a multiplier. The modulated sound IF signal is multiplied in phase with the limited SIF signal. The demodulator output signal is supplied to the output via a low-pass filter for attenuation of the carrier harmonics. The AM signal is supplied to the output (AUDOUT/AMOUT) via the volume control.

Switching between the QSS output and AM output is made by means of the AM bit in subaddress 29H (see also Table 1).

FM demodulator

The FM demodulator is realised as narrow-band PLL with external loop filter, which provides the necessary selectivity without using an external band-pass filter. To obtain a good selectivity a linear phase detector and a constant input signal amplitude are required. For this reason the intercarrier signal is internally supplied to the demodulator via a gain controlled amplifier and AGC circuit. To improve the selectivity an internal bandpass filter is connected in front of the PLL circuit.

The nominal frequency of the demodulator is tuned to the required frequency (4.5/5.5/6.0/6.5 MHz) by means of a calibration circuit which uses the clock frequency of the $\mu\text{-Controller/Teletext}$ decoder as a reference. The setting to the wanted frequency is realised by means of the control bits FMA/FMB in the control bit 29H.

From the output status bytes it can be read whether the PLL frequency is inside or outside the window and whether the PLL is in lock or not. With this information it is possible to make an automatic search system for the incoming sound frequency. This can be realised by means of a software loop which switches the demodulator to the various frequencies and then select the frequency on which a lock condition has been found.

The deemphasis output signal amplitude is independent of the TV standard and has the same value for a frequency deviation of ± 25 kHz at the 4.5 MHz standard and for a deviation of ± 50 Khz for the other standards.

TDA935X/6X/8X PS/N2 series

Audio circuit and input signal selection

The audio control circuit contains an audio switch with 1 external input and a volume control circuit. The selection of the various inputs is made by means of the ADX bit. In various versions the Automatic Volume Levelling (AVL) function can be activated. The pin to which the external capacitor has to be connected depends on the IC version. For the 90° types the capacitor is connected to the EW output pin. For the 110° types a choice must be made between the AVL function and a sub-carrier output for comb filter applications. This choice is made via the CBM0/1 bits (in subaddress 22H). When the AVL is active it automatically stabilises the audio output signal to a certain level.

It is possible to use the deemphasis pin as additional audio input. In that case the internal signal must, of course, be switched off. This can be realised by means of the sound mute bit (SM in subaddress 29H). When the IF circuit is switched to positive modulation the internal signal on the deemphasis pin is automatically muted.

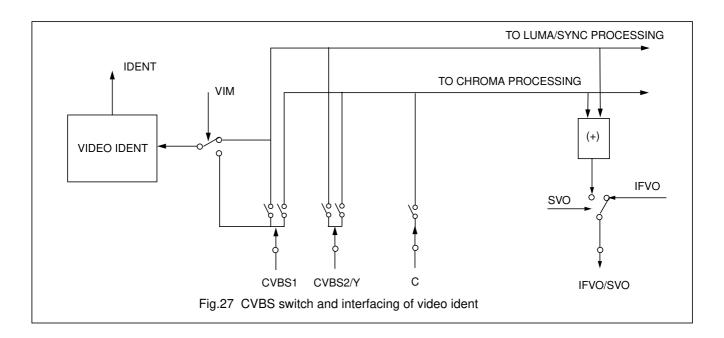
CVBS and Y/C input signal selection

The circuit has 2 inputs for external CVBS signals and one input can also be used as one Y/C input (see Fig. 27).

It is possible to supply the selected CVBS signal to the demodulated IF video output pin. This mode is selected by means of the SVO bit in subaddress 22H. The vision IF amplifier is switched off in this mode.

The video ident circuit can be connected to the incoming 'internal' video signal or to the selected signal. This ident circuit is independent of the synchronisation and can be used to switch the time-constant of the horizontal PLL depending on the presence of a video signal (via the VID bit). In this way a very stable OSD can be realised.

The subcarrier output is combined with a 3-level output switch (0 V, 2.3 V and 4.5 V). The output level and the availability of the subcarrier signal is controlled by the CMB1 and CMB0 bits. The output can be used to switch sound traps etc. It is also possible to use this pin for the connection of the AVL capacitor or as AM output.



TDA935X/6X/8X PS/N2 series

Synchronisation circuit

The IC contains separator circuits for the horizontal and vertical sync pulses and a data-slicing circuit which extracts the digital teletext data from the analog signal.

The horizontal drive signal is obtained from an internal VCO which is running at a frequency of 25 MHz. This oscillator is stabilised to this frequency by using a 12 MHz signal coming from the reference oscillator of the $\mu\text{-Controller}.$

The horizontal drive is switched on and off via the soft start/stop procedure. This function is realised by means of variation of the T_{ON} of the horizontal drive pulses. In addition the horizontal drive circuit has a 'low-power start-up' function.

The vertical synchronisation is realised by means of a divider circuit. The vertical ramp generator needs an external resistor and capacitor. For the vertical drive a differential output current is available. The outputs must be DC coupled to the vertical output stage.

In the types which are intended for 90° picture tubes the following geometry parameters can be adjusted:

- · Horizontal shift
- · Vertical amplitude
- · Vertical slope
- S-correction
- · Vertical shift

The types which are intended to be used in combination with 110° picture tubes have an East-West control circuit in stead of the AVL function. The additional controls for these types are:

- EW width
- EW parabola width
- EW upper and lower corner parabola correction
- · EW trapezium correction
- Vertical zoom
- horizontal parallelogram and bow correction.

When the vertical amplitude is compressed (zoom factor <1) it is still possible to display the black current measuring lines in the overscan. This function is activated by means of the bit OSVE in subaddress 26H.

Chroma, luminance and feature processing

The chroma band-pass and trap circuits (including the SECAM cloche filter) are realised by means of gyrators and are tuned to the right frequency by comparing the tuning frequency with the reference frequency of the colour decoder. The luminance delay line and the delay cells for the peaking circuit are also realised with gyrators.

The circuit contains the following picture improvement features:

- Peaking control circuit. The ratio of the positive and negative overshoots of the peaking can be adjusted by means of the bits RPO1/RPO0 in subaddress 2EH.
- Black stretch. This function corrects the black level for incoming signals which have a difference between the black level and the blanking level.

TDA935X/6X/8X PS/N2 series

Colour decoder

The ICs can decode PAL, SECAM and NTSC signals. The PAL/NTSC decoder does not need external reference crystals but has an internal clock generator which is stabilised to the required frequency by using the 12 MHz clock signal from the reference oscillator of the $\mu\text{-Controller}.$

Under bad-signal conditions (e.g. VCR-playback in feature mode), it may occur that the colour killer is activated although the colour PLL is still in lock. When this killing action is not wanted it is possible to overrule the colour killer by forcing the colour decoder to the required standard and to activate the FCO-bit (Forced Colour On) in subaddress 21H.

The Automatic Colour Limiting (ACL) circuit (switchable via the ACL bit in subaddress 20H) prevents that oversaturation occurs when PAL/NTSC signals with a high chroma-to-burst ratio are received. The ACL circuit is designed such that it only reduces the chroma signal and not the burst signal. This has the advantage that the colour sensitivity is not affected by this function.

The SECAM decoder contains an auto-calibrating PLL demodulator which has two references, viz: the divided 12 MHz reference frequency (obtained from the $\mu\text{-Controller})$ which is used to tune the PLL to the desired free-running frequency and the bandgap reference to obtain the correct absolute value of the output signal. The VCO of the PLL is calibrated during each vertical blanking period, when the IC is in search or SECAM mode.

The base-band delay line (TDA 4665 function) is integrated. This delay line is also active during NTSC to obtain a good suppression of cross colour effects. The demodulated colour difference signals are internally supplied to the delay line.

RGB output circuit and black-current stabilization

In the RGB control circuit the signal is controlled on contrast, brightness and saturation. The ICs have a linear input for external RGB/YUV signals. Switching between RGB and the YUV mode can be realised via the YUV bit in subaddress 2BH. The signals for OSD and text are internally supplied to the control circuit. The output signal has an amplitude of about 2 V black-to-white at nominal input signals and nominal settings of the various controls.

To obtain an accurate biasing of the picture tube the 'Continuous Cathode Calibration' (CCC) system has been included in these ICs. When required the operation of the CCC system can be changed into a one-point black current system. The switching between the 2 possibilities

is realised by means of the OPC bit in subaddress 2BH. When used as one-point control loop the system will control the black level of the RGB output signals to the 'low' reference current and not on the cut off point of the cathode. In this way spreads in the picture tube characteristics will no take into account. A further consequence is that the RGB output signals have a fixed amplitude (2 V_{P-P} under nominal conditions) and that the 'cathode drive level' bits (CL3-CL0) have no effect on these amplitudes. For this reason the gain of the RGB output stages has to be adapted to the required drive level of the cathodes.

A black level off-set can be made with respect to the level which is generated by the black current stabilization system. In this way different colour temperatures can be obtained for the bright and the dark part of the picture. In the V_{g2} adjustment mode (AVG = 1) the black current stabilization system checks the output level of the 3 channels and indicates whether the black level of the highest output is in a certain window (WBC-bit) or below or above this window (HBC-bit). This indication can be read from the status byte 01 and can be used for automatic adjustment of the V_{g2} voltage during the production of the TV receiver. During this test the vertical scan remains active so that the indication of the 2 bits can be made visible on the TV screen.

The control circuit contains a beam current limiting circuit and a peak white limiting circuit. To prevent that the peak white limiting circuit reacts on the high frequency content of the video signal a low-pass filter is inserted in front of the peak detector.

During switch-off of the TV receiver a fixed beam current is generated by the black current control circuit. This current ensures that the picture tube capacitance is discharged. During the switch-off period the vertical deflection can be placed in an overscan position so that the discharge is not visible on the screen.

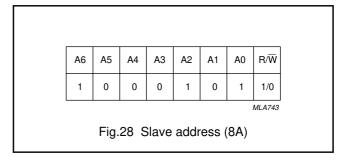
A wide blanking pulse can be activated in the RGB outputs by means of the HBL bit in subaddress 2BH. The timing of this blanking can be adjusted by means of the bits WBF/R bits in subaddress 03H.

TDA935X/6X/8X PS/N2 series

SOFTWARE CONTROL

The CPU communicates with the peripheral functions using Special function Registers (SFRs) which are addressed as RAM locations. The registers for the Teletext decoder appear as normal SFRs in the $\mu\text{-Controller}$ memory map and are written to these functions by using a serial bus. This bus is controlled by dedicated hardware which uses a simple handshake system for software synchronisation.

For compatibility reasons and possible re-use of software blocks, the I²C-bus control for the TV processor is organised as in the stand-alone TV signal processors. The TV processor registers cannot be read, so when the content of these registers is needed in the software, a copy should be stored in Auxiliary RAM or Non Volatile RAM. The slave address of the TV signal processor is given in Fig.28.



Valid subaddresses: 05H to 2EH, subaddress FE and FF are reserved for test purposes. Auto-increment mode available for subaddresses.

TDA935X/6X/8X PS/N2 series

DESCRIPTION OF THE I2C-BUS SUBADDRESSES

Table 33 Inputs TV-processor

FUNCTION	SUBADDR				DATA	BYTE				POR
FUNCTION	(HEX)	D7	D6	D5	D4	D3	D2	D1	D0	Value
Off-set IF demodulator	05	0	0	A5	A4	A3	A2	A1	A0	20
Horizontal parallelogram (1)	06	0	0	A5	A4	A3	A2	A1	A0	20
Horizontal bow (1)	07	0	0	A5	A4	A3	A2	A1	A0	20
Hue	08	0	0	A5	A4	A3	A2	A1	A0	00
Horizontal shift (HS)	09	0	0	A5	A4	A3	A2	A1	A0	20
EW width (EW) (1)	0A	0	0	A5	A4	A3	A2	A1	A0	20
EW parabola/width (PW) (1)	0B	0	0	A5	A4	A3	A2	A1	A0	20
EW upper corner parabola(1)	0C	0	0	A5	A4	A3	A2	A1	A0	20
EW lower corner parabola(1)	0D	0	0	A5	A4	A3	A2	A1	A0	20
EW trapezium (TC) (1)	0E	0	0	A5	A4	A3	A2	A1	A0	20
Vertical slope (VS)	0F	0	0	A5	A4	A3	A2	A1	A0	20
Vertical amplitude (VA)	10	0	0	A5	A4	A3	A2	A1	A0	20
S-correction (SC)	11	0	0	A5	A4	A3	A2	A1	A0	20
Vertical shift (VSH)	12	0	0	A5	A4	A3	A2	A1	A0	20
Vertical zoom (VX) (1)	13	0	0	A5	A4	A3	A2	A1	A0	20
Black level offset R	14	0	0	A5	A4	A3	A2	A1	A0	20
Black level offset G	15	0	0	A5	A4	A3	A2	A1	A0	20
White point R	16	0	0	A5	A4	A3	A2	A1	A0	20
White point G	17	0	0	A5	A4	A3	A2	A1	A0	20
White point B	18	0	0	A5	A4	A3	A2	A1	A0	20
Peaking	19	PF1	PF0	A5	A4	A3	A2	A1	A0	20
Luminance delay time	1A	0	0	0	0	YD3	YD2	YD1	YD0	00
Brightness	1B	0	0	A5	A4	A3	A2	A1	A0	20
Saturation	1C	0	0	A5	A4	A3	A2	A1	A0	20
Contrast	1D	0	0	A5	A4	A3	A2	A1	A0	20
AGC take-over	1E	0	0	A5	A4	A3	A2	A1	A0	20
Volume control	1F	0	0	A5	A4	A3	A2	A1	A0	20
Colour decoder 0	20	CM3	CM2	CM1	CM0	MAT	MUS	ACL	CB	00
Colour decoder 1	21	0	0	0	0	0	0	BPS	FCO	00
AV-switch 0	22		0	SVO	CMB1	CMB0	INA	INB	0	00
AV-switch 1	23	0	0	0	0	0	0	0	RGBL	00
Synchronisation 0	24	0	HP2	FOA	FOB	POC	STB	VIM	VID	00
Synchronisation 1	25	0	0	FSL	OSO	FORF	FORS	DL	NCIN	00
Deflection	26	0	AFN	DFL	XDT	SBL	AVG	EVG	HCO ⁽¹⁾	00
Vision IF 0	27	IFA	IFB	IFC	VSW	MOD	AFW	IFS	STM	00
Vision IF 1	28	SIF	0	0	IFLH	0	AGC1	AGC0	FFI	00
Sound 0	29	AGN	SM1	FMWS	AM ⁽³⁾	SM0	0	FMB	FMA	00
Control 0	2A	0	IE2	RBL	AKB	CL3	CL2	CL1	CL0	00
Control 1	2B	0	IVG	0	0	0	YUV	0	HBL ⁽¹⁾	00
Sound 1	2C	0	0	ADX	0	0	AVL ⁽²⁾	0	0	00
Features 0	2D	0	0	0	0	0	0	0	BKS	00
Features 1	2E	0	0	RPO1	RPO0	0	0	0	0	00

Note

- 1. These functions are only available in versions which have the East-West drive output.
- 2. The AVL function is only available in versions which have no East-West output or when the subcarrier output is used for the connection of the AVL capacitor (via the bits CMB1 and CMB0 in subaddress 22H).
- 3. Only available in types with QSS sound IF circuit and AM demodulator.

TDA935X/6X/8X PS/N2 series

Table 34 Outputs TV-processor

FUNCTION	SUBADDR	DATA BYTE									
FUNCTION		D7	D6	D5	D4	D3	D2	D1	D0		
Output status bytes	00	POR	IFI	LOCK	SL	CD3	CD2	CD1	CD0		
	01	XPR	NDF	FSI	IVW	WBC	HBC	BCF	IN2		
	02	SUP	Х	Х	QSS	AFA	AFB	FMW	FML		
	03	ID7	ID6	ID5	ID4	ID3	ID2	ID1	ID0		
	04	SN1	SN0	X	X	Х	Х	Х	X		

Explanation input control data TV-processor

Table 35 Off-set IF demodulator

DAC SETTING	CONTROL
0	negative correction
20	no correction
3F	positive correction

Table 36 Horizontal parallelogram

DAC SETTING	CONTROL
0	screen top 0.5 μs delayed and screen bottom 0.5 μs advanced with respect to centre
20	no correction
3F	screen top 0.5 μs advanced and screen bottom 0.5 μs delayed with respect to centre

Table 37 Horizontal bow

DAC SETTING	CONTROL
0	screen top and bottom 0.5 μs delayed with respect to centre
20	no correction
3F	screen top and bottom 0.5 μs advanced with respect to centre

Table 38 Hue control

DAC SETTING	CONTROL
0	-45°
20	0°
3F	+45°

Table 39 Horizontal shift

DAC SETTING	CONTROL
0	–2 μs
20	0
3F	+2 μs

Table 40 EW width

DAC SETTING	CONTROL
0	output current 700 μA
3F	output current 0 μA

Table 41 EW parabola/width

DAC SETTING	CONTROL
0	output current 0 μA
3F	output current 440 µA at top and bottom of screen

Table 42 EW upper/lower corner parabola

DAC SETTING	CONTROL
0	output current +76 μA
11	output current 0 μA
3F	output current –207 μA

Table 43 EW trapezium

DAC SETTING	CONTROL
0	output current at top of screen 100 μA lower that at bottom
20	no correction
3F	output current at top of screen 100 μA higher than at bottom

TDA935X/6X/8X PS/N2 series

Table 44 Vertical slope

DAC SETTING	CONTROL
0	correction –20%
20	no correction
3F	correction +20%

Table 45 Vertical amplitude

DAC SETTING	CONTROL
0	amplitude 80%
20	amplitude 100%
3F	amplitude 120%

Table 46 S-correction

DAC SETTING	CONTROL
0	correction –10%
0E	no correction
3F	correction 25%

Table 47 Vertical shift

DAC SETTING	CONTROL
0	shift -5%
20	no correction
3F	shift +5%

Table 48 Vertical zoom

DAC SETTING	CONTROL
0	amplitude 75%
20	amplitude 100%
3F	amplitude 138%

Table 49 Black level off-set R/G

DAC SETTING	CONTROL
0	off-set of –160 mV
20	no off-set
3F	off-set of +160 mV

Table 50 White point R/G/B

DAC SETTING	CONTROL
0	gain –3 dB
20	no correction
3F	gain +3 dB

Table 51 Peaking centre frequency

PF1	PF0	CENTRE FREQUENCY
0	0	2.7 MHz
0	1	3.1 MHz
1	0	3.5 MHz
1	1	spare

Table 52 Peaking control (overshoot in direction 'black')

DAC SETTING	CONTROL
0	depeaking (overshoot -22%)
10	no peaking
3F	overshoot 80%

Table 53 Y-delay adjustment; note 1

YD0 to YD3	Y-DELAY
YD3	YD3 × 160 ns +
YD2	YD2 × 80 ns +
YD1	YD1 × 80 ns +
YD0	YD0 × 40 ns

Note

 For an equal delay of the luminance and chrominance signal the delay must be set at a value of 160 ns. This is only valid for a CVBS signal without group delay distortions.

Table 54 Brightness control

DAC SETTING	CONTROL
0	correction –0.7V
20	no correction
3F	correction +0.7V

Table 55 Saturation control

DAC SETTING	CONTROL
0	colour off (-52 dB)
17	saturation nominal
3F	saturation +300%

Table 56 Contrast control

DAC SETTING	CONTROL
0	RGB amplitude –14 dB
20	RGB amplitude nominal
3F	RGB amplitude +6 dB

TDA935X/6X/8X PS/N2 series

Table 57 AGC take-over

DAC SETTING	CONTROL
0	tuner take-over at IF input signal of 0.4 mV
3F	tuner take-over at IF input signal of 80 mV

Table 58 Volume control

DAC SETTING	CONTROL
0	attenuation 80 dB
3F	no attenuation

Table 59 Colour decoder mode, note 1

СМЗ	CM2	CM1	СМО	DECODER MODE	FREQ
0	0	0	0	PAL/NTSC/SECAM	Α
0	0	0	1	PAL/SECAM	Α
0	0	1	0	PAL	Α
0	0	1	1	NTSC	Α
0	1	0	0	SECAM	
0	1	0	1	PAL/NTSC	В
0	1	1	0	PAL	В
0	1	1	1	NTSC	В
1	0	0	0	PAL/NTSC/SECAM	ABCD
1	0	0	1	PAL/NTSC	С
1	0	1	0	PAL	С
1	0	1	1	NTSC	С
1	1	0	0	PAL/NTSC	BCD
				(Tri-Norma)	
1	1	0	1	PAL/NTSC	D
1	1	1	0	PAL	D
1	1	1	1	NTSC	D

Note

1. The decoder frequencies for the various standards are obtained from an internal clock generator which is synchronised by a 12 MHz reference signal which is obtained from the μ -Controller clock generator.

These frequencies are:

a) A: 4.433619 MHz

b) B: 3.582056 MHz (PAL-N)

c) C: 3.575611 MHz (PAL-M)

d) D: 3.579545 MHz (NTSC-M)

Table 60 PAL-SECAM/NTSC matrix

MAT	MATRIX POSITION
0	adapted to standard
1	PAL matrix

Table 61 NTSC matrix

MUS	MATRIX POSITION
0	Japanese matrix
1	USA matrix

Table 62 Automatic colour limiting

ACL	COLOUR LIMITING
0	not active
1	active

Table 63 Chroma bandpass centre frequency

СВ	CENTRE FREQUENCY
0	F _{SC}
1	1.1 × F _{SC}

Table 64 Bypass of chroma base-band delay line

BPS	DELAY LINE MODE
0	active
1	bypassed

Table 65 Forced Colour-On

FCO		CONDITION
0	off	
1	on	

Table 66 Selected video out

svo	CONDITION
0	IF video available at output
1	selected CVBS available at output

Table 67 Condition AVL/SNDIF/REFO

CMB1	СМВ0	CONDITION	
0	0	AVL/SNDIF active (depends on SIF bit)	
0	1	output voltage 2.3 V + subcarrier;	
1	0	output voltage low (<0.8 V)	
1	1	output voltage high (>4.5V)	

TDA935X/6X/8X PS/N2 series

Table 68 Source select

INA	INB	INC	SELECTED SIGNALS
0	0	0	CVBS1
0	1	0	CVBS2
1	0	0	Y/C

Table 69 Blanking of RGB outputs

RGBL	CONDITION
0	normal operation
1	RGB outputs blanked continuously

Table 70 Synchronization of OSD/TEXT display

HP2	μ-CONTROLLER COUPLED TO
0	φ1 loop
1	φ2 loop

Table 71 Phase 1 (ϕ_1) time constant

FOA	FOB	MODE
0	0	normal
0	1	slow
1	0	slow/fast
1	1	fast

Table 72 Synchronization mode

POC	MODE
0	active
1	not active

Table 73 Stand-by

STB	MODE
0	stand-by
1	normal

Table 74 Video ident mode

VIM	MODE
0	ident coupled to internal CVBS (pin 40)
1	ident coupled to selected CVBS

Table 75 Video ident mode

VID	VIDEO IDENT MODE
0	φ1 loop switched on and off
1	not active

Table 76 Forced slicing level for vertical sync

FSL	SLICING LEVEL
0	slicing level dependent on noise detector
1	fixed slicing level of 60%

Table 77 Switch-off in vertical overscan

oso	MODE
0	Switch-off undefined
1	Switch-off in vertical overscan

Table 78 Forced field frequency

FORF	FORS	FIELD FREQUENCY
0	0	auto (60 Hz when line not in sync)
0	1	60 Hz
1	0	keep last detected field frequency
1	1	auto (50 Hz when line not in sync)

Table 79 Interlace

DL	STATUS
0	interlace
1	de-interlace

Table 80 Vertical divider mode

NCIN	VERTICAL DIVIDER MODE
0	normal operation
1	switched to search window

Table 81 AFC switch

AFN	MODE
0	normal operation
1	AFC not active

Table 82 Disable flash protection

DFL	MODE
0	flash protection active
1	flash protection disabled

TDA935X/6X/8X PS/N2 series

Table 83 X-ray detection

XDT	MODE
0	protection mode, when a too high EHT is detected the receiver is switched to stand-by and the XPR-bit is set to 1
1	detection mode, the receiver is not switched to stand-by and only the XPR-bit is set to 1

Table 84 Service blanking

SBL	SERVICE BLANKING MODE
0	off
1	on

Table 85 Adjustment V_{g2} voltage

AVG	MODE	
0	normal operation	
1	V_{g2} adjustment (WBC and HBC bits in output byte 01 can be read)	

Table 86 Enable vertical guard (RGB blanking)

EVG	VERTICAL GUARD MODE	
0	not active	
1	active	

Table 87 EHT tracking mode

нсо	TRACKING MODE	
0	EHT tracking only on vertical	
1	EHT tracking on vertical and EW	

Table 88 PLL demodulator frequency adjust

IFA	IFB	IFC	IF FREQUENCY
0	0	0	58.75 MHz
0	0	1	45.75 MHz
0	1	0	38.90 MHz
0	1	1	38.00 MHz
1	0	0	33.40 MHz
1	1	0	33.90 MHz

Table 89 Video mute

vsw	STATE	
0	normal operation	
1	IF-video signal switched off	

Table 90 Modulation standard

MOD	MODULATION	
0	negative	
1	positive	

Table 91 AFC window

AFW	AFC WINDOW	
0	normal	
1	enlarged	

Table 92 IF sensitivity

IFS	IF SENSITIVITY	
0	normal	
1	reduced	

Table 93 Search tuning mode

STM	MODE	
0	normal operation	
1	reduced sensitivity of video indent circuit	

Table 94 Selection external input for sound IF circuit

SIF	MODE	
0	IF input not selected	
1	IF input selected (see also table 1)	

Table 95 Calibration of IF PLL demodulator

IFLH	MODE			
0	calibration system active			
1	calibration system not active			

Table 96 IF AGC speed

AGC1	AGC0	AGC SPEED			
0	0	0.7 × norm			
0	1	norm			
1	0	3 × norm			
1	1	6×norm			

Table 97 Fast filter IF-PLL

FFI	CONDITION		
0	normal time constant		
1	increased time constant		

TDA935X/6X/8X PS/N2 series

Table 98 Gain FM demodulator

AGN	MODE		
0	normal operation		
1	gain +6 dB, to be used for the demodulation of mono signals in the NTSC system		

Table 99 Sound mute

SM1	SM0	CONDITION		
0	1	see note 1		
1	0	mute on		
1	1	mute off		

Note

1. The mute is activated when the digital acquisition help is out-of-window.

Table 100Window selection of Narrow-band sound PLL

FMWS	FUNCTION		
0	small window		
1	large window		

Table 101 Selection QSS out or AM out

АМ	MODE			
0	QSS output selected			
1	AM output selected			

Table 102Nominal frequency FM demodulator

FMB	FMA	FREQUENCY		
0	0	5.5 MHz		
0	1	6.0 MHz		
1	0	4.5 MHz		
1	1	6.5 MHz		

Table 103Enable fast blanking ext.RGB/YUV

IE2	FAST BLANKING			
0	not active			
1	active			

Table 104 RGB blanking

RBL	RGB BLANKING		
0	not active		
1	active		

Table 105 Black current stabilization

AKB	MODE		
0	active		
1	not active		

Table 106 Cathode drive level (15 steps; 3.5 V/step)

CL3	CL2	CL1	CL0	SETTING CATHODE DRIVE AMPLITUDE; NOTE 1
0	0	0	0	50 V _{BL-WH}
0	1	1	1	75 V _{BL-WH}
1	1	1	1	95 V _{BL-WH}

Note

- 1. The given values are valid for the following conditions:
 - a) Nominal CVBS input signal
 - b) Nominal settings for contrast, WPA and peaking
 - c) Black- and blue-stretch switched-off
 - d) Gain of output stage such that no clipping occurs
 - e) Beam current limiting not active
 - f) The tolerance on these values is about \pm 3 V.

Table 107Activation vertical guard

IVG	MODE
0	vertical guard connected to black current input
1	vertical guard connected to BCL input and "Peak-White-Limiting" disabled

Table 108 RGB / YUV switch

YUV	STATUS
0	RGB input activated
1	YUV input activated

Table 109RGB blanking mode (110° types)

HBL	MODE
0	normal blanking (horizontal flyback)
1	wide blanking

TDA935X/6X/8X PS/N2 series

Table 110Audio signal selection

ADX	SELECTED SIGNAL		
0	internal audio signal		
1	external audio signal		

Table 111 Auto Volume Levelling

AVL	MODE
0	not active
1	active

Table 112Black stretch

BKS		BLACK STRETCH MODE
0	off	
1	on	

Table 113 Ratio pre- and overshoot

RPO1	RPO0	RATIO PRE-/OVERSHOOT
0	0	1:1
0	1	1:1.25
1	0	1:1.5
1	1	1:1.8

Explanation output control data TV-processor

Table 114 Power-on-reset

POR	MODE
0	normal
1	power-down

Table 115 Output video identification

IFI	VIDEO SIGNAL		
0	no video signal identified		
1	video signal identified		

Table 116 IF-PLL lock indication

LOCK	INDICATION
0	not locked
1	locked

Table 117 Phase 1 (ϕ_1) lock indication

SL	INDICATION
0	not locked
1	locked

Table 118 Colour decoder mode, note 1

CD2	CD3 CD2 CD1 CD0 STANDARD			
CD3	CDZ	CDI	CDU	STANDARD
0	0	0	0	no colour standard identified
0	0	0	1	NTSC with freq. A
0	0	1	0	PAL with freq. A
0	0	1	1	NTSC with freq. B
0	1	0	0	PAL with freq. B
0	1	0	1	NTSC with freq. C
0	1	1	0	PAL with freq. C
0	1	1	1	NTSC with freq. D
1	0	0	0	PAL with freq. D
1	0	1	0	SECAM

Note

1. The values for the various frequencies can be found in the note of table 59.

Table 119 X-ray protection

XPR	OVERVOLTAGE		
0	no overvoltage detected		
1	overvoltage detected		

TDA935X/6X/8X PS/N2 series

Table 120 Output vertical guard

NDF	VERTICAL OUTPUT STAGE
0	OK
1	failure

Table 121 Field frequency indication

FSI		FREQUENCY
0	50 Hz	
1	60 Hz	

Table 122 Condition vertical divider

IVW	STANDARD VIDEO SIGNAL		
0	no standard video signal		
1	standard video signal (525 or 625 lines)		

Table 123 Indication output black level in/out window

	WBC	CONDITION		
	0	black current stabilisation outside window		
black current stabilisation inside window		black current stabilisation inside window		

Table 124 Indication output black level

нвс	CONDITION		
0	black current stabilisation below window		
1	black current stabilisation above window		

Table 125 Condition black current loop

BCF	CONDITION		
0	black current loop is stabilised		
1	black current loop is not stabilised		

Table 126 Indication RGB-2 input condition

IN2	RGB INSERTION	
0	no	
1	yes	

Table 127 Supply voltage indication

SUP	CONDITION	
0	supply voltage (8 Volt) not present	
1	supply voltage (8 Volt) present	

Table 128 Version indication

QSS	IC VERSION		
0	version with intercarrier mono sound circuit		
1	version with QSS-IF circuit		

Table 129 AFC output

AFA	AFB	CONDITION	
0	0	outside window; RF too low	
0	1	outside window; RF too high	
1	0	in window; below reference	
1	1	in window; above reference	

Table 130 Indication FM-PLL in/out window

FMW	CONDITION			
0	FM-PLL in window			
1	FM-PLL out of window			

Table 131Indication FM-PLL in/out lock

FML	CONDITION		
0	FM-PLL out of lock		
1	FM-PLL locked		

TDA935X/6X/8X PS/N2 series

LIMITING VALUES

In accordance with the Absolute Maximum Rating System (IEC 134).

SYMBOL	PARAMETER	CONDITIONS	MIN.	MAX.	UNIT
V _P	supply voltage		_	9	٧
V_{DD}	supply voltage (all digital supplies)		-0.5	5.0	V
V _I	digital inputs	note 1	-0.5	V _{DD} + 0.5	V
Vo	digital outputs	note 1	-0.5	V _{DD} + 0.5	٧
Io	output current (each output)		_	±10	mA
I _{IOK}	DC input or output diode current		_	±20	mA
T _{stg}	storage temperature		-25	+150	°C
T _{amb}	operating ambient temperature		0	70	°C
T _{sol}	soldering temperature	for 5 s	_	260	°C
T _j	operating junction temperature		_	150	°C
V _{es}	electrostatic handling	HBM; all pins; notes 2 and 3	-2000	+2000	٧
		MM; all pins; notes 2 and 4	-300	+300	V

Notes

- 1. This maximum value has an absolute maximum of $5.5\ V$ independent of V_{DD} .
- 2. All pins are protected against ESD by means of internal clamping diodes.
- 3. Human Body Model (HBM): $R = 1.5 \text{ k}\Omega$; C = 100 pF.
- 4. Machine Model (MM): $R = 0 \Omega$; C = 200 pF.

THERMAL CHARACTERISTICS

SYMBOL	PARAMETER	VALUE	UNIT
R _{th j-a}	thermal resistance from junction to ambient in free air	35	K/W

QUALITY SPECIFICATION

In accordance with "SNW-FQ-611E".

Latch-up

At an ambient temperature of 70 °C all pins meet the following specification:

- $I_{trigger} \ge 100 \text{ mA or } \ge 1.5 V_{DD(max)}$
- $I_{trigger} \le -100 \text{ mA or } \le -0.5 V_{DD(max)}$.

TDA935X/6X/8X PS/N2 series

CHARACTERISTICS OF MICRO-COMPUTER AND TEXT DECODER

 V_{DD} = 3.3 V \pm 10%; V_{SS} = 0 V; T_{amb} = –20 to +70 °C; unless otherwise specified

NUMBER	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
Supplies			'	•	-1	
VM.1.1	supply voltage (V _{DDA/P/C})		3.0	3.3	3.6	V
VM.1.2	periphery supply current (I _{DDP})	note 1	1	-	_	mA
VM.1.3	core supply current (I _{DDC})		_	15	tbf	mA
VM.1.4	analog supply current (I _{DDA})		_	45	tbf	mA
Digital input	ts					
RESET						
1.1.1	low level input voltage		_	_	0.8	V
I.1.2	high level input voltage		2.0	_	5.5	V
I.1.3	hysteresis of Schmitt Trigger input		0.4	_	0.7	V
1.1.4	input leakage current	$V_I = 0$	_	-	1	μΑ
I.1.5	equivalent pull down resistance	$V = V_{DD}$	_	33	_	kΩ
I.1.6	capacitance of input pin		_	_	10	pF
Digital input	t/outputs					
P1.0 TO P1.3	3, P2.0 TO P2.6 AND P3.0 TO P3.3					
IO.1.1	low level input voltage		_	-	0.8	V
IO.1.2	high level input voltage		2.0	-	5.5	V
IO.1.3	hysteresis of Schmitt Trigger input		0.4	_	0.7	V
IO.1.4	low level output voltage	I _{OL} = 4 mA	_	-	0.4	V
IO.1.5	high level output voltage	open drain	_	-	5.5	V
IO.1.6	high level output voltage	I _{OH} = 4 mA	2.4	_	_	V
IO.1.7	output rise time (push-pull only) 10% to 90%	load 100 pF	_	16	_	ns
IO.1.8	output fall time 10% to 90%	load 100pF	_	14	_	ns
IO.1.9	load capacitance		_	_	100	pF
IO.1.10	capacitance of input pin		_	_	10	pF

TDA935X/6X/8X PS/N2 series

NUMBER	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
P0.5 AND P0	.6		'	•	-1	
IO.2.1	low level input voltage		_	_	0.8	V
IO.2.2	high level input voltage		2.0	_	5.5	V
IO.2.3	hysteresis of Schmitt Trigger input		0.4	_	0.7	V
IO.2.4	low level output voltage	I _{OL} = 8mA	_	_	0.4	V
IO.2.5	high level output voltage	open drain	_	_	5.5	V
IO.2.6	high level output voltage	I _{OH} = 8mA	2.4	_	_	V
IO.2.7	output rise time (push-pull only) 10% to 90%	load 100 pF	-	16	_	ns
IO.2.8	output fall time 10% to 90%	load 100pF	_	14	_	ns
IO.2.9	load capacitance		_	_	100	pF
IO.2.10	capacitance of input pin		_	_	10	pF
P1.6 AND P1	.7			•		•
IO.3.1	low level input voltage (V _{IL})		_	_	1.5	V
IO.3.2	high level input voltage (VIH)		3.0	_	5.5	V
IO.3.3	hysteresis of Schmitt-trigger input		0.2	_	_	V
IO.3.4	low level output voltage	sink current 8mA	0	_	0.4	V
IO.3.5	high level output voltage	open drain	_	_	5.5	V
IO.3.6	output fall time (V _{IH} to V _{IL} for C _L)		20+0.1× C _L	_	250	ns
IO.3.7	bus load capacitance		10	_	400	pF
IO.3.8	capacitance of IO pin		_	_	10	pF
Crystal osci	illator		,	•		
OSCIN; NOTE	2					
X.1.1	resonator frequency		_	12	_	MHz
X.1.2	input capacitance (C _i)		_	4.0	_	pF
X.1.3	output capacitance (Co)		-	5.0	_	pF
X.1.4	$C_{x1} = C_{x2}$		12	_	56	pF
X.1.5	R _i (crystal)		_	Ī-	100	Ω

Note

- 1. Peripheral current is dependent on external components and voltage levels on I/Os
- 2. The simplified circuit diagram of the oscillator is given in Fig.29.

A suitable crystal for this oscillator is the Saronix type 9922 520 00169. The nominal tuning of the crystal is important to obtain a symmetrical catching range for the PLL in the colour decoder. This tuning can be adapted by means of the values of the capacitors C_{x1} and C_{x2} in Fig.29. Good results were obtained with capacitor values of 39 pF, however, for a new application the optimum value should be determined by checking the symmetry of the catching range of the colour decoder.

I

TV signal processor-Teletext decoder with embedded $\mu\text{-}Controller$

TDA935X/6X/8X PS/N2 series

CHARACTERISTICS OF TV-PROCESSORS

 $V_P = 5 \text{ V}$; $T_{amb} = 25 \,^{\circ}\text{C}$; unless otherwise specified.

NUMBER	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
Supplies	-			ļ.	1	
MAIN SUPPLY;	NOTE 1					
V.1.1	supply voltage		7.2	8.0	8.4	V
V.1.2	total supply current		_	135	_	mA
V.1.4	total power dissipation		-	1085	-	mW
IF circuit			•	•	•	
VISION IF AME	PLIFIER INPUTS					
	input sensitivity (RMS value)	note 2				
M.1.1		f _i = 38.90 MHz	_	75	150	μV
M.1.2		f _i = 45.75 MHz	_	75	150	μV
M.1.3		f _i = 58.75 MHz	_	75	150	μV
M.1.4	input resistance (differential)	note 3	_	2	_	kΩ
M.1.5	input capacitance (differential)	note 3	_	3	_	pF
M.1.6	gain control range		64	_	_	dB
M.1.7	maximum input signal (RMS value)		150	-	_	mV
PLL DEMODUL	LATOR; NOTES 4 AND 5		•			•
M.2.1	Free-running frequency of VCO	PLL not locked, deviation from nominal setting	-500	-	+500	kHz
M.2.2	Catching range PLL	without SAW filter	_	±1	_	MHz
M.2.3	delay time of identification	via LOCK bit	Ī-	_	20	ms
VIDEO AMPLIF	IER OUTPUT; NOTES 7 AND 8		•	•	•	•
M.3.1	zero signal output level	negative modulation; note 9	-	4.7	-	V
M.3.2		positive modulation; note 9	_	2.0	_	V
M.3.3	top sync level	negative modulation	1.9	2.1	2.3	V
M.3.4	white level	positive modulation	_	4.5	_	V
M.3.5	difference in amplitude between negative and positive modulation		_	0	15	%
M.3.6	video output impedance		Ī-	50	Ī-	Ω
M.3.7	internal bias current of NPN emitter follower output transistor		1.0	_	_	mA
M.3.8	maximum source current		_	_	5	mA
M.3.9	bandwidth of demodulated output signal	at -3 dB	6	7	_	MHz
M.3.10	differential gain	note 10	_	2	5	%
M.3.11	differential phase	notes 10 and 6	1-	_	5	deg

TDA935X/6X/8X PS/N2 series

NUMBER	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNI
VIDEO AMPLIF	IER (CONTINUED)					'
M.3.12	video non-linearity	note 11	_	_	5	%
M.3.13	white spot clamp level		_	5.3	_	٧
M.3.14	noise inverter clamping level	note 12	_	1.5	-	٧
M.3.15	noise inverter insertion level (identical to black level)	note 12	-	2.8	_	V
	intermodulation	notes 6 and 13				
M.3.16	blue	$V_0 = 0.92 \text{ or } 1.1 \text{ MHz}$	60	66	-	dB
M.3.17		$V_0 = 2.66 \text{ or } 3.3 \text{ MHz}$	60	66	_	dB
M.3.18	yellow	V _o = 0.92 or 1.1 MHz	56	62	_	dB
M.3.19		$V_0 = 2.66 \text{ or } 3.3 \text{ MHz}$	60	66	_	dB
	signal-to-noise ratio	notes 6 and 14				
M.3.20		weighted	56	60	_	dB
M.3.21		unweighted	49	53	_	dB
M.3.22	residual carrier signal	note 6	_	5.5	_	mV
M.3.23	residual 2nd harmonic of carrier signal	note 6	_	2.5	-	mV
IF AND TUNER	AGC; NOTE 15			1	•	•
Timing of IF-	4 <i>GC</i>					
M.4.1	modulated video interference	30% AM for 1 mV to 100 mV; 0 to 200 Hz (system B/G)	_	-	10	%
M.4.2	response time to IF input signal amplitude increase of 52 dB	positive and negative modulation	_	2	-	ms
M.4.3	response to an IF input signal	negative modulation	_	50	_	ms
M.4.4	amplitude decrease of 52 dB	positive modulation	_	100	_	ms
Tuner take-ov	ver adjustment (via I ² C-bus)			<u>'</u>	<u>'</u>	
M.5.1	minimum starting level for tuner take-over (RMS value)		_	0.4	8.0	mV
M.5.2	maximum starting level for tuner take-over (RMS value)		80	150	_	mV
Tuner control	output					
M.6.1	maximum tuner AGC output voltage	maximum tuner gain; note 3	_	-	8	V
M.6.2	output saturation voltage	minimum tuner gain; I _O = 2 mA	_	_	300	mV
M.6.3	maximum tuner AGC output swing		5	-	_	mA
M.6.4	leakage current RF AGC		_	_	1	μΑ
M.6.5	input signal variation for complete tuner control		0.5	2	4	dB

TDA935X/6X/8X PS/N2 series

NUMBER	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
AFC OUTPUT	(VIA I ² C-BUS); NOTE 16				•	•
M.7.1	AFC resolution		_	2	_	bits
M.7.2	window sensitivity		_	125	_	kHz
M.7.3	window sensitivity in large window mode		_	275	_	kHz
VIDEO IDENTIF	FICATION OUTPUT (VIA IFI BIT IN OUTF	PUT BYTE 00)				
M.8.1	delay time of identification after the AGC has stabilized on a new transmitter		-	-	10	ms

TDA935X/6X/8X PS/N2 series

NUMBER	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
QSS Sound	IF circuit		-!	-1		
SOUND IF AMI	PLIFIER					
	input sensitivity (RMS value)					
Q.1.1		FM mode (-3 dB)	_	30	70	μV
Q.1.2		AM mode (-3 dB)	_	60	100	μV
	maximum input signal					
Q.1.3	(RMS value)	FM mode	50	70	_	mV
Q.1.4		AM mode	80	140	_	mV
Q.1.5	input resistance (differential)	note 3	_	2	_	kΩ
Q.1.6	input capacitance (differential)	note 3	_	3	_	pF
Q.1.7	gain control range		64	_	_	dB
Q.1.8	crosstalk attenuation between SIF and VIF input		50	_	_	dB
SOUND IF INT	ERCARRIER OUTPUT; WITH AM = 0					
Q.2.1	output signal amplitude (RMS value)	SC-1; sound carrier 2 off	75	100	125	mV
Q.2.2	bandwidth (-3 dB)		7.5	10	_	MHz
Q.2.3	residual IF sound carrier (RMS value)		-	2	_	mV
Q.2.4	output resistance		_	300	_	Ω
Q.2.5	DC output voltage		_	2.5	_	٧
Q.2.6	internal bias current of emitter follower		_	1.0	_	mA
Q.2.7	maximum AC and DC sink current		_	1.0	_	mA
Q.2.8	maximum AC and DC source current		-	1.0	_	mA
Q.2.9	weighted S/N ratio (SC1/SC2).	black picture	53/48	58/55	_	dB
Q.2.10	Ratio of PC/SC1 at vision IF	white picture	52/47	55/53	_	dB
Q.2.11	input of 40 dB or higher, note 17	6 kHz sinewave (black-to-white modulation)	44/42	48/46	_	dB
Q.2.12		250 kHz sine wave (black-to-white modulation)	44/25	48/30	-	dB
Q.2.13		sound carrier subharmonics (f=2.75 MHz ± 3 kHz)	45/44	51/50	_	dB
Q.2.14		sound carrier subharmonics (f=2.87 MHz ± 3 kHz)	46/45	52/51	_	dB

TDA935X/6X/8X PS/N2 series

NUMBER	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
AM SOUND OL	JTPUT; DEPENDING ON SETTING OF C	MB0/CMB1 AND AM BITS	'	-	-	•
Q.3.1	AF output signal amplitude (RMS value)	54% modulation	400	500	600	mV
Q.3.2	total harmonic distortion	54% modulation	_	0.5	1.0	%
Q.3.21	total harmonic distortion	80% modulation	_	2.0	5.0	%
Q.3.3	AF bandwidth	−3 dB	100	125	_	kHz
Q.3.4	weighted signal-to-noise ratio		47	53	_	dB
Q.3.5	DC output voltage		_	2.5	_	٧
Q.3.6	power supply ripple rejection		_	40	_	dB
FM demodul	ator and audio amplifier			•	•	
FM-PLL DEM	ODULATOR; NOTE 18					
G.1.2	gain control range AGC amplifier		26	30	_	dB
G.1.3	catching range PLL	note 19	_	±225	_	kHz
G.1.4	maximum phase detector output current		_	±100	_	μΑ
G.1.5	VCO steepness Δf _{FM} /ΔV _C (K ₀)		_	3.3	_	MHz/\
G.1.6	phase detector steepness $\Delta I_{C}/\Delta \phi_{VFM}$ (K_{D})		_	9	_	μA/rac
G.1.7	AM rejection	note 20	40	46	_	dB
EXTERNAL SO	UND IF INPUT (SNDIF, WHEN SELECT	ED)				
G.1.8	input limiting for lock-in of PLL (RMS value)		-	1	2	mV
G.1.9	input resistance	note 3	_	50	_	kΩ
G.1.10	input capacitance	note 3	_	_	1.0	pF
DE-EMPHASIS	OUTPUT; NOTE 22		·			
G.2.1	output signal amplitude (RMS value)	notes 19 and 21	-	500	-	mV
G.2.2	output resistance		_	15	_	kΩ
G.2.3	DC output voltage		_	3.2	_	V
G.2.31	signal-to-noise ratio (RMS value)	note 23	_	50	_	dB
AUDIO INPUT V	/IA DEEMPHASIS OUTPUT; NOTE 22					
G.2.4	input signal amplitude (RMS value)		_	500	_	mV
G.2.5	input resistance		_	15	_	kΩ
G.2.6	voltage gain between input and output	maximum volume	-	9	_	dB

TDA935X/6X/8X PS/N2 series

NUMBER	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
Audio Ampli	ifier					1
AUDIO OUTPU	T OR VOLUME CONTROLLED AM-OUT					
A.1.1	controlled output signal amplitude (RMS value)	-6 dB; nominal audio input signal	500	700	900	mV
A.1.2	output resistance		_	500	_	Ω
A.1.3	DC output voltage		_	3.6	_	٧
A.1.4	total harmonic distortion	note 24	_	_	0.5	%
A.1.6	power supply rejection	note 6	_	2.5	_	dB
A.1.7	internal signal-to-noise ratio	note 6 + 23 + 26	_	50	_	dB
A.1.8	external signal-to-noise ratio	note 6 + 26	_	60	_	dB
A.1.10	control range	see also Fig.30	_	80	_	dB
A.1.11	suppression of output signal when mute is active		_	80	_	dB
A.1.12	DC shift of the output when mute is active		_	10	50	mV
EXTERNAL AU	DIO INPUT			1	1	
A.2.1	input signal amplitude (RMS value)		_	500	2000	mV
A.2.2	input resistance		_	25	_	kΩ
A.2.3	voltage gain between input and output	maximum volume	_	9	_	dB
A.2.4	crosstalk between internal and external audio signals		60	_	_	dB
AUTOMATIC V	DLUME LEVELLING; NOTE 28					
A.3.1	gain at maximum boost		_	6	_	dB
A.3.2	gain at minimum boost		_	-14	_	dB
A.3.3	charge (attack) current		_	1	_	mA
A.3.4	discharge (decay) current		_	200	_	nA
A.3.5	control voltage at maximum boost		_	1	_	V
A.3.6	control voltage at minimum boost		_	5	_	V

I

TV signal processor-Teletext decoder with embedded $\mu\text{-}Controller$

TDA935X/6X/8X PS/N2 series

NUMBER	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
CVBS, Y/C a	nd RGB/YUV INPUTS		•	-1	'	•
CVBS-Y/C sv	WITCH					
S.1.1	CVBS or Y input voltage (peak-to-peak value)	note 29	_	1.0	1.4	V
S.1.2	CVBS or Y input current		_	4	_	μΑ
S.1.3	suppression of non-selected CVBS input signal	notes 6 and 30	50	_	_	dB
S.1.4	chrominance input voltage (burst amplitude)	note 3 and 31	_	0.3	1.0	V
S.1.5	chrominance input impedance		_	50	_	kΩ
CVBS OUTPU	TON IFVO2					
S.1.9	output signal amplitude (peak-to-peak value)		_	2.0	-	V
S.1.10	top sync level		_	1.8	_	٧
S.1.11	output impedance		_	_	50	Ω
EXTERNAL RO	GB/YUV INPUT			•	-	•
S.2.1	RGB input signal amplitude for an output signal of 2 V (black-to-white) (peak-to-peak value)	note 32	_	0.7	0.8	V
S.2.2	RGB input signal amplitude before clipping occurs (peak-to-peak value)	note 6	1.0	_	_	V
S.2.3	Y input signal amplitude (peak-to-peak value)	input signal amplitude for an output signal of 2 V	_	1.4/1.0	2.0	V
S.2.4	U/P _B input signal amplitude (peak-to-peak value)	(black-to-white); when activated via the YUV1/YUV0	_	-1.33/ +0.7	2.0	V
S.2.5	V/P _R input signal amplitude (peak-to-peak value)	bits; note 33	_	-1.05/ +0.7	1.5	V
S.2.6	difference between black level of internal and external signals at the outputs		-	_	20	mV
S.2.7	input currents	no clamping; note 3	_	0.1	1	μΑ
S.2.8	delay difference for the three channels	note 6	_	0	20	ns

TDA935X/6X/8X PS/N2 series

NUMBER	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
FAST INSERTION	ON		•	-	1	•
S.3.1	input voltage	no data insertion]_	_	0.4	V
S.3.2		data insertion	0.9	_	_	V
S.3.3	maximum input pulse	insertion	_	_	3.0	٧
S.3.4	delay time from RGB in to RGB out	data insertion; note 6	_	-	20	ns
S.3.5	delay difference between insertion to RGB out and RGB in to RGB out	data insertion; note 6	_	_	20	ns
S.3.6	input current		_	_	0.2	mA
S.3.7	suppression of internal RGB signals	notes 6 and 30; insertion; f _i = 0 to 5 MHz	_	55	_	dB
S.3.8	suppression of external RGB signals	notes 6 and 30; no insertion; $f_i = 0$ to 5 MHz	_	55	_	dB
Chrominano	e and Luminance filters				1	•
CHROMINANC	E TRAP CIRCUIT; NOTE 34					
F.1.1	trap frequency		_	f _{osc}	_	MHz
F.1.2	Bandwidth at f _{SC} = 3.58 MHz	−3 dB	_	2.8	_	MHz
F.1.3	Bandwidth at f _{SC} = 4.43 MHz	−3 dB	_	3.4	_	MHz
F.1.4	colour subcarrier rejection		24	26	_	dB
F.1.5	trap frequency during SECAM reception		_	4.3	-	MHz
CHROMINANC	E BANDPASS CIRCUIT					•
F.2.1	centre frequency (CB = 0)		_	f _{osc}	_	MHz
F.2.2	centre frequency (CB = 1)		_	1.1×f _{osc}	_	MHz
F.2.3	bandpass quality factor		_	3	_	
CLOCHE FILTE	iR					•
F.3.1	centre frequency		4.26	4.29	4.31	MHz
F.3.2	Bandwidth		241	268	295	kHz
Y DELAY LINE						
F.4.1	delay time	note 6	_	480	_	ns
F.4.2	tuning range delay time	8 steps	-160	_	+160	ns
F.4.3	bandwidth of internal delay line	note 6	8	_	_	MHz

TDA935X/6X/8X PS/N2 series

NUMBER	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
Picture Impr	ovement Features		<u>'</u>	•	1	!
PEAKING CON	TROL; NOTE 35					
P.1.1	width of preshoot or overshoot	note 3	_	160	_	ns
P.1.2	peaking signal compression threshold		_	50	_	IRE
P.1.3	overshoot at maximum peaking	positive	_	45	_	%
P.1.4		negative	_	80	_	%
P.1.5	Ratio negative/positive overshoot; note 36		_	1.8	_	
P.1.6	peaking control curve	63 steps	see Fig.3	31		
P.1.7	peaking centre frequency	setting PF1/PF0 = 0/0	_	2.7	_	MHz
P.1.8		setting PF1/PF0 = 0/1	_	3.1	_	MHz
P.1.9		setting PF1/PF0 = 1/0	_	3.5	_	MHz
BLACK LEVEL	STRETCHER; NOTE 37	•	,	•	,	•
P.2.1	Maximum black level shift		15	21	27	IRE
P.2.2	level shift at 100% peak white		-1	0	1	IRE
P.2.3	level shift at 50% peak white		-1	-	3	IRE
P.2.4	level shift at 15% peak white		6	8	10	IRE

TDA935X/6X/8X PS/N2 series

NUMBER	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
Horizontal a	nd vertical synchronization and	drive circuits	<u>'</u>	1	!	
SYNC VIDEO I	NPUT					
H.1.1	sync pulse amplitude	note 3	50	300	350	mV
H.1.2	slicing level for horizontal sync	note 38	_	50	_	%
H.1.3	slicing level for vertical sync	note 38	_	35	_	%
HORIZONTAL (OSCILLATOR	,			•	
H.2.1	free running frequency		_	15625	_	Hz
H.2.2	spread on free running frequency		-	-	±2	%
H.2.3	frequency variation with respect to the supply voltage	$V_P = 8.0 \text{ V} \pm 10\%$; note 6	_	0.2	0.5	%
H.2.4	frequency variation with temperature	T _{amb} = 0 to 70 °C; note 6	_	_	80	Hz
FIRST CONTR	OL LOOP; NOTE 39					
H.3.1	holding range PLL		_	±0.9	±1.2	kHz
H.3.2	catching range PLL	note 6	±0.6	±0.9	_	kHz
H.3.3	signal-to-noise ratio of the video input signal at which the time constant is switched		_	24	_	dB
H.3.4	hysteresis at the switching point		_	3	_	dB
SECOND CON	TROL LOOP					•
H.4.1	control sensitivity		_	150	_	μs/μs
H.4.2	control range from start of horizontal output to flyback at nominal shift position		-	19	-	μs
H.4.3	horizontal shift range	63 steps	±2	_	_	μs
H.4.4	control sensitivity for dynamic compensation		_	7.6	_	μs/V
H.4.5	Voltage to switch-on the 'flash' protection	note 40	6.0	-	_	V
H.4.6	Input current during protection		_	-	1	mA
H.4.7	control range of the parallelogram correction	note 41	-	±0.5	_	μs
H.4.8	control range of the bow correction	note 41	_	±0.5	_	μs

TDA935X/6X/8X PS/N2 series

NUMBER	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
HORIZONTAL	OUTPUT; NOTE 42		Į.		1	
H.5.1	LOW level output voltage	I _O = 10 mA	_	_	0.3	٧
H.5.2	maximum allowed output current		10	_	_	mA
H.5.3	maximum allowed output voltage		_	_	V _P	V
H.5.4	duty factor	V _{OUT} = LOW (T _{ON})	_	55	_	%
H.5.5	switch-on time of horizontal drive pulse		_	1175	_	ms
H.5.6	switch-off time of horizontal drive pulse		_	43	_	ms
FLYBACK PUL	SE INPUT AND SANDCASTLE OUTPUT					
H.6.1	required input current during flyback pulse	note 3	100	_	300	μΑ
H.6.2	output voltage	during burst key	4.8	5.3	5.8	٧
		during blanking	2.3	2.5	2.7	V
H.6.3	clamped input voltage during flyback		2.6	3.0	3.4	V
H.6.4	pulse width	burst key pulse	3.3	3.5	3.7	μs
H.6.5		vertical blanking, note 43	_	14/9.5	_	lines
H.6.6	delay of start of burst key to start of sync		4.6	4.8	5.0	μs
VERTICAL OS	CILLATOR; NOTE 44		•			•
H.7.1	free running frequency		_	50/60	_	Hz
H.7.2	locking range		45	_	64.5/72	Hz
H.7.3	divider value not locked		_	625/525	_	lines
H.7.4	locking range		434/488	_	722	lines/ frame
VERTICAL RA	MP GENERATOR					
H.8.1	sawtooth amplitude (peak-to-peak value)	VS = 1FH; C = 100 nF; R = 39 kΩ	_	3.0	_	V
H.8.2	discharge current		_	1	_	mA
H.8.3	charge current set by external resistor	note 45	_	16	_	μΑ
H.8.4	vertical slope	63 steps; see Fig. 47	-20	_	+20	%
H.8.5	charge current increase	f = 60 Hz	_	19	_	%
H.8.6	LOW level of ramp		_	2.3	_	٧
VERTICAL DE	RIVE OUTPUTS					
H.9.1	differential output current (peak-to-peak value)	VA = 1FH	_	0.95	_	mA
H.9.2	common mode current		_	400	_	μΑ
H.9.3	output voltage range		0	_	4.0	V

TDA935X/6X/8X PS/N2 series

NUMBER	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
EHT TRACKIN	G/OVERVOLTAGE PROTECTION				-	
H.10.1	input voltage		1.2	_	2.8	V
H.10.2	scan modulation range		- 5	_	+5	%
H.10.3	vertical sensitivity		_	6.3	_	%/V
H.10.4	EW sensitivity	when switched-on	_	-6.3	_	%/V
H.10.5	EW equivalent output current		+100	_	-100	μΑ
H.10.6	overvoltage detection level	note 40	_	3.9	Ī-	V
DE-INTERLACE				•	•	
H.11.1	first field delay		_	0.5H	_	
EW WIDTH; N	OTE 46			•	•	1
H.12.1	control range	63 steps; see Fig. 50	100	_	65	%
H.12.2	equivalent output current		0	_	700	μΑ
H.12.3	EW output voltage range		1.0	_	5.0	V
H.12.4	EW output current range		0	_	1200	μΑ
EW PARABOLA	A/WIDTH			•		•
H.13.1	control range	63 steps; see Fig. 51	0	_	23	%
H.13.2	equivalent output current	EW=3FH; CP=11H; TC=1FH	0	_	450	μΑ
EW UPPER/LO	OWER CORNER/PARABOLA		!	•		1
H.14.1	control range	63 steps; see Fig. 52	-46	_	+17	%
H.14.2	equivalent output current	PW=3FH; EW=3FH; TC=1FH	-207	_	+76	μΑ
EW TRAPEZIU	M			1	1	· I
H.15.1	control range	63 steps; see Fig. 53	- 5	_	+5	%
H.15.2	equivalent output current	EW=1FH; CP=11H; PW=1FH	-100	_	+100	μΑ
VERTICAL AME	PLITUDE				.!	1
H.16.1	control range	63 steps; see Fig. 46	80	Ī-	120	%
H.16.2	equivalent differential vertical drive output current	SC = 0EH	760	_	1140	μА
	(peak-to-peak value)					
VERTICAL SHI	FT			•	•	•
H.17.1	control range	63 steps; see Fig. 48	-5	_	+5	%
H.17.2	equivalent differential vertical drive output current (peak-to-peak value)		-50	-	+50	μА
S-CORRECTIO				1	1	
		CO stansuasa Fig. 40	10		05	0/
H.18.1	control range	63 steps; see Fig. 49	-10	⁻ - 47	25	%
	OM MODE (OUTPUT CURRENT VARIATIO	ON WITH RESPECT TO NOMINAL SO		1	14.00	1
H.19.1	vertical expand factor		0.75	-	1.38	
H.19.2	output current limiting and RGB blanking		_	1.05	_	

TDA935X/6X/8X PS/N2 series

NUMBER	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
Colour demo	odulation part		-	-1	-1	!
CHROMINANC	E AMPLIFIER					
D.1.1	ACC control range	note 48	26	_	_	dB
D.1.2	change in amplitude of the output signals over the ACC range		_	_	2	dB
D.1.3	threshold colour killer ON		-30	_	_	dB
D.1.4	hysteresis colour killer OFF	strong signal conditions; S/N ≥ 40 dB; note 6	_	+3	_	dB
D.1.5		noisy input signals; note 6	_	+1	_	dB
ACL CIRCUIT;	NOTE 49					
D.2.1	chrominance burst ratio at which the ACL starts to operate		_	3.0	_	
REFERENCE F	ART					
Phase-locked	d loop					
D.3.1	catching range		±500	_	_	Hz
D.3.2	phase shift for a ±400 Hz deviation of the oscillator frequency	note 6	-	-	2	deg
HUE CONTROI	_		•			•
D.5.1	hue control range	63 steps; see Fig.32	±35	±40	_	deg
D.5.2	hue variation for ±10% V _P	note 6	_	0	_	deg
D.5.3	hue variation with temperature	$T_{amb} = 0 \text{ to } 70 ^{\circ}\text{C}; \text{ note } 6$	_	0	_	deg
DEMODULATO	RS					
General						
D.6.3	spread of signal amplitude ratio between standards	note 6	-1	-	+1	dB
D.6.5	bandwidth of demodulators	-3 dB; note 50	_	650	_	kHz
PAL/NTSC d	emodulator					
D.6.6	gain between both demodulators G(B–Y) and G(R–Y)		1.60	1.78	1.96	
D.6.12	change of output signal amplitude with temperature	note 6	-	0.1	_	%/K
D.6.13	change of output signal amplitude with supply voltage	note 6	-	-	±0.1	dB
D.6.14	phase error in the demodulated signals	note 6	-	-	±5	deg

TDA935X/6X/8X PS/N2 series

NUMBER	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNI.	
SECAM dem	nodulator		!		.!	_	
D.7.1	black level off-set		_	_	7	kHz	
D.7.2	pole frequency of deemphasis		77	85	93	kHz	
D.7.3	ratio pole and zero frequency		_	3	_		
D.7.4	non linearity		_	_	3	%	
D.7.5	calibration voltage		1.8	2.3	2.8	V	
Base-band d	lelay line						
D.8.1	variation of output signal for adjacent time samples at constant input signals		-0.1	_	0.1	dB	
D.8.2	residual clock signal (peak-to-peak value)		-	_	5	mV	
D.8.3	delay of delayed signal		63.94	64.0	64.06	μs	
D.8.4	delay of non-delayed signal		40	60	80	ns	
D.8.5	difference in output amplitude with delay on or off		_	_	5	%	
COLOUR DIFF	ERENCE MATRICES (IN CONTROL CIF	RCUIT)			-	•	
PAL/SECAM	mode; (R-Y) and (B-Y) not affect	ted					
D.9.1	ratio of demodulated signals (G–Y)/(R–Y)		_	-0.51 ±10%	_		
D.9.2	ratio of demodulated signals (G-Y)/(B-Y)		-	-0.19 ±25%	_		
NTSC mode,	the matrix results in the following	g signals (nominal hue setti	ng)	1	•	'	
MUS-bit = 0							
D.9.6	(B-Y) signal: 2.03/0°			2.03U _R			
D.9.7	(R-Y) signal: 1.59/95°		-0.	14U _R + 1.5	58V _R		
D.9.8	(G-Y) signal: 0.61/240°		-0.	31U _R – 0.5	3V _R		
MUS-bit = 1		•	<u>.</u>			•	
D.9.9	(B-Y) signal: 2.20/-1°		2.2	20U _R – 0.0	4V _R		
D.9.10	(R-Y) signal: 1.53/99°		-0.	-0.24U _R + 1.51V _R			
D.9.11	(G-Y) signal: 0.70/223°		-0.	-0.51U _R - 0.48V _R			
REFERENCE S	SIGNAL OUTPUT/SWITCH OUTPUT; NO	DTE 51					
D.10.1	reference frequency	CMB1/CMB0 = 01		3.58/4.43		MHz	
D.10.2	output signal amplitude (peak-to-peak value)	CMB1/CMB0 = 01	0.2	0.25	0.3	V	
D.10.3	output level (mid position)	CMB1/CMB0 = 01	2.3	2.5	2.7	V	
D.10.4	output level LOW	CMB1/CMB0 = 10	_	_	0.8	V	
D.10.5	output level HIGH	CMB1/CMB0 = 11	4.5			V	

TDA935X/6X/8X PS/N2 series

NUMBER	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNI
Control part	1		•	•	-	•
SATURATION C	CONTROL; NOTE 32					
C.1.1	saturation control range	63 steps; see Fig.33	52	_	_	dB
CONTRAST CO	ONTROL; NOTE 32			•	•	1
C.2.1	contrast control range	63 steps; see Fig.34	_	20	_	dB
C.2.2	tracking between the three channels over a control range of 10 dB		-	_	0.5	dB
C.2.6	contrast reduction		_	10	_	dB
BRIGHTNESS (CONTROL					
C.3.1	brightness control range	63 steps; see Fig.35	_	±0.7	_	V
RGB AMPLIFIE	ERS					
C.4.1	output signal amplitude (peak-to-peak value)	at nominal luminance input signal, nominal settings for contrast, white-point adjustment and cathode drive level(CL3-CL0 = 0111)	_	2.0	_	V
C.4.2	maximum signal amplitude (black-to-white)	note 52	_	5.5	-	V
C.4.3	maximum peak white level		_	5.5	_	V
C.4.4	output signal amplitude for the 'red' channel (peak-to-peak value)	at nominal settings for contrast and saturation control and no luminance signal to the input (R-Y, PAL)	_	2.1	_	V
C.4.5	nominal black level voltage		_	2.5	_	V
C.4.6	black level voltage	when black level stabilisation is switched-off (via AKB bit)	_	2.5	-	V
C.4.61	black level voltage control range	AVG bit active; note 53	1.8	2.5	3.2	V
C.4.7	width of video blanking with HBL bit active	note 54	13.4	13.7	14.0	μs
C.4.71	timing of video blanking with	start of blanking; note 54	3.5	_	5.9	μs
C.4.72	respect to mid sync (HBL = 1)	end of blanking; note 54	7.8	_	10.2	μs
C.4.8	control range of the black-current stabilisation		_	±1	_	V
C.4.81	RGB output level when RGBL=1		_	0.8	_	V
C.4.9	blanking level	difference with black level,	_	-0.5	_	V
C.4.10	level during leakage measurement	note 52	_	-0.1	_	V
C.4.11	level during 'low' measuring pulse		_	0.25	_	V
C.4.12	level during 'high' measuring pulse; note 55		_	0.5	_	V

TDA935X/6X/8X PS/N2 series

NUMBER	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
C.4.13	adjustment range of the cathode drive level	note 52	_	±3	_	dB
C.4.131	gain control range to compensate spreads in picture tube characteristics for the 2-point black -current stabilization system; note 55		_	±6	_	dB
C.4.14	variation of black level with temperature	note 6	_	1.0	_	mV/K
C.4.141	black level off-set adjustment on the Red and Green channel	63 steps	_	±160	_	mV
C.4.21	signal-to-noise ratio of the output	RGB input; note 56	60	_	_	dB
C.4.22	signals	CVBS input; note 56	50	_	_	dB
C.4.23	residual voltage at the RGB	at f _{osc}	_	_	15	mV
C.4.24	outputs (peak-to-peak value)	at 2f _{osc} plus higher harmonics	_	_	15	mV
C.4.25	bandwidth of output signals	RGB input; at -3 dB	_	15	_	MHz
C.4.26		CVBS input; at –3 dB; f _{osc} = 3.58 MHz	_	2.8	-	MHz
C.4.27		CVBS input; at –3 dB; f _{osc} = 4.43 MHz	_	3.4	_	MHz
C.4.28		S-VHS input; at –3 dB	5	_	_	MHz
WHITE-POINT	ADJUSTMENT			•	-1	1
C.5.1	I ² C-bus setting for nominal gain	HEX code	_	20H	_	
C.5.2	adjustment range of the relative R, G and B drive levels		_	±3	_	dB
2-POINT BLAC	K-CURRENT STABILIZATION, NOTES 57	7			•	•
C.6.1	amplitude of 'low' reference current		_	8	_	μΑ
C.6.2	amplitude of 'high' reference current; note 55		_	40	_	μΑ
C.6.3	acceptable leakage current		_	±75	_	μΑ
C.6.4	maximum current during scan		_	2	_	mA
C.6.5	input impedance		_	500	_	Ω
C.6.7	minimum input current to activate the guard circuit	IVG bit = "0", note 58	_	0.1	_	mA
BEAM CURREN	-	L	ļ	ļ	ļ	1
C.7.1	contrast reduction starting voltage		_	2.8	-	V
C.7.2	voltage difference for full contrast reduction		_	1.8	_	V
C.7.3	brightness reduction starting voltage		_	1.7	_	V
C.7.4	voltage difference for full brightness reduction		_	0.9	_	V

TDA935X/6X/8X PS/N2 series

NUMBER	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
C.7.5	internal bias voltage		_	3.3	_	V
C.7.6	detection level vertical guard	IVG bit = "1"; note 58	-	3.45	_	V
C.7.7	minimum input current to activate the guard circuit	IVG bit = "1"; note 58	_	100	_	μΑ
C.7.8	maximum allowable current		_	1	_	mA
FIXED BEAM C	CURRENT SWITCH-OFF; NOTE 59					
C.8.1	discharge current during switch-off		0.85	1.0	1.15	mA
C.8.2	discharge time of picture tube		_	38	_	ms

Notes

- 1. When the 3.3 V supply is present and the μ-Controller is active a 'low-power start-up' mode can be activated. When all sub-address bytes have been sent and the POR and XPR flags have been cleared the horizontal output can be switched-on via the STB-bit (subaddress 24H). In this condition the horizontal drive signal has the nominal T_{OFF} and the T_{ON} grows gradually from zero to the nominal value. As soon as the 8 V supply is present the switch-on procedure (e.g. closing of the second loop) is continued.
- 2. On set AGC.
- 3. This parameter is not tested during production and is just given as application information for the designer of the television receiver.
- 4. Loop bandwidth BL = 60 kHz (natural frequency fN = 15 kHz; damping factor d = 2; calculated with top sync level as FPLL input signal level).
- 5. The IF-PLL demodulator uses an internal VCO (no external LC-circuit required) which is calibrated by means of a digital control circuit which uses the clock frequency of the μ-Controller as a reference. The required IF frequency for the various standards is set via the IFA-IFC bits in subaddress 27H. When the system is locked the resulting IF frequency is very accurate with a deviation from the nominal value of less than 25 kHz.
- 6. This parameter is not tested during production but is guaranteed by the design and qualified by means of matrix batches which are made in the pilot production period.
- 7. Measured at 10 mV (RMS) top sync input signal.
- 8. Via this pin both the demodulated IF signal and the selected CVBS (or Y+C) signal can be supplied to the output. The selection between both signals is realised by means of the SVO bit in subaddress 22H.
- 9. So called projected zero point, i.e. with switched demodulator.
- 10. Measured in accordance with the test line given in Fig.36. For the differential phase test the peak white setting is reduced to 87%.

The differential gain is expressed as a percentage of the difference in peak amplitudes between the largest and smallest value relative to the subcarrier amplitude at blanking level.

The phase difference is defined as the difference in degrees between the largest and smallest phase angle.

- 11. This figure is valid for the complete video signal amplitude (peak white-to-black), see Fig.37.
- 12. The noise inverter is only active in the 'strong signal mode' (no noise detected in the incoming signal)
- 13. The test set-up and input conditions are given in Fig.38. The figures are measured with an input signal of 10 mV RMS. This test can only be carried out in a test set-up in which the test options of the IC can be activated. This because the IF-AGC control input is not available in this IC.

TDA935X/6X/8X PS/N2 series

- 14. Measured at an input signal of 10 mV_{RMS}. The S/N is the ratio of black-to-white amplitude to the black level noise voltage (RMS value). B = 5 MHz. Weighted in accordance with CCIR 567.
- 15. The time-constant of the IF-AGC is internal and the speed of the AGC can be set via the bits AGC1 and AGC0 in subaddress 28H. The AGC response time is also dependent on the acquisition time of the PLL demodulator. The values given are valid for the 'norm' setting (AGC1-AGC0 = 0-1) and when the PLL is in lock.
- 16. The AFC control voltage is generated by the digital tuning system of the PLL demodulator. This system uses the clock frequency of the μ-Controller/Teletext decoder as a reference and is therefore very accurate. For this reason no maximum and minimum values are given for the window sensitivity figures (parameters M.7.2 and M.7.3). The tuning information is supplied to the tuning system via the AFA and AFB bits in output byte 02H. The AFC value is valid only when the LOCK-bit is 1.
- 17. The weighted S/N ratio is measured under the following conditions:
 - a) The vision IF modulator must meet the following specifications:

Incidental phase modulation for black-to-white jumps less than 0.5 degrees.

QSS AF performance, measured with the television-demodulator AMF2 (audio output, weighted S/N ratio) better than 60 dB (deviation 27 kHz) for 6 kHz sine wave black-to-white modulation.

Picture-to-sound carrier ratio: PC/SC1 = 13 dB (transmitter).

- b) The measurements must be carried out with the Siemens SAW filters G3962 for vision IF and G9350 for sound IF. Input level for sound IF 10 mV_{BMS} with 27 kHz deviation.
- c) The PC/SC ratio at the vision IF input is calculated as the addition of the TV transmitter ratio and the SAW filter PC/SC ratio. This PC/SC ratio is necessary to achieve the S/N(W) values as indicated.
- 18. Calculation of the FM-PLL filter can be done approximately by use of the following equations:

$$f_O \,=\, \frac{1}{2\pi} \sqrt{\frac{K_0 K_D}{C_P}}$$

$$v = \frac{1}{2R\sqrt{K_0K_DC_P}}$$

$$BL_{-3dB} = f_0(1.55 - v^2)$$

These equations are only valid under the conditions that $v \le 1$ and $C_S > 5C_P$.

Definitions:

 $K_0 = VCO$ steepness in rad/V

 K_D = phase detector steepness μ A/rad

R = loop filter resistor

C_S = series capacitor

C_P = parallel capacitor

f₀ = natural frequency of PLL

 BL_{-3dB} = loop bandwidth for -3dB

v = damping factor

Some examples for these values are given in table 132

- 19. Modulation frequency: 1 kHz, $\Delta f = \pm 50$ kHz.
- 20. f = 4.5/5.5 MHz; FM: 70 Hz, ± 50 kHz deviation; AM: 1.0 kHz, 30% modulation.
- 21. This figure is independent of the TV standard and valid for a frequency deviation of ± 25 kHz at a carrier frequency of 4.5 MHz or a deviation of ± 50 kHz at a carrier frequency of 5.5/6.0/6.5 MHz.

TDA935X/6X/8X PS/N2 series

- 22. The deemphasis pin can also be used as additional audio input. In that case the internal (demodulated FM signal) must be switched off. This can be realised by means of the SM (sound mute) bit. When the vision IF amplifier is switched to positive modulation the signal from the FM demodulator is automatically switched off. The external signal must be switched off when the internal signal is selected.
- 23. The signal-to-noise ratio is measured under the following conditions:
 - a) Input signal to the SNDIF pin (activated via SIF bit) with an amplitude of 100mV_{RMS} , $f_{MOD} = 1 \text{ kHz}$ and $\Delta f = 27 \text{ kHz}$
 - Output signal measured at the AUDEEM pin. The noise (RMS value) is measured according to the CCIR 468 definition.
- 24. Audio input signal 200 mV_{RMS}. Measured with a bandwidth of 15 kHz and the audio attenuator at –6 dB.
- 25. Audio input signal 1 V_{RMS} and the volume control setting such that no clipping occurs in the audio output.
- 26. Unweighted RMS value, audio input signal 500 mV_{RMS}, audio attenuator at -6 dB.
- 27. Audio attenuator at -20 dB; temperature range 10 to 50 °C.
- 28. In various versions the Automatic Volume Levelling (AVL) function can be activated. The pin to which the external capacitor has to be connected depends on the IC version. For the 90° types the capacitor is connected to the EW output pin. For the 110° types a choice can be made between the AVL function and a sub-carrier output / general purpose switch output. The selection must be made by means of the CMB0 and CMB1 bit in subaddress 22H. More details about the sub-carrier output are given in the parameters D.10.
 - The Automatic Volume Levelling (AVL) circuit stabilises automatically the audio output signal to a certain level which can be set by means of the volume control. This AVL function prevents big audio output fluctuations due to variation of the modulation depth of the transmitter. The AVL can be switched on and off via the AVL bit in subaddress 29H.
 - The AVL is active over an input voltage range (measured at the deemphasis output) of 150 to 1500 mV_{RMS}. The AVL control curve is given in Fig.39. The control range of +6 dB to -14 dB is valid for input signals with 50% of the maximum frequency deviation.
- 29. Signal with negative-going sync. Amplitude includes sync pulse amplitude.
- 30. This parameter is measured at nominal settings of the various controls.
- 31. Indicated is a signal for a colour bar with 75% saturation (chroma: burst ratio = 2.2:1).
- 32. The contrast and saturation control is active on the internal signal (YUV) and on the external RGB/YUV input. The Text/OSD input can be controlled on brightness only. Nominal contrast is specified with the DAC in position 20 HEX. Nominal saturation as maximum –10 dB.
- 33. The YUV input signal amplitudes are based on a colour bar signal with 75/100% saturation.
- 34. When the decoder is forced to a fixed subcarrier frequency (via the CM-bits) the chroma trap is always switched-on, also when no colour signal is identified. In the automatic mode the chroma trap is switched-off when no colour signal is identified.
- 35. Valid for a signal amplitude on the Y-input of 0.7 V black-to-white (100 IRE) with a rise time (10% to 90%) of 70 ns and the video switch in the Y/C mode. During production the peaking function is not tested by measuring the overshoots but by measuring the frequency response of the Y output.
- 36. The ratio between the positive and negative peaks can be varied by means of the bits RPO1 and RPO0 in subaddress 2EH. For ratios which are smaller than 1.8 the positive peak is not affected and the negative peak is reduced.
- 37. For video signals with a black level which deviates from the back-porch blanking level the signal is "stretched" to the blanking level. The amount of correction depends on the IRE value of the signal (see Fig.40). The black level is detected by means of an internal capacitor. The black level stretcher can be switched on and off via the BKS bit in subaddress 2DH. The values given in the specification are valid only when the luminance input signal has an amplitude of 1 V_{p-p}.

TDA935X/6X/8X PS/N2 series

- 38. The slicing level is independent of sync pulse amplitude. The given percentage is the distance between the slicing level and the black level (back porch). When the amplitude of the sync pulse exceeds the value of 350 mV the sync separator will slice the sync pulse at a level of 175 mV above top sync. The maximum sync pulse amplitude is 4 V_{p-p} . The vertical slicing level is dependent on the S/N ratio of the incoming video signal. For a S/N \leq 24 dB the slicing level is 35%, for a S/N \geq 24 dB the slicing level is 60%. With the bit FSL (Forced Slicing Level) the vertical slicing level can be forced to 60%.
- 39. To obtain a good performance for both weak signal and VCR playback the time constant of the first control loop is switched depending on the input signal condition and the condition of the POC, FOA, FOB and VID bits in subaddress 24H. The circuit contains a noise detector and the time constant is switched to 'slow' when too much noise is present in the signal. In the 'fast' mode during the vertical retrace time the phase detector current is increased 50% so that phase errors due to head-switching of the VCR are corrected as soon as possible. Switching of the time constant can be automatically or can be set by means of the control bits.
 - The circuit contains a video identification circuit which is independent of the first loop. This identification circuit can be used to close or open the first control loop when a video signal is present or not present on the input. This enables a stable On Screen Display (OSD) when just noise is present at the input.
 - To prevent that the horizontal synchronisation is disturbed by anti copy signals like Macrovision the phase detector is gated during the vertical retrace period so that pulses during scan have no effect on the output voltage. The width of the gate pulse is about 22 μ s. During weak signal conditions (noise detector active) the gating is active during the complete scan period and the width of the gate pulse is reduced to 5.7 μ s so that the effect of noise is reduced to a minimum.
 - The output current of the phase detector in the various conditions are shown in Table 133.
- 40. The ICs have 2 protection inputs. The protection on the second phase detector pin is intended to be used as 'flash' protection. When this protection is activated the horizontal drive is switched-off immediately and then switched-on again via the slow start procedure.
 - The protection on the EHT input is intended for overvoltage (X-ray) protection. When this protection is activated the horizontal drive is directly switched-off (via the slow stop procedure).
 - The EHT protection input can also be used to switch-off the TV receiver in a correct way when it is switched off via the mains power switch or when the power supply is interrupted by pulling the mains plug. This can be realised by means of a detection circuit which monitors the main supply voltage of the receiver. When this voltage suddenly decreases the EHT protection input must be pulled HIGH and then the horizontal drive is switched off via the slow stop procedure. Whether the EHT capacitor is discharged in the overscan or not during the switch-off period depends on the setting of the OSO bit (subaddress 25H, D4). See also note 59.
- 41. The control range indicates the maximum phase difference at the top and the bottom of the screen. Compared with the phase position at the centre of the screen the maximum phase difference at the top and the bottom of the screen is $\pm 0.5 \,\mu s$ for both the parallelogram and the bow correction.
- 42. During switch-on the horizontal drive starts-up in a soft-start mode. The horizontal drive starts with a very short T_{ON} time of the horizontal output transistor, the 'off time' of the transistor is identical to the 'off time' in normal operation. The starting frequency during switch-on is therefore about 2 times higher than the normal value. The 'on time' is slowly increased to the nominal value in a time of about 1175 ms (see Fig.43). The rather slow rise of the T_{ON} between 75% and 100% of T_{ON} is introduced to obtain a sufficiently slow rise of the EHT for picture tubes with Dynamic Astigmatic Focus (DAF) guns. When the nominal frequency is reached the PLL is closed in such a way that only very small phase corrections are necessary. This ensures a safe operation of the output stage.
 - During switch-off the soft-stop function is active. This is realised by decreasing the T_{ON} of the output transistor complimentary to the start-up behaviour. The switch-off time is about 43 ms (see Fig.43). When the 'switch off command' is received the soft-stop procedure is started after a delay of about 2 ms. During the switch-off time the EHT capacitor of the picture tube is discharged with a fixed beam current which is forced by the black current loop (see also note 59). The discharge time is about 38 ms.

The horizontal output is gated with the flyback pulse so that the horizontal output transistor cannot be switched-on during the flyback time.

TDA935X/6X/8X PS/N2 series

- 43. The vertical blanking pulse in the RGB outputs has a width of 27 or 22 lines (50 or 60 Hz system). The vertical pulse in the sandcastle pulse has a width of 14 or 9.5 lines (50 or 60 Hz system). This to prevent a phase distortion on top of the picture due to a timing modulation of the incoming flyback pulse.
- 44. The timing pulses for the vertical ramp generator are obtained from the horizontal oscillator via a divider circuit. During TV reception this divider circuit has 3 modes of operation:
 - a) Search mode 'large window'.
 - This mode is switched on when the circuit is not synchronized or when a non-standard signal (number of lines per frame outside the range between 311 and 314(50 Hz mode) or between 261 and 264 (60 Hz mode) is received). In the search mode the divider can be triggered between line 244 and line 361 (approximately 45 to 64.5 Hz).
 - b) Standard mode 'narrow window'.
 - This mode is switched on when more than 15 succeeding vertical sync pulses are detected in the narrow window. When the circuit is in the standard mode and a vertical sync pulse is missing the retrace of the vertical ramp generator is started at the end of the window. Consequently, the disturbance of the picture is very small. The circuit will switch back to the search window when, for 6 successive vertical periods, no sync pulses are found within the window.
 - c) Standard TV-norm (divider ratio 525 (60 Hz) or 625 (50 Hz).
 - When the system is switched to the narrow window it is checked whether the incoming vertical sync pulses are in accordance with the TV-norm. When 15 standard TV-norm pulses are counted the divider system is switched to the standard divider ratio mode. In this mode the divider is always reset at the standard value even if the vertical sync pulse is missing.
 - When 3 vertical sync pulses are missed the system switches back to the narrow window and when also in this window no sync pulses are found (condition 3 missing pulses) the system switches over to the search window.

The vertical divider needs some waiting time during channel-switching of the tuner. When a fast reaction of the divider is required during channel-switching the system can be forced to the search window by means of the NCIN bit in subaddress 25H.

When RGB signals are inserted the maximum vertical frequency is increased to 72 Hz. This has the consequence that the circuit can also be synchronised by signals with a higher vertical frequency like VGA.

- 45. Conditions: frequency is 50 Hz; normal mode; VS = 1F.
- 46. The output range percentages mentioned for E-W control parameters are based on the assumption that 400 μ A variation in E-W output current is equivalent to 20% variation in picture width.
- 47. The ICs have a zoom adjustment possibility for the horizontal and vertical deflection. For this reason an extra DAC has been added in the vertical amplitude control which controls the vertical scan amplitude between 0.75 and 1.38 of the nominal scan. At an amplitude of 1.06 of the nominal scan the output current is limited and the blanking of the RGB outputs is activated. This is illustrated in Fig. 42.
 - a) The nominal scan height must be adjusted at a position of 19 HEX of the vertical 'zoom' DAC.
- 48. At a chrominance input voltage of 660 mV (p-p) (colour bar with 75% saturation i.e. burst signal amplitude 300 mV (p-p)) the dynamic range of the ACC is +6 and -20 dB.
- 49. The ACL function can be activated by via the ACL bit in the subaddress 20H. The ACL circuit reduces the gain of the chroma amplifier for input signals with a chroma-to-burst ratio which exceeds a value of 3.0.
- 50. This parameter indicates the bandwidth of the complete chrominance circuit including the chrominance bandpass filter. The bandwidth of the low-pass filter of the demodulator is approximately 1 MHz.
- 51. The subcarrier output is combined with a 3-level switch output which can be used to switch external circuits like sound traps etc. This output is controlled by the CMB1 and CMB0 bits in control byte 22H. The subcarrier signal is available when CMB1/0 are set to 0/1. During the demodulation of SECAM signals the subcarrier signal is only available during the vertical retrace period. The frequency is 4.43 MHz in this condition. When CMB1/0 are set to 00 in versions for 90° picture tubes (no EW output) the output is high ohmic.

TDA935X/6X/8X PS/N2 series

52. Because of the 2-point black current stabilization circuit both the black level and the amplitude of the RGB output signals depend on the drive characteristic of the picture tube. The system checks whether the returning measuring currents meet the requirement and adapts the output level and gain of the circuit when necessary. Therefore the typical value of the black level and amplitude at the output are just given as an indication for the design of the RGB output stage.

The 2-point black level system adapts the drive voltage for each cathode in such a way that the 2 measuring currents have the right value. This has the consequence that a change in the gain of the output stage will be compensated by a gain change of the RGB control circuit. Because different picture tubes may require different drive voltage amplitudes the ratio between the output signal amplitude and the inserted measuring pulses can be adapted via the I²C-bus. This is indicated in the parameter 'Adjustment range of the cathode drive level'.

Because of the dependence of the output signal amplitude on the application the soft clipping limiting has been related to the input signal amplitude.

- 53. The alignment system for the V_{g2} voltage of the picture tube can be activated by means of the AVG bit. In that condition a certain black level is inserted at the RGB outputs during a few lines. The value of this level can be adjusted by means of the brightness control DAC. An automatic adjustment of the V_{g2} of the picture tube can be realised by using the WBC and HBC bits in output byte 01. For a black level feedback current between 12 and 20 μ A the WBC = 1, for a higher or lower current WBC = 0. Whether the current is too high or too low can be found from the HBC bit. The indication of these bits can be made visible on the screen via OSD so that this alignment procedure can also be used for service purposes.
- 54. When the reproduction of 4:3 pictures on a 16:9 picture tube is realised by means of a reduction of the horizontal scan amplitude the edges of the picture may slightly be disturbed. This effect can be prevented by adding an additional blanking to the RGB signals. The blanking pulse is derived form the horizontal oscillator and is directly related to the incoming video signal (independent of the flyback pulse). This blanking is activated with the HBL bit.
- 55. This parameter is valid only when the CCC loop is active.
- 56. Signal-to-noise ratio (S/N) is specified as peak-to-peak signal with respect to RMS noise (bandwidth 5 MHz).
- 57. This is a current input. The timing of the measuring pulses and the vertical blanking for the 50/60 Hz standard are given in Fig.44

The start-up procedure is as follows.

When the TV receiver is switched-on the RGB outputs are blanked and the black-current loop will try to adjust the picture tube to the right bias levels. The RGB drive signals are switched-on as soon as the black current loop is stabilised. This results in the shortest switch-on time.

When this switch-on system results in a visible disturbance of the picture it is possible to add a further switch-on delay via a software routine. In that case the RGB outputs must be blanked by means of the RBL bit. As soon as the black current loop is stabilised the BCF-bit is set to 0 (output byte 01). This information can then be used to switch-on the RGB outputs with some additional delay.

58. The input of the vertical guard function can be connected to the black current measuring input (BLKIN) or to the beam current limiting input (BCLIN). The switching between these modes is realised by means of the IVG bit in subaddress 2BH. When the black current input is chosen it should be noted that for a reliable operation of the protection system and the black current stabilization system the end of the protection pulse during normal operation should not overlap the measuring pulses (see also Fig.44). Therefore this pulse must end before line 14.

TDA935X/6X/8X PS/N2 series

- 59. During switch-off the magnitude of the discharge current of the picture tube is controlled by the black current loop. Dependent on the setting of the OSO bit the vertical scan can be stopped in an overscan position during that time so that the discharge is not visible on the screen. The switch-off procedure is as follows:
 - a) When the switch-off command is received the RGB outputs are blanked for a time of about 2 ms.
 - b) If OSO = 1 the vertical scan is placed in an overscan position
 - c) If OSO = 0 the vertical deflection will keep running during the switch-off time
 - d) The soft-stop procedure is started with a reduction of the T_{ON} of the output stage from nominal to zero
 - e) The fixed beam current is forced via the black current loop
 - f) The soft-stop time has a value of 43 ms, the fixed beam current is flowing during a time of 38 ms.

Table 132 Some examples for the FM-PLL filter

BL _{-3dB} (kHz)	C _S (nF)	C _P (nF)	R (k Ω)	ν
100	4.7	820	2.7	0.5
160	4.7	330	3.9	0.5

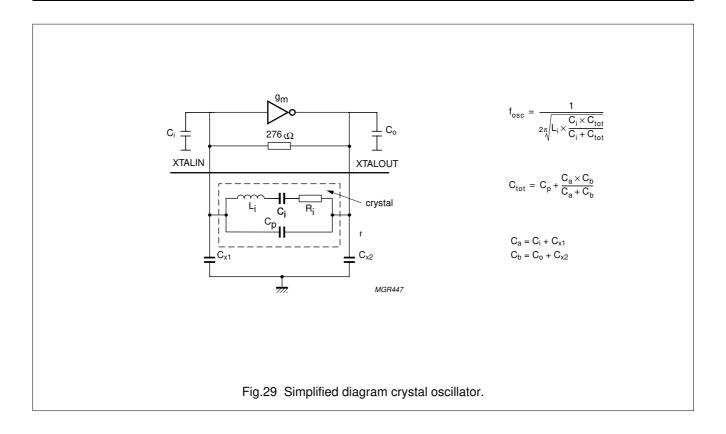
Table 133 Output current of the phase detector in the various conditions

I	² C-BUS C	OMMANDS	3	IC	CONDITIO	NS	φ-1 CURRENT/MODE			
VID	POC	FOA	FOB	IFI	SL	NOISE	SCAN	V-RETR	GATING	MODE
_	0	0	0	yes	yes	no	200	300	yes (1)	normal
_	0	0	0	yes	yes	yes	30	30	yes ⁽²⁾	normal
_	0	0	0	yes	no	_	200	300	no	normal
_	0	0	1	yes	yes	_	30	30	yes ⁽²⁾	slow
_	0	0	1	yes	no	-	200	300	no	slow
_	0	1	0	yes	yes	no	200	300	yes ⁽²⁾	slow/fast
_	0	1	0	yes	yes	yes	30	30	yes ⁽²⁾	slow/fast
_	_	1	1	-	_	-	200	300	yes ⁽¹⁾	fast
0	0	_	_	no	_	_	6	6	no	OSD
_	1	_	_	_	_	_	_	_	_	off

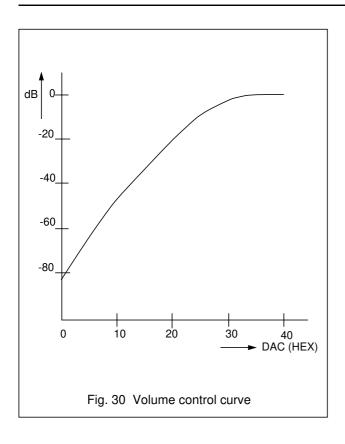
Note

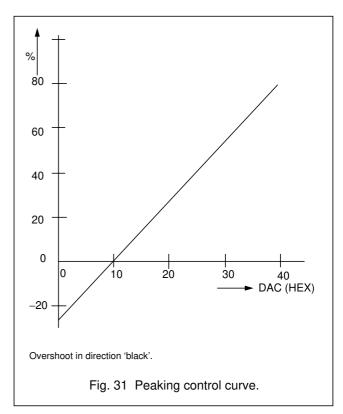
- 1. Gating is active during vertical retrace, the width is $22 \mu s$. This gating prevents disturbance due to Macro Vision Anti Copy signals.
- 2. Gating is continuously active and is 5.7 µs wide

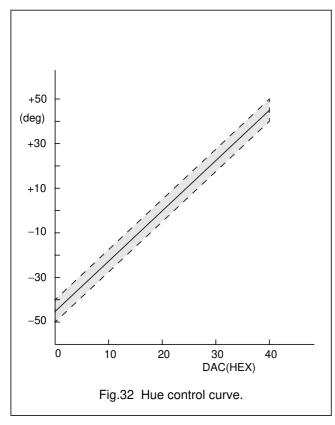
TDA935X/6X/8X PS/N2 series

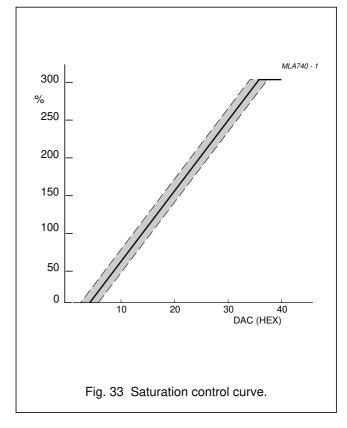


TDA935X/6X/8X PS/N2 series

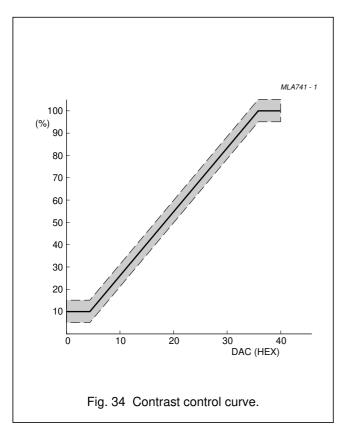


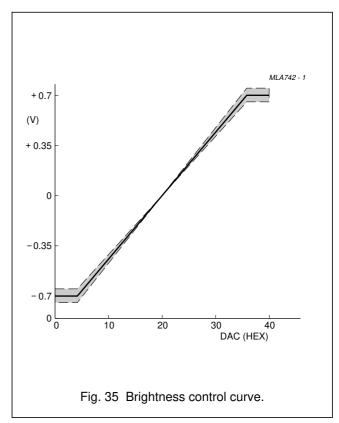


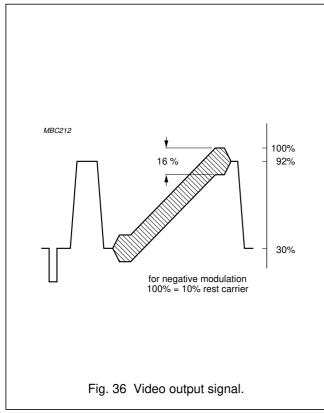


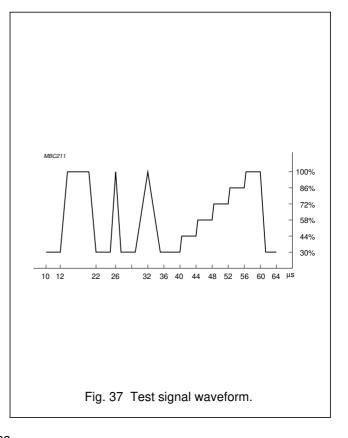


TDA935X/6X/8X PS/N2 series

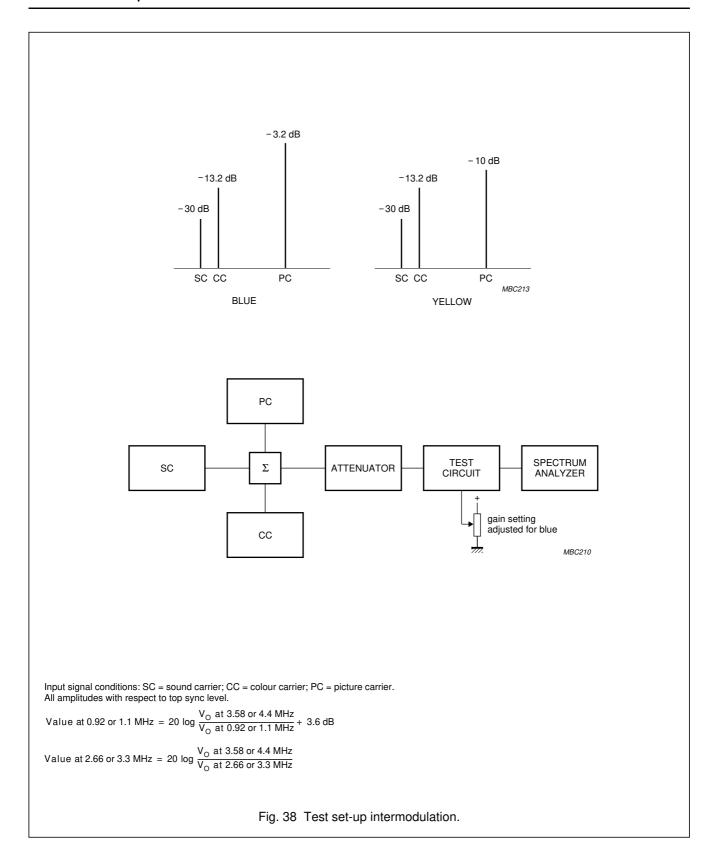






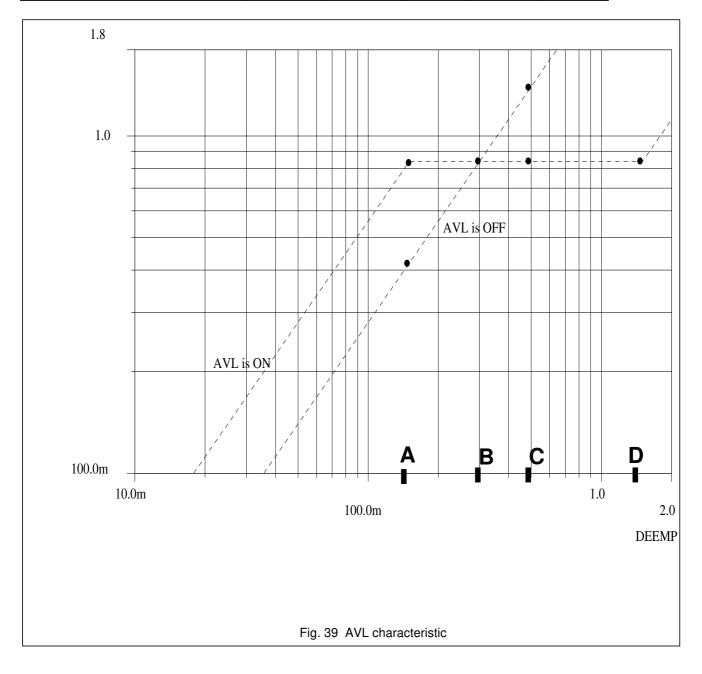


TDA935X/6X/8X PS/N2 series

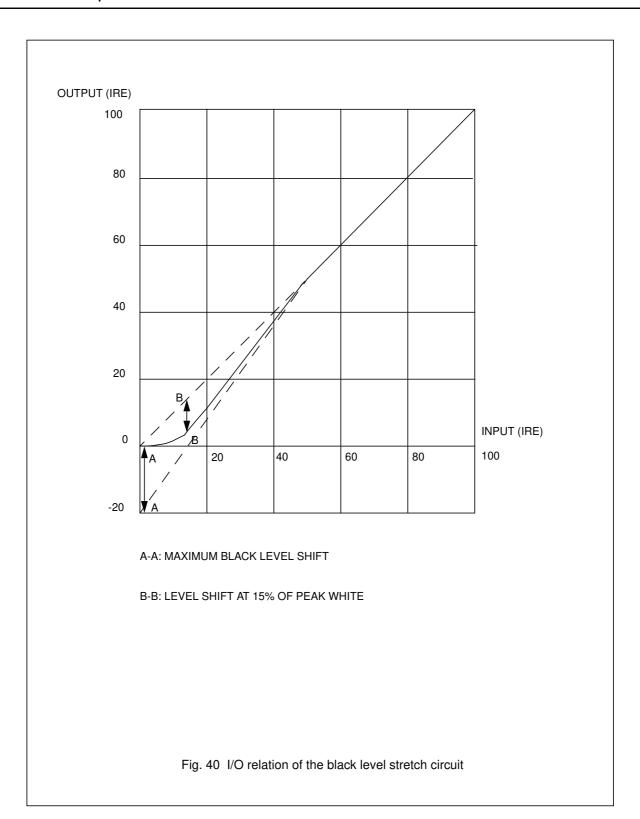


TDA935X/6X/8X PS/N2 series

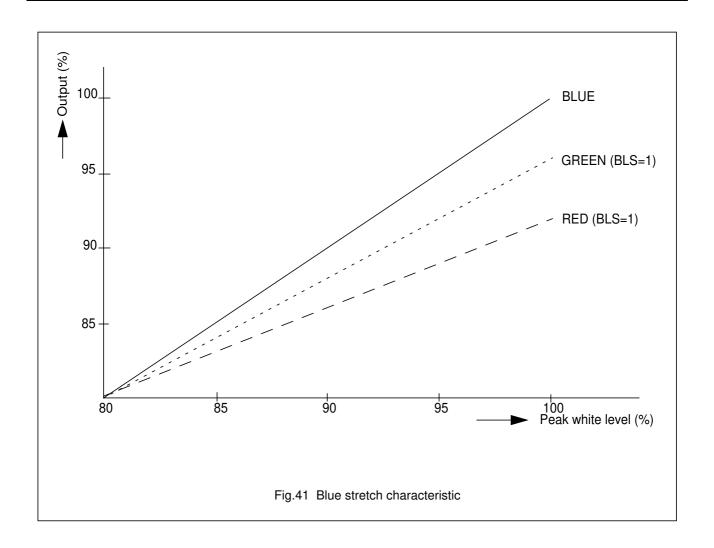
CHARACTERISTIC POINTS AVL	Α	В	С	D	UNIT
Deemphasis voltage	150	300	500	1500	mV _{RMS}
FM swing	15	30	50	150	kHz



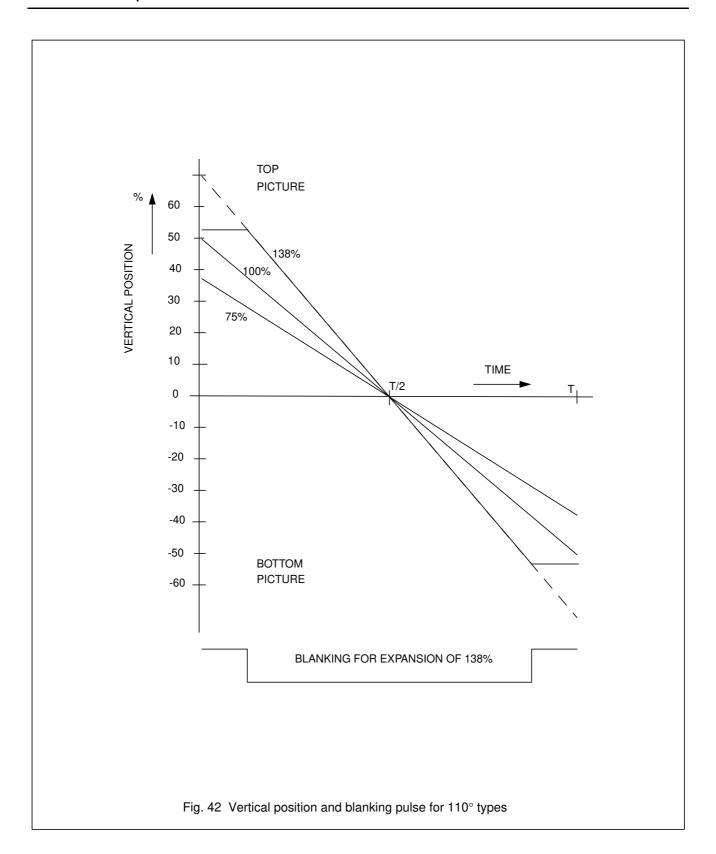
TDA935X/6X/8X PS/N2 series

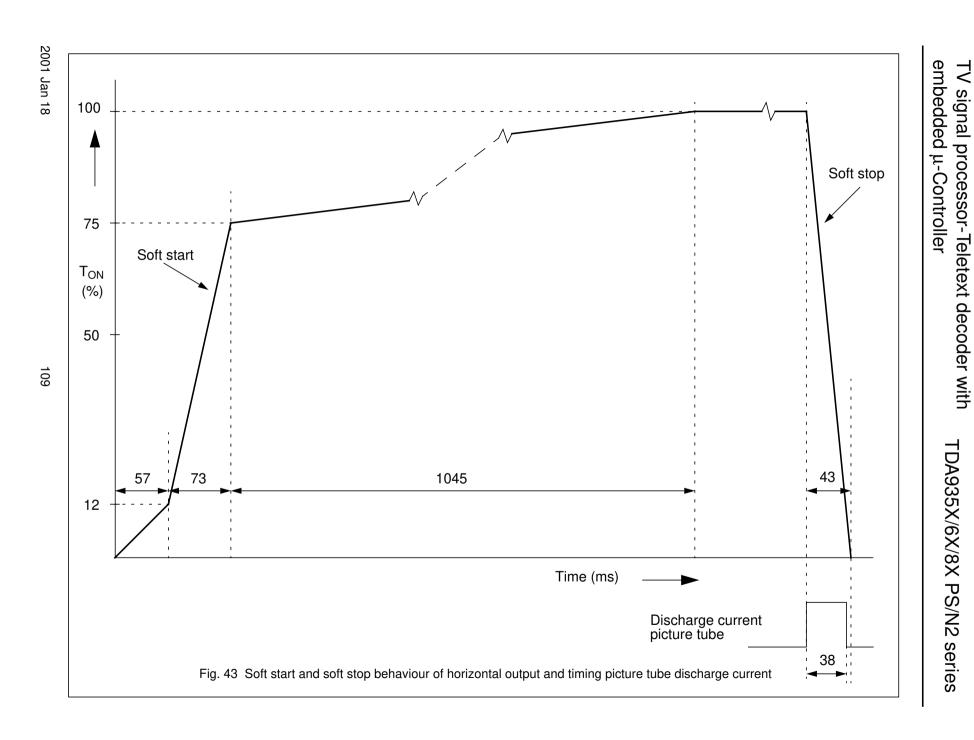


TDA935X/6X/8X PS/N2 series



TDA935X/6X/8X PS/N2 series

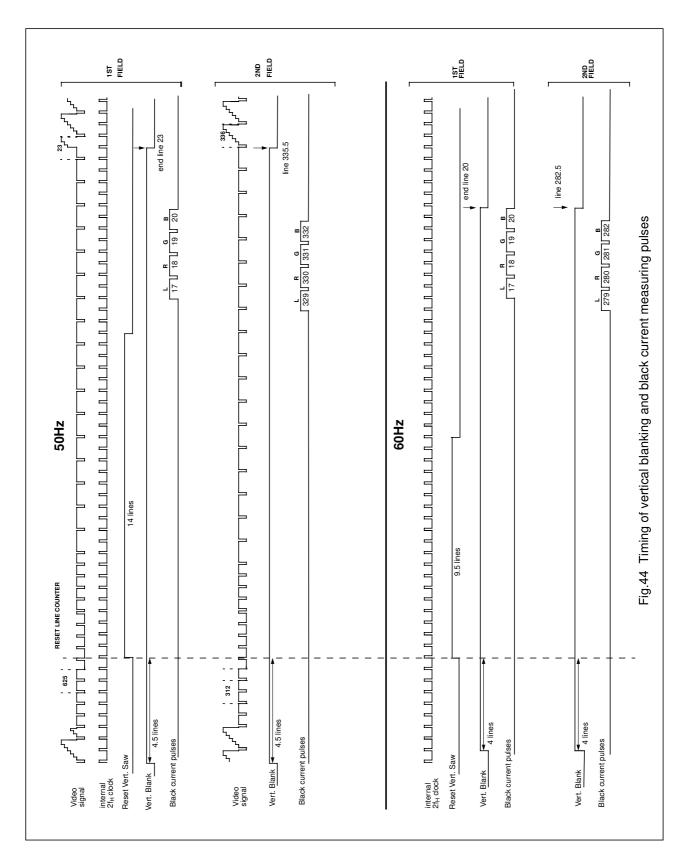




Philips Semiconductors

Tentative Device Specification

TDA935X/6X/8X PS/N2 series



TDA935X/6X/8X PS/N2 series

TEST AND APPLICATION INFORMATION

East-West output stage

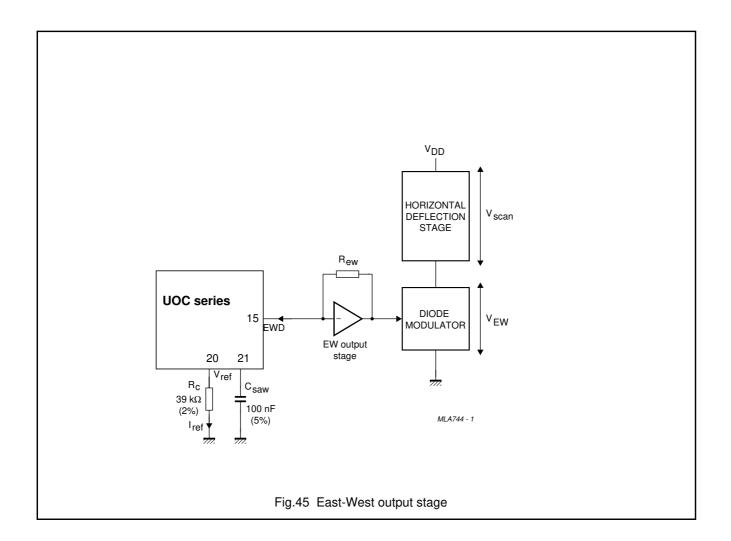
In order to obtain correct tracking of the vertical and horizontal EHT-correction, the EW output stage should be dimensioned as illustrated in Fig.45.

Resistor R_{EW} determines the gain of the EW output stage. Resistor R_c determines the reference current for both the vertical sawtooth generator and the geometry processor. The preferred value of R_c is 39 k Ω which results in a reference current of 100 μA (V $_{ref}$ = 3.9 V).

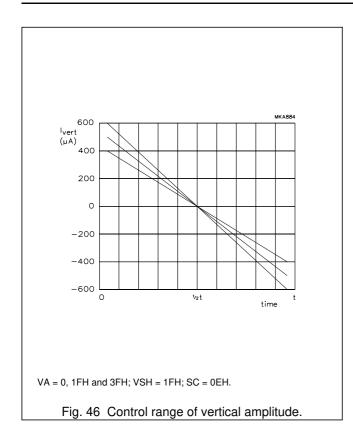
The value of R_{EW} must be:

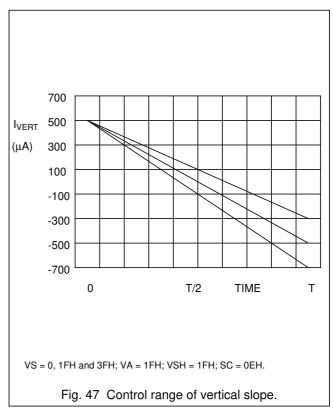
$$R_{EW} = R_c \times \frac{V_{scan}}{18 \times V_{ref}}$$

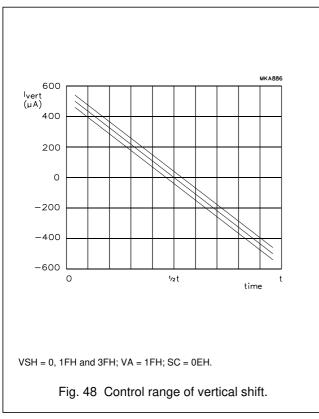
Example: With V_{ref} = 3.9 V; R_c = 39 $k\Omega$ and V_{scan} = 120 V then R_{EW} = 68 $k\Omega.$

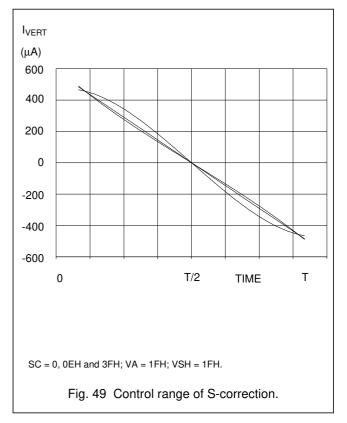


TDA935X/6X/8X PS/N2 series

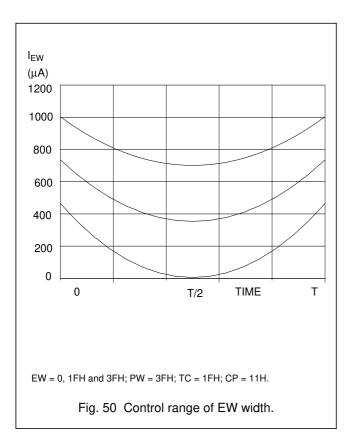


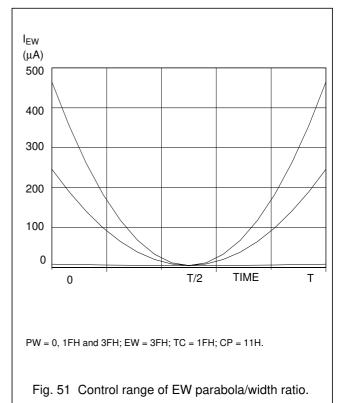


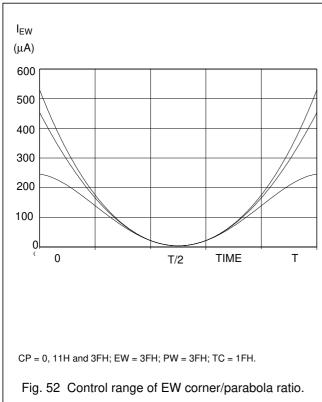


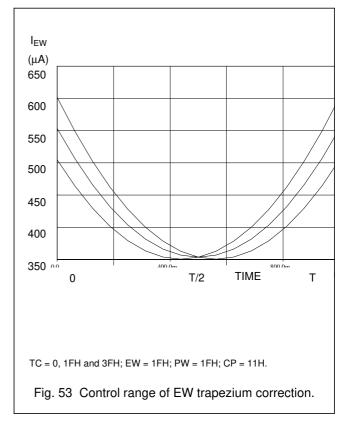


TDA935X/6X/8X PS/N2 series









TDA935X/6X/8X PS/N2 series

Adjustment of geometry control parameters

The deflection processor offers 5 control parameters for picture alignment, viz:

- S-correction
- · vertical amplitude
- vertical slope
- · vertical shift
- · horizontal shift.

The 110° types offer in addition:

- EW width
- · EW parabola width
- EW upper/lower corner parabola
- EW trapezium correction.
- Vertical zoom
- · Horizontal parallelogram and bow correction

It is important to notice that the ICs are designed for use with a DC-coupled vertical deflection stage. This is the reason why a vertical linearity alignment is not necessary (and therefore not available).

For a particular combination of picture tube type, vertical output stage and EW output stage it is determined which are the required values for the settings of S-correction, EW parabola/width ratio and EW corner/parabola ratio. These parameters can be preset via the I²C-bus, and do not need any additional adjustment. The rest of the parameters are preset with the mid-value of their control range (i.e. 1FH), or with the values obtained by previous TV-set adjustments.

The vertical shift control is meant for compensation of off-sets in the external vertical output stage or in the picture tube. It can be shown that without compensation these off-sets will result in a certain linearity error, especially with picture tubes that need large S-correction. The total linearity error is in first order approximation proportional to the value of the off-set, and to the square of the S-correction needed. The necessity to use the vertical shift alignment depends on the expected off-sets in vertical output stage and picture tube, on the required value of the S-correction, and on the demands upon vertical linearity.

For adjustment of the vertical shift and vertical slope independent of each other, a special service blanking mode can be entered by setting the SBL bit HIGH. In this mode the RGB-outputs are blanked during the second half of the picture. There are 2 different methods for alignment of the picture in vertical direction. Both methods make use of the service blanking mode.

The first method is recommended for picture tubes that have a marking for the middle of the screen. With the vertical shift control the last line of the visible picture is positioned exactly in the middle of the screen. After this adjustment the vertical shift should not be changed. The top of the picture is placed by adjustment of the vertical amplitude, and the bottom by adjustment of the vertical slope.

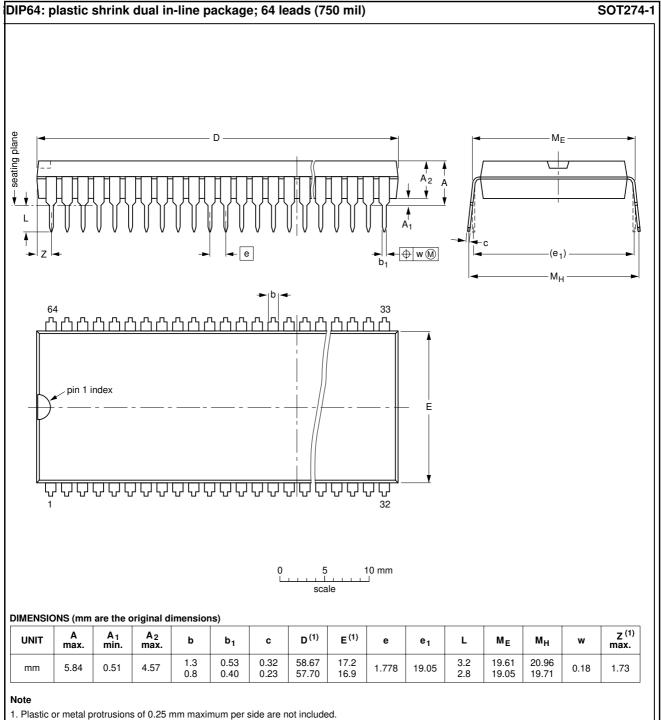
The second method is recommended for picture tubes that have no marking for the middle of the screen. For this method a video signal is required in which the middle of the picture is indicated (e.g. the white line in the circle test pattern). With the vertical slope control the beginning of the blanking is positioned exactly on the middle of the picture. Then the top and bottom of the picture are placed symmetrical with respect to the middle of the screen by adjustment of the vertical amplitude and vertical shift. After this adjustment the vertical shift has the right setting and should not be changed.

If the vertical shift alignment is not required VSH should be set to its mid-value (i.e. VSH = 1F). Then the top of the picture is placed by adjustment of the vertical amplitude and the bottom by adjustment of the vertical slope. After the vertical picture alignment the picture is positioned in the horizontal direction by adjustment of the EW width and the horizontal shift. Finally (if necessary) the left- and right-hand sides of the picture are aligned in parallel by adjusting the EW trapezium control.

To obtain the full range of the vertical zoom function the adjustment of the vertical geometry should be carried out at a nominal setting of the zoom DAC at position 19 HEX.

TDA935X/6X/8X PS/N2 series

PACKAGE OUTLINE



VERSION IEC JEDEC EIAJ PROJECTION	OUTLINE		REFER	ENCES	EUROPEAN	ISSUE DATE	
— + 02 10 :	VERSION	IEC	JEDEC	EIAJ	PROJECTION	ISSUE DATE	
1 S(1) = 1 I I I I I I I I I	SOT274-1					92-10-13 95-02-04	

2001 Jan 18 115

TDA935X/6X/8X PS/N2 series

SOLDERING

Introduction

There is no soldering method that is ideal for all IC packages. Wave soldering is often preferred when through-hole and surface mounted components are mixed on one printed-circuit board. However, wave soldering is not always suitable for surface mounted ICs, or for printed-circuits with high population densities. In these situations reflow soldering is often used.

This text gives a very brief insight to a complex technology. A more in-depth account of soldering ICs can be found in our "IC package Databook" (order code 9398 652 90011).

SDIP

SOLDERING BY DIPPING OR BY WAVE

The maximum permissible temperature of the solder is 260 °C; solder at this temperature must not be in contact with the joint for more than 5 seconds. The total contact time of successive solder waves must not exceed 5 seconds.

The device may be mounted up to the seating plane, but the temperature of the plastic body must not exceed the specified maximum storage temperature ($T_{stg\ max}$). If the printed-circuit board has been pre-heated, forced cooling may be necessary immediately after soldering to keep the temperature within the permissible limit.

REPAIRING SOLDERED JOINTS

Apply a low voltage soldering iron (less than 24 V) to the lead(s) of the package, below the seating plane or not more than 2 mm above it. If the temperature of the soldering iron bit is less than 300 $^{\circ}$ C it may remain in contact for up to 10 seconds. If the bit temperature is between 300 and 400 $^{\circ}$ C, contact may be up to 5 seconds.

QFP

REFLOW SOLDERING

Reflow soldering techniques are suitable for all QFP packages.

The choice of heating method may be influenced by larger plastic QFP packages (44 leads, or more). If infrared or vapour phase heating is used and the large packages are not absolutely dry (less than 0.1% moisture content by weight), vaporization of the small amount of moisture in them can cause cracking of the plastic body. For details, refer to the Drypack information in our "Quality Reference Handbook" (order code 9397 750 00192).

Reflow soldering requires solder paste (a suspension of fine solder particles, flux and binding agent) to be applied to the printed-circuit board by screen printing, stencilling or pressure-syringe dispensing before package placement.

Several methods exist for reflowing; for example, infrared/convection heating in a conveyor type oven. Throughput times (preheating, soldering and cooling) vary between 50 and 300 seconds depending on heating method. Typical reflow peak temperatures range from 215 to 250 °C.

WAVE SOLDERING

Wave soldering is **not** recommended for QFP packages. This is because of the likelihood of solder bridging due to closely-spaced leads and the possibility of incomplete solder penetration in multi-lead devices.

If wave soldering cannot be avoided, for QFP packages with a pitch (e) larger than 0.5 mm, the following conditions must be observed:

- A double-wave (a turbulent wave with high upward pressure followed by a smooth laminar wave) soldering technique should be used.
- The footprint must be at an angle of 45° to the board direction and must incorporate solder thieves downstream and at the side corners.

During placement and before soldering, the package must be fixed with a droplet of adhesive. The adhesive can be applied by screen printing, pin transfer or syringe dispensing. The package can be soldered after the adhesive is cured.

Maximum permissible solder temperature is 260 °C, and maximum duration of package immersion in solder is 10 seconds, if cooled to less than 150 °C within 6 seconds. Typical dwell time is 4 seconds at 250 °C.

A mildly-activated flux will eliminate the need for removal of corrosive residues in most applications.

REPAIRING SOLDERED JOINTS

Fix the component by first soldering two diagonally-opposite end leads. Use only a low voltage soldering iron (less than 24 V) applied to the flat part of the lead. Contact time must be limited to 10 seconds at up to 300 $^{\circ}$ C. When using a dedicated tool, all other leads can be soldered in one operation within 2 to 5 seconds between 270 and 320 $^{\circ}$ C.

TDA935X/6X/8X PS/N2 series

DEFINITIONS

Data sheet status	
Objective specification	This data sheet contains target or goal specifications for product development.
Preliminary specification	This data sheet contains preliminary data; supplementary data may be published later.
Product specification	This data sheet contains final product specifications.
Limiting values	

Limiting values given are in accordance with the Absolute Maximum Rating System (IEC 134). Stress above one or more of the limiting values may cause permanent damage to the device. These are stress ratings only and operation of the device at these or at any other conditions above those given in the Characteristics sections of the specification is not implied. Exposure to limiting values for extended periods may affect device reliability.

Application information

Where application information is given, it is advisory and does not form part of the specification.

LIFE SUPPORT APPLICATIONS

These products are not designed for use in life support appliances, devices, or systems where malfunction of these products can reasonably be expected to result in personal injury. Philips customers using or selling these products for use in such applications do so at their own risk and agree to fully indemnify Philips for any damages resulting from such improper use or sale.

PURCHASE OF PHILIPS I2C COMPONENTS



Purchase of Philips I²C components conveys a license under the Philips' I²C patent to use the components in the I²C system provided the system conforms to the I²C specification defined by Philips. This specification can be ordered using the code 9398 393 40011.

TDA935X/6X/8X PS/N2 series